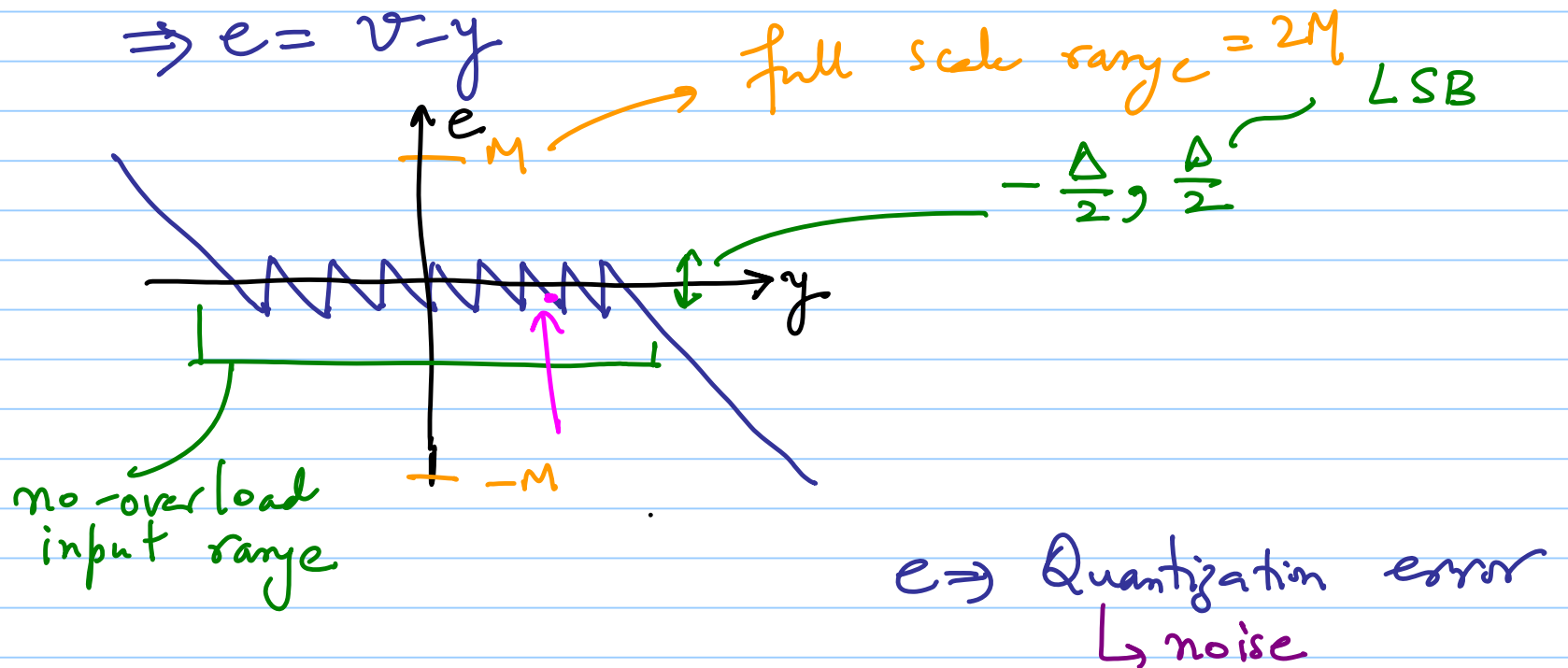


ECE 615 - Lecture 8

Note Title

2/4/2016

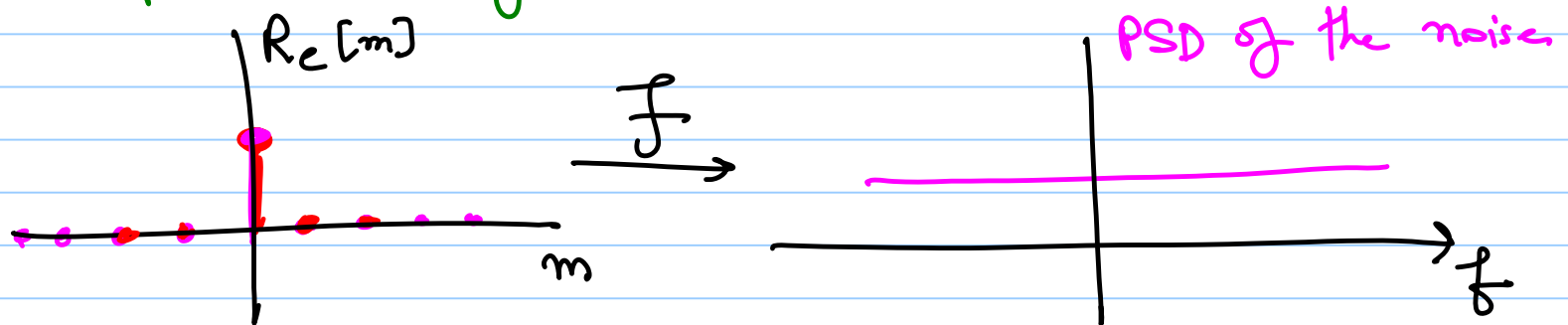


Conditions to model $e[n]$ as noise

① y (input) stays within the no-overload input range

