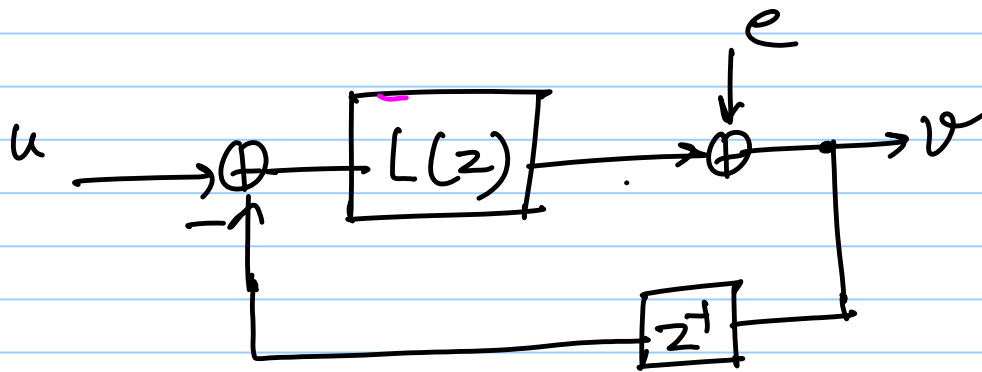


ECE 615 - Lecture 10

Note Title

2/11/2016



$$|L(z)| \rightarrow \infty \text{ at } z=1$$
$$\Rightarrow \omega=0$$
$$\Rightarrow \text{DC}$$

$|u-v| \rightarrow 0$ in the signal band

$$V(z) = \underbrace{\frac{L(z)}{1+z^{-1}L(z)}}_{\text{STF}(z) \text{ signal transfer function}} \cdot u(z) + \underbrace{\frac{1}{1+z^{-1}L(z)}}_{\text{NTF}(z) : \text{noise transfer function}} \cdot e(z)$$

Uncorrelated with $u(z)$

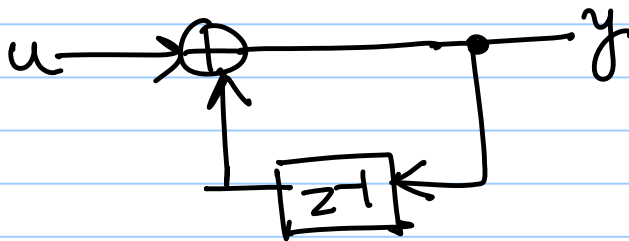
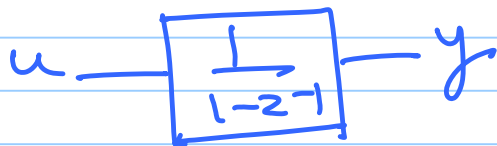
Let $L(z) = \frac{1}{1-z^{-1}}$

$$\frac{Y(z)}{U(z)} = \frac{1}{1-z^{-1}}$$

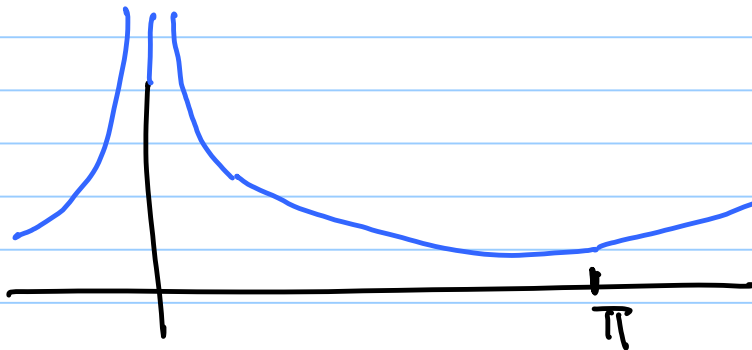
$$\Rightarrow Y(z) - z^{-1}Y(z) = U(z)$$

$$\Rightarrow y(n) - y(n-1] = u(n)$$

$$\Rightarrow y(n] = y(n-1] + u(n)$$



accumulator
discrete integrator

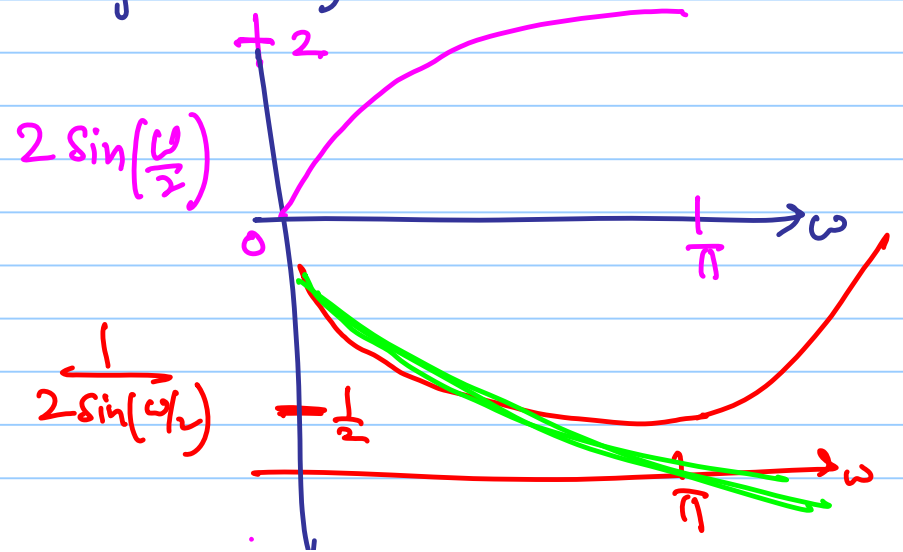


$$X(z) = \frac{1}{1-z^{-1}}$$

$$X(e^{j\omega}) = \frac{1}{1-e^{-j\omega}} = \frac{1}{e^{j\frac{\omega}{2}} (e^{j\frac{\omega}{2}} - e^{-j\frac{\omega}{2}})}$$

$$= e^{j\frac{\omega}{2}} \cdot \frac{1}{2j \sin(\omega/2)}$$

$$|X(e^{j\omega})| = \frac{1}{2 \sin(\omega/2)}$$



$$STF(z) = \frac{L(z)}{1+z^{-1}L(z)} = \frac{1}{1+z^{-1}} = \frac{1}{1+z^{-1}} = \frac{1}{1+z^{-1}} = 1$$

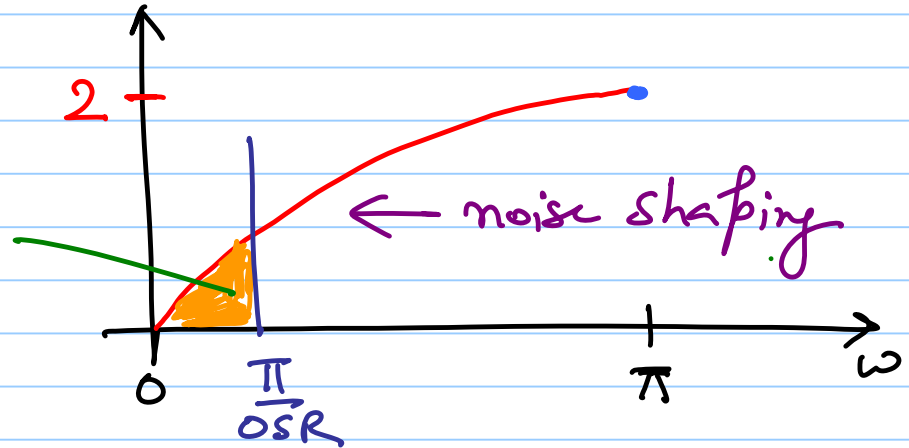
$$NTF(z) = \frac{1}{1+z^{-1}L(z)} = \frac{1}{1+z^{-1}} = 1-z^{-1}$$

$$\boxed{NTF(z) = 1-z^{-1}} \quad \& \quad STF = 1$$

$$NTF(e^{j\omega}) = 1 - e^{-j\omega} = e^{-\frac{j\omega}{2}} \times 2j \sin\left(\frac{\omega}{2}\right)$$