② $e[n]$ is uncorrelated with the input $y[n]$
"input must be busy"

③ Spectrum of $e[n]$ is "white"



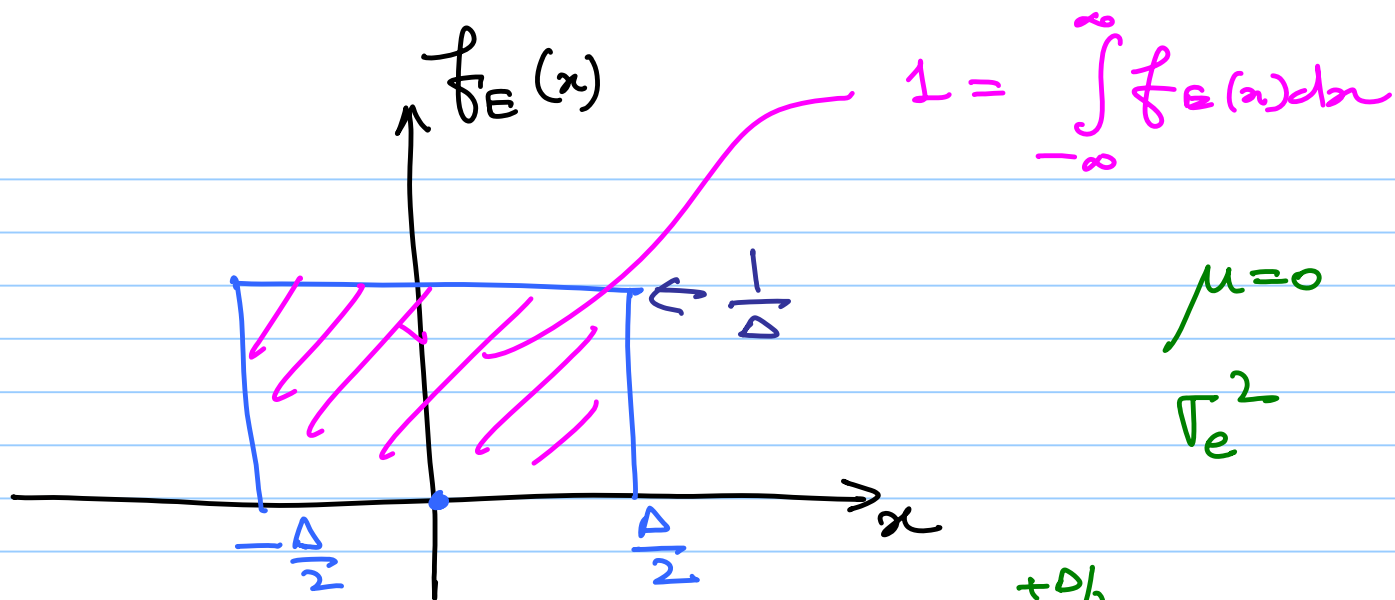
PSD of "e"

$$S_e(f) = \mathcal{F} \left[\underbrace{E(e[n] e[n-m])}_{\text{autocorrelation}} \right] = \delta[m]$$