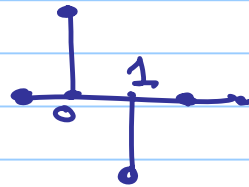
$$|WTF(e^{j\omega})| = |2 \sin(\omega/2)| \leftarrow \text{High-pass response}$$

(in-band noise) IBN



$$WTF(z) = 1 - z^{-1}$$

$$h[n] = [1, -1]$$



$$\sum_n h[n] = 0 \rightarrow H(e^{j0})$$

$$\sum_n (-1)^n h[n] = 2 \rightarrow H(e^{j\pi})$$

intuitively HPF

output quantization noise

$$E(z) \cdot \text{NTF}(z) \\ e[n] \otimes h[n]$$

PSD of o/p quantization noise

$$\begin{aligned} S_{n_o}(\omega) &= S_e(\omega) \cdot \underbrace{|\text{NTF}(e^{j\omega})|^2}_{\text{NTF magnitude squared}} \\ &= \frac{\Delta^2}{12 \cdot \pi} \cdot |2 \sin(\omega/2)|^2 \end{aligned}$$

$$\begin{aligned} \text{for } \omega < \frac{\pi}{\text{OSR}} \\ \Rightarrow \omega \ll \pi \end{aligned}$$

$$\approx \boxed{\frac{\Delta^2}{12\pi} \cdot \omega^2} \quad \text{at low frequencies}$$



noise → $S_y(\omega) = S_x(\omega) |H(e^{j\omega})|^2$
power spectral density (PSD)

Inband Noise

$$IBN = \int_0^{\pi/OSR} \frac{\Delta^2}{12\pi} \omega^2 d\omega$$

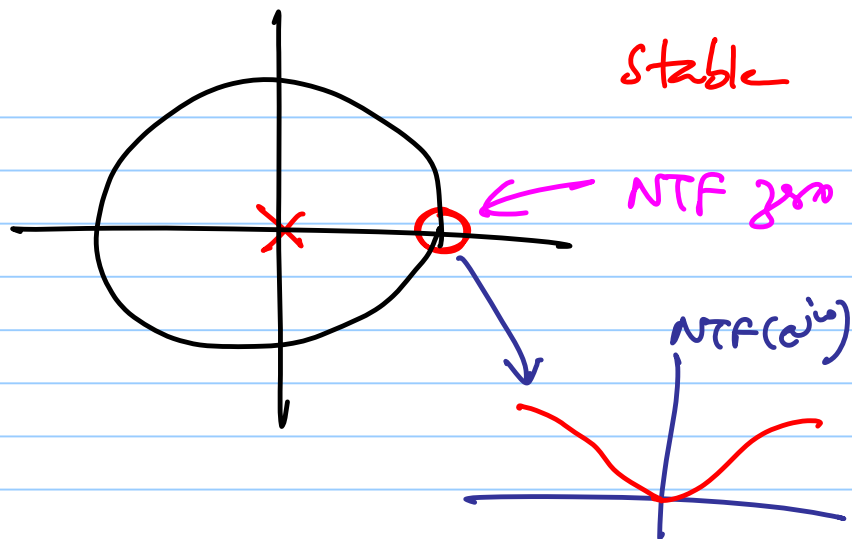
$$= \frac{\Delta^2}{12\pi} \cdot \left. \frac{\omega^3}{3} \right|_0^{\pi/OSR} = \frac{\Delta^2}{12\pi} \left(\frac{\pi}{OSR} \right)^3 \cdot \frac{1}{3}$$