PSD

④ Quantization error is uniformly distributed

$$e \sim U\left[-\frac{\Delta}{2}, \frac{\Delta}{2}\right]$$



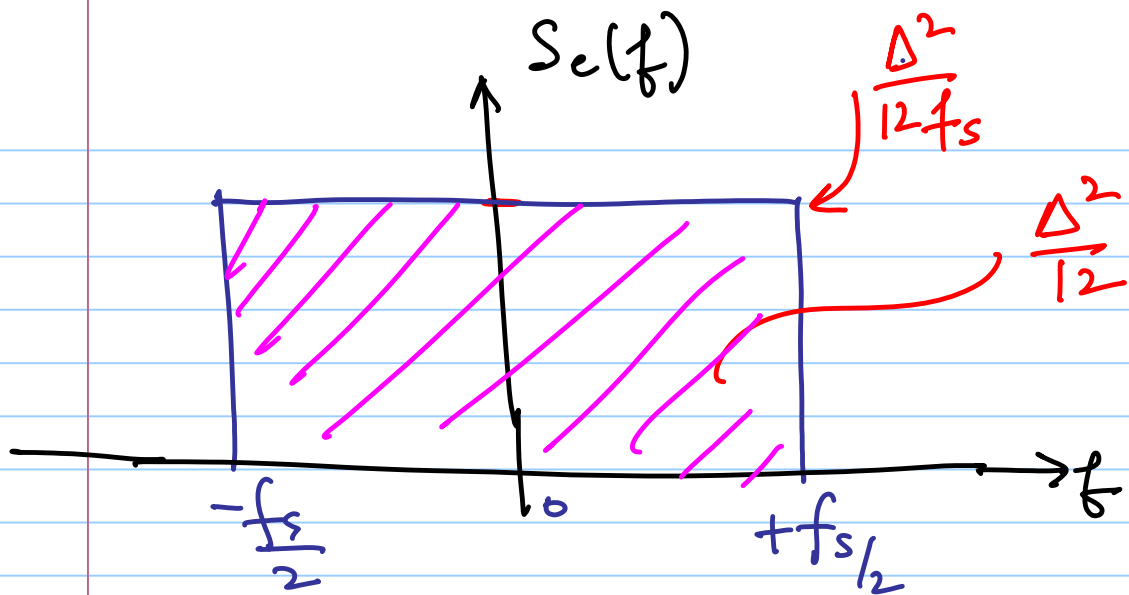
$$\sigma_e^2 = E[(e - \mu)^2] = E[e^2] = \int_{-\Delta/2}^{+\Delta/2} x^2 f_E(x) dx$$

$$= \frac{1}{\Delta} \left[\frac{x^3}{3} \right]_{-\Delta/2}^{+\Delta/2} = \frac{2}{\Delta} \times \frac{x^3}{3} \Big|_0^{+\Delta/2} = \frac{2}{\Delta} \left(\frac{\Delta}{2} \right)^3 \frac{1}{3}$$

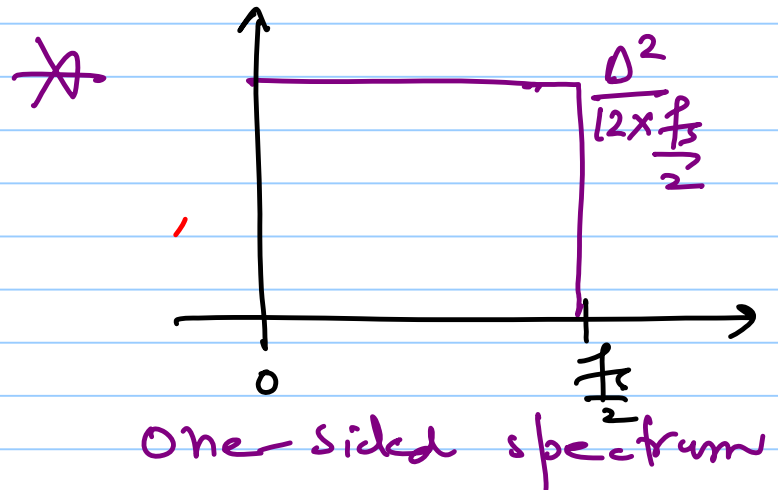
$$= \frac{\Delta^2}{12}$$

$$\text{Noise power} = \text{variance} = \sigma_e^2$$

$$\boxed{\text{Q-noise power} = \frac{\Delta^2}{12}} = \frac{V_{\text{LSB}}^2}{12}$$