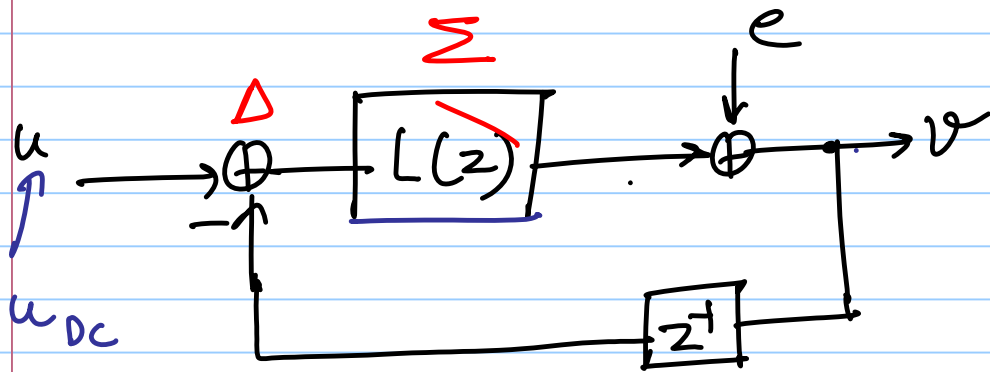
$$= \frac{\Delta^2 \pi^2}{36 \cdot OSR^3}$$

$$IBN = \frac{\Delta^2 \pi^2}{36} \cdot OSR^{-3}$$

$\frac{\Delta^2}{12} OSR^{-1}$ for
plain
oversampling

$$\begin{aligned} \text{NTF}(z) &= 1 - z^{-1} \\ &= \frac{z - 1}{z} \end{aligned}$$





$$L(z) = \frac{1}{1-z^{-1}} \leftarrow 1^{\text{st}}\text{-order}$$

1st-order modulator

$$\bar{u} = \lim_{N \rightarrow \infty} \frac{1}{N} \sum_n v[n]$$

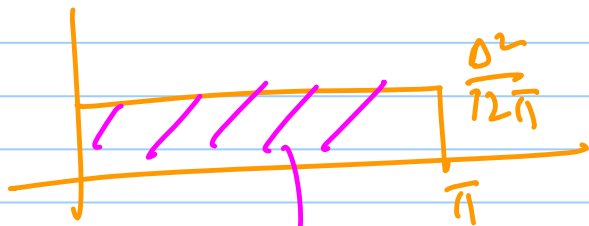
oversampling

$$IBN = \frac{\Delta^2}{12} OSR^{-1}$$

\Rightarrow $2 \times OSR$ Noise \downarrow 3dB

\Rightarrow SNR \uparrow by 3dB

\Rightarrow ENOB \uparrow 0.5 bits



Total noise = $\frac{\Delta^2}{12}$

first-order noise-shaping

$$IBN = \frac{\Delta^2 \pi^2}{36} OSR^{-3}$$

\Rightarrow Noise \downarrow 9dB

SNR \uparrow 9dB

ENOB \Rightarrow 1.5 bits per

doubling with OSR

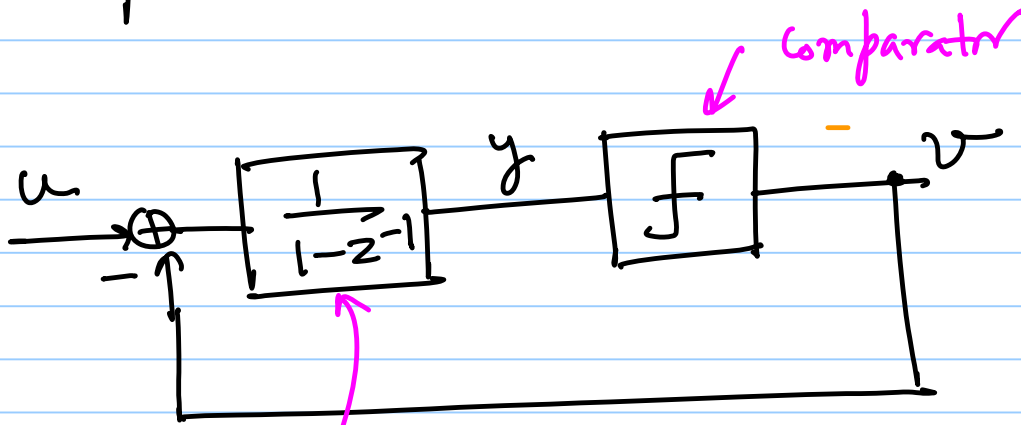
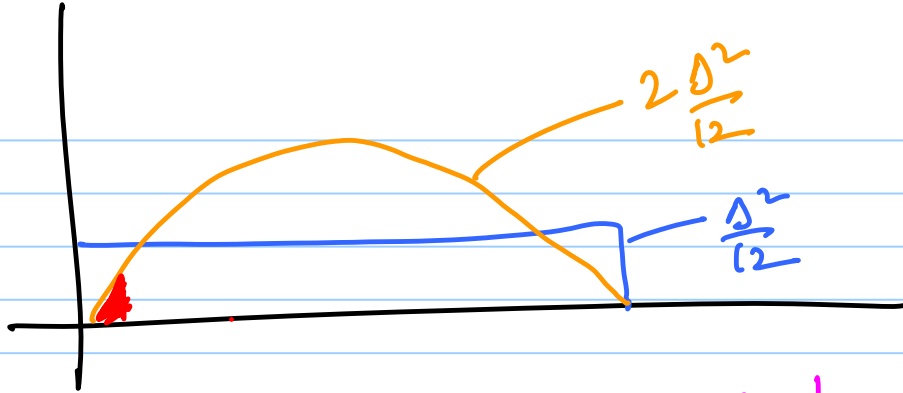
Total noise

$$= \frac{\Delta^2}{12\pi} \int_0^\pi |NTP(e^{j\omega})|^2 d\omega$$

Parseval's Theorem

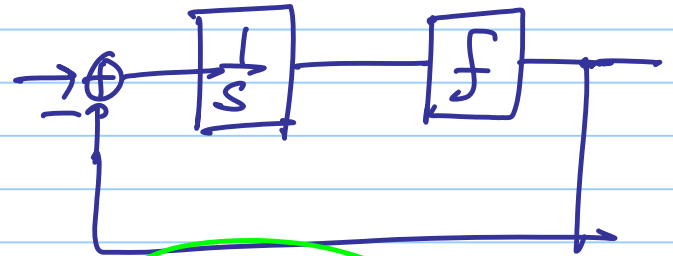
$$= \frac{\Delta^2}{12\pi} \left(\pi \sum_n |h(n)|^2 \right)$$

$$= \frac{\Delta^2}{12\pi} \times 2\pi = 2 \left(\frac{\Delta^2}{12} \right)$$

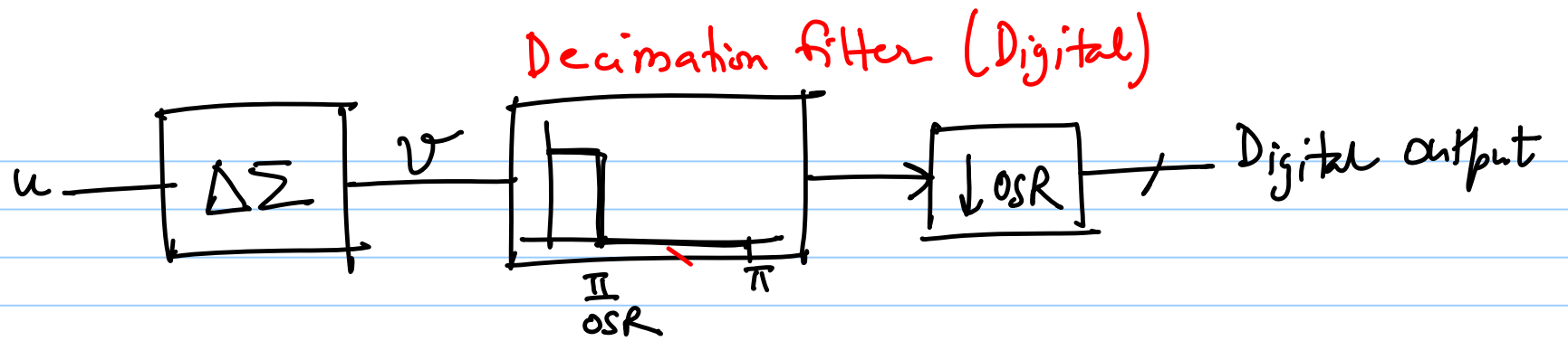


Switched Capacitor

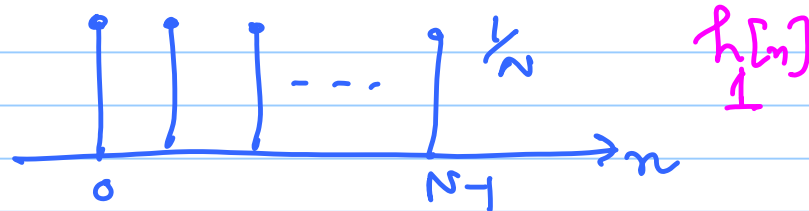