Sampled white noise
 distributed from
 $-\frac{f_s}{2}$ to $\frac{f_s}{2}$



$y = A \sin(\omega_{in} t)$ is quantized with a quantizer which has $LSB = \Delta$

Signal-to-quantization-noise ratio

$$SQNR = \frac{\text{Signal power}}{\text{Q-noise power}} = \frac{P_s}{\Delta^2/12}$$

Signal power $\frac{A^2}{2}$
for sine input

* We have an N-bit ADC

$$\cdot \text{full scale range} = 2^N \Delta$$

$$\text{maximum amplitude } (A_{\max}) = \frac{F_S}{2} = 2^{N-1} \Delta$$

$$\text{max. signal power} = \frac{(2^{N-1} \Delta)^2}{2} = \frac{A_{\max}^2}{2}$$

$$\text{Peak SQNR} = \frac{(2^{N-1} \Delta)^2}{2 \times \frac{\Delta^2}{12}} = \frac{2^{2N-2}}{2} \times 12 = \frac{3}{2} \times 2^{2N}$$

full scale
sine input

$$\text{SQNR}_{\text{dB}} = 10 \log_{10} \left(\frac{3}{2} \cdot 2^{2N} \right)$$

$$\text{SQNR}_{\text{dB}} = 6.02N + 1.76 \text{ dB}$$

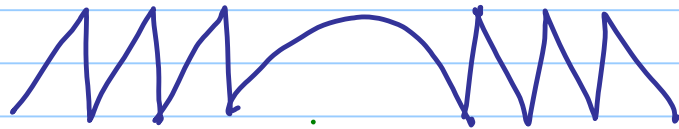
for $\Delta = 0.1 \Rightarrow \# \text{ of periods} = 44$

for $\Delta = 0.2 \Rightarrow \# \text{ of periods} = 20$

When Δ is halved Δ^2

$$\text{tone power} = \left(\frac{1}{20}\right)^2 = \frac{1}{4} = -6 \text{ dB}$$