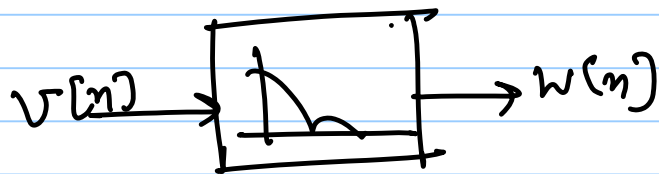
Discrete-time $\Delta\Sigma$ Integrator



CT- $\Delta\Sigma$



FIR filter



$$w(n) = \sum_{i=-\infty}^{\infty} h[i] v[n-i]$$

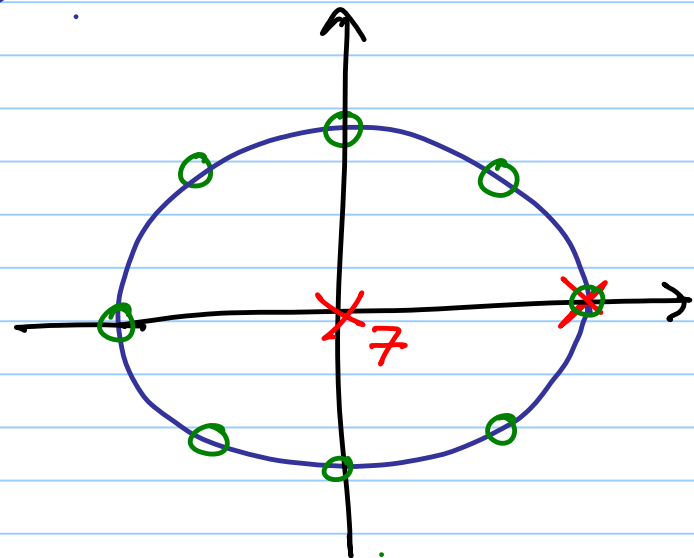
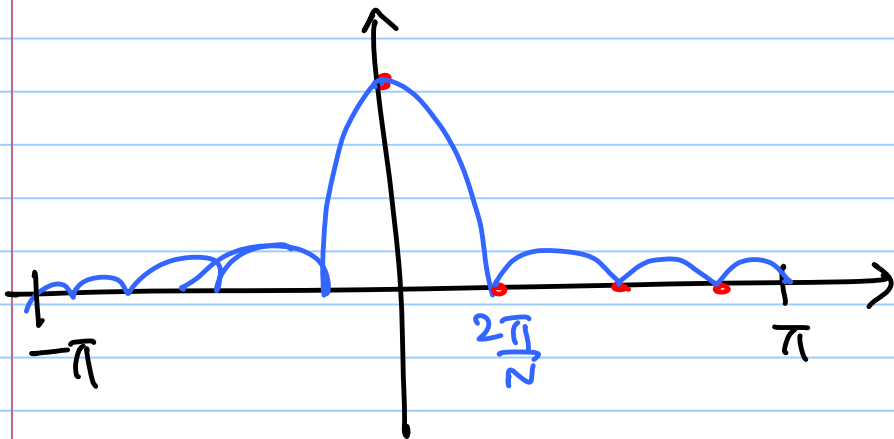
$$= \frac{1}{N} \sum_{i=0}^{N-1} v[n-i] \rightarrow \text{moving average filter over the modulator output.}$$