$e(t)$



$\Delta \downarrow 2$
→



$e(t)$

$e(2t)$

$E(f)$

$\frac{1}{2} E(f/2)$

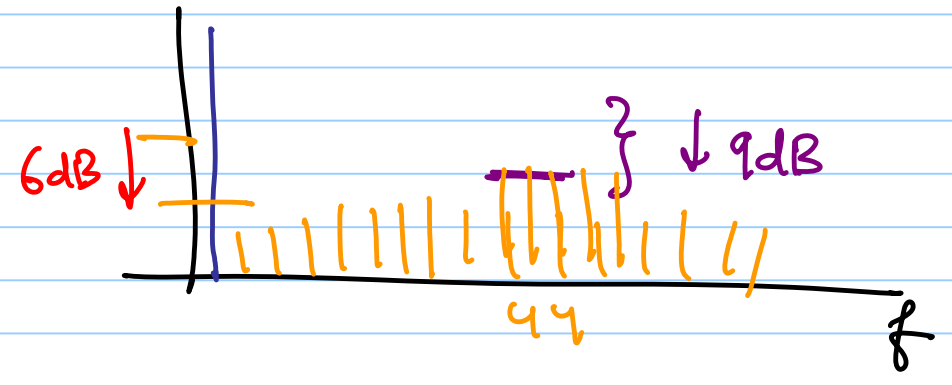
\Rightarrow 3dB reduction in tone power

total tone power reduction = 9dB by halving Δ

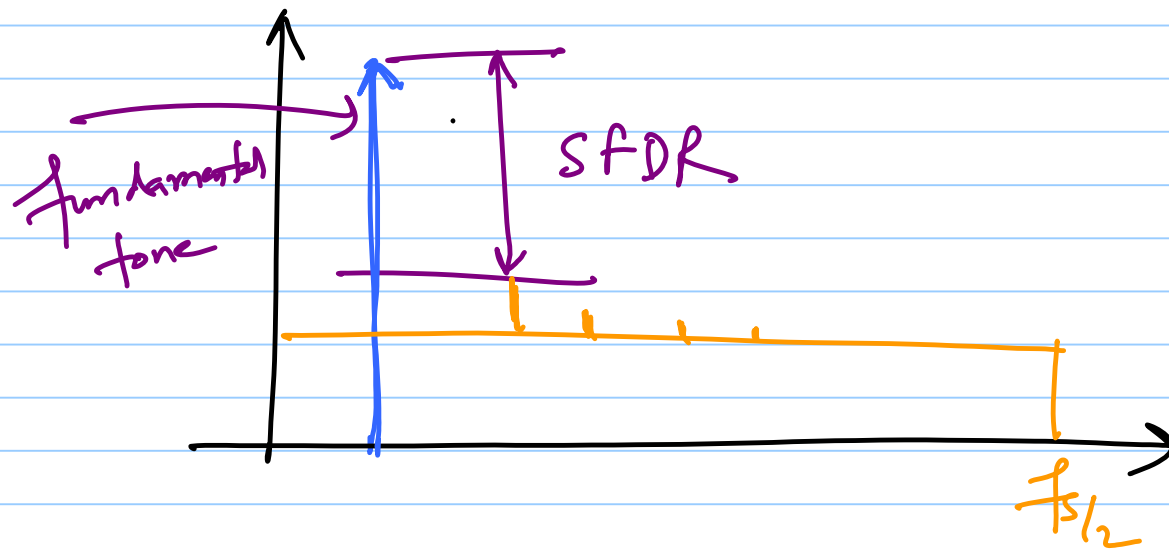
$$\Delta = 0.2$$



$$\Delta = 0.1$$

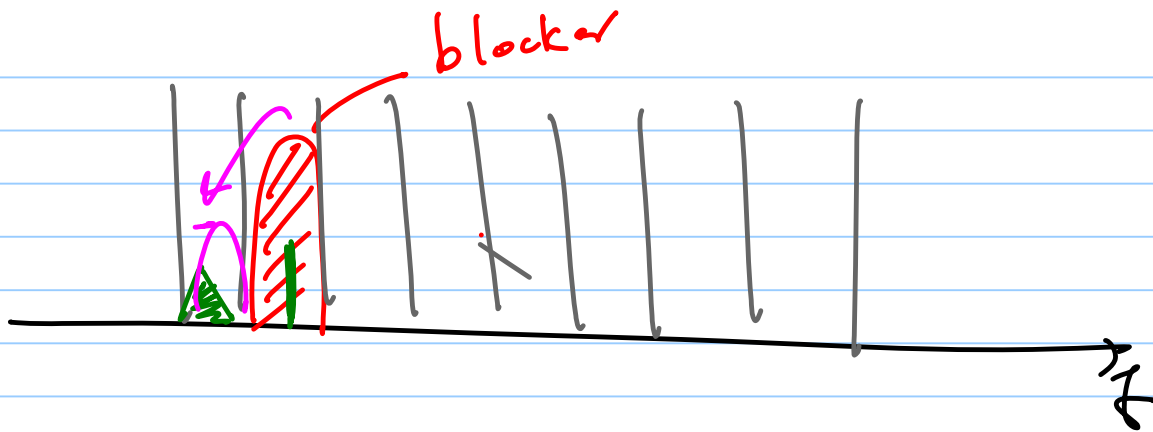


SFDR (Spurious-free dynamic range)



Spurs \Rightarrow spurious tones

SFDR = difference (in dB) b/w the signal tone power & the largest spurious tone "in the bandwidth of interest."



halving Δ by 2 \Rightarrow 9dB better SFDR
 & 6dB higher SQNR

frequency domain measurements

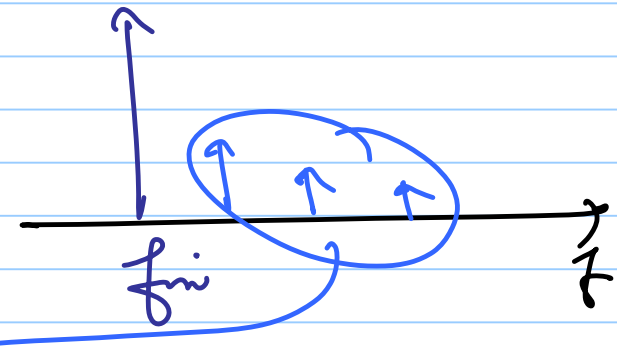
$$SQNR = 10 \log_{10} \left(\frac{P_s}{P_Q} \right)$$

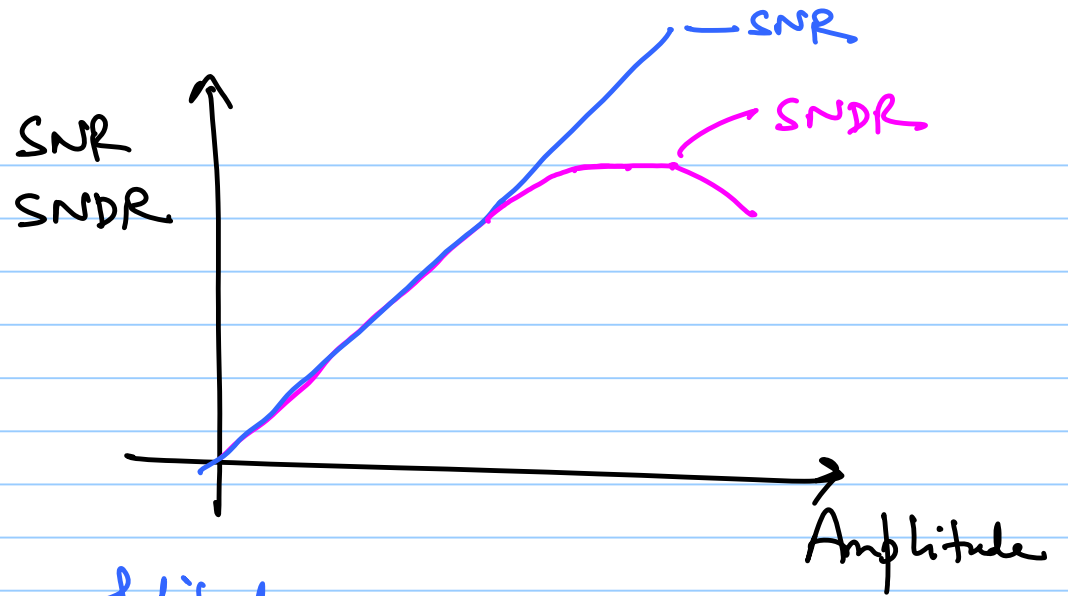
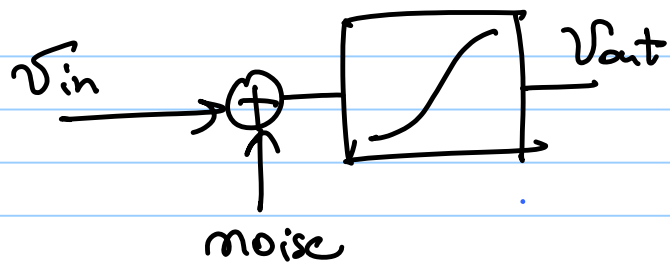
$P_Q \Rightarrow$ quantization noise power

$$SNR = 10 \log_{10} \left(\frac{P_s}{P_N} \right)$$

$$SINAD = 10 \log_{10} \left(\frac{P_s}{P_s + P_{\text{distortion}}} \right)$$

Signal to noise and distortion ratio (SINAD)





@ higher amplitudes
⇒ higher distortion
⇒ lower SNDR

$$\text{ENOB} = \frac{\text{SNDR} - 1.76}{6.02}$$

effective number
of bits

$$\text{SQNR} = 6.02N + 1.76$$
$$N = \frac{\text{SQNR} - 1.76}{6.02}$$

$$\text{SFDR} = 10 \log_{10} \left(\frac{\text{Signal power}}{\text{largest spurious power}} \right)$$

$$\text{Dynamic Range (DR)} = 10 \log_{10} \left(\frac{\text{Maximum signal power detected}}{\text{Smallest signal power}} \right)$$
$$= 10 \log_{10} \left(\frac{A_{\text{max}}}{A_{\text{min}}} \right) \text{ detected}$$