$$h_1[n] = \begin{cases} \frac{1}{N}, & 0 \leq n \leq N-1 \\ 0, & \text{otherwise} \end{cases}$$

$$H_1(z) = \frac{1}{N} \left(\frac{1-z^{-N}}{1-z^{-1}} \right)$$

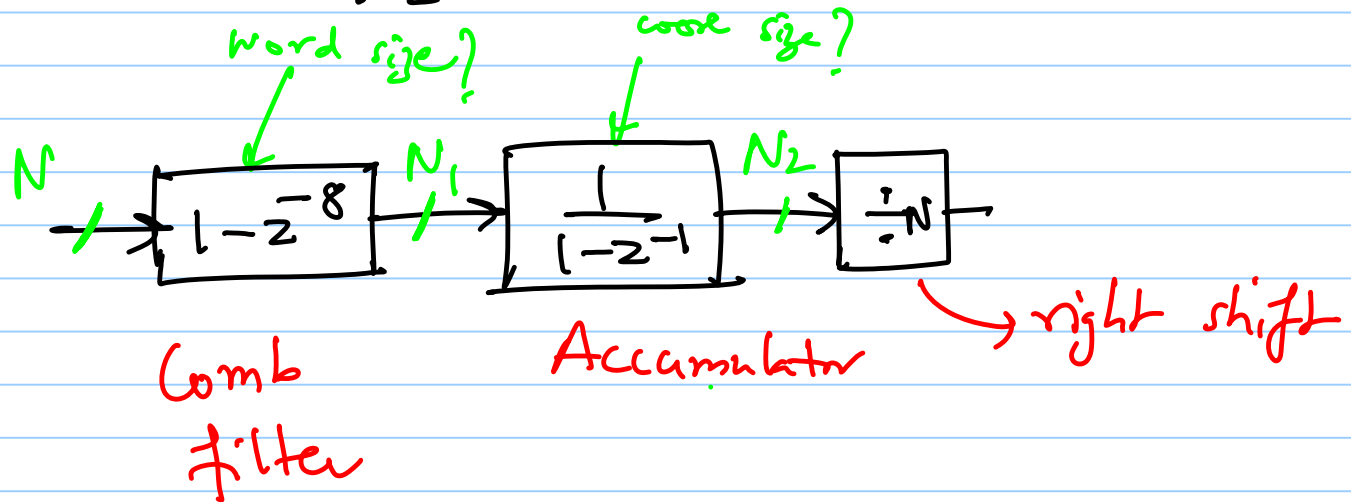
$$z^N - 1 = 0$$

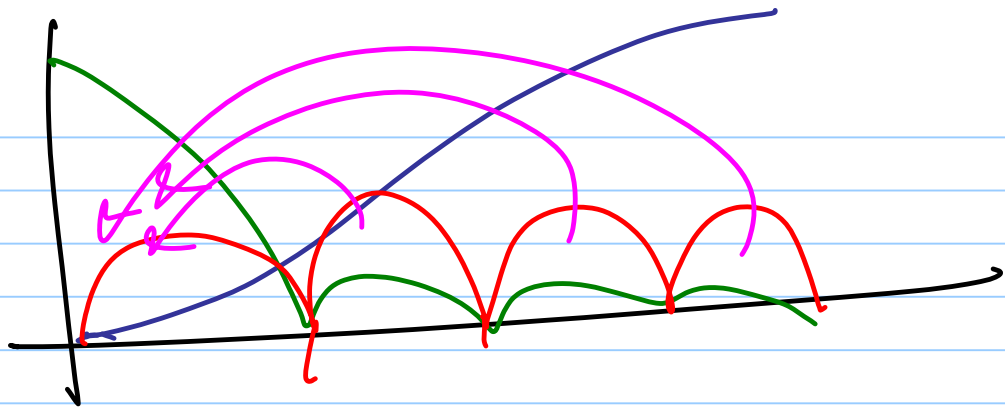
for $N=8$



$$H_1(z) = \frac{1}{2} \cdot \frac{1-z^{-N}}{1-z^{-1}}$$

$N=8$





"Sinc filters"