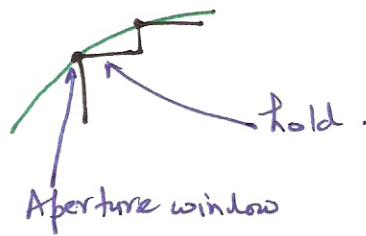
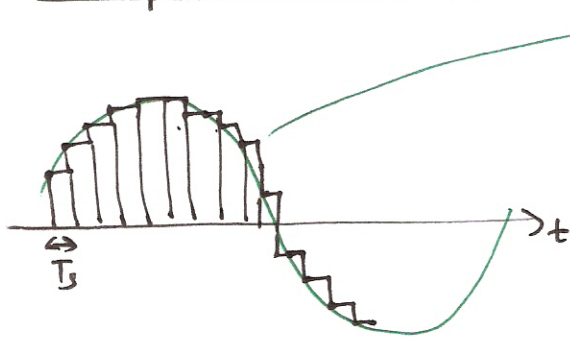


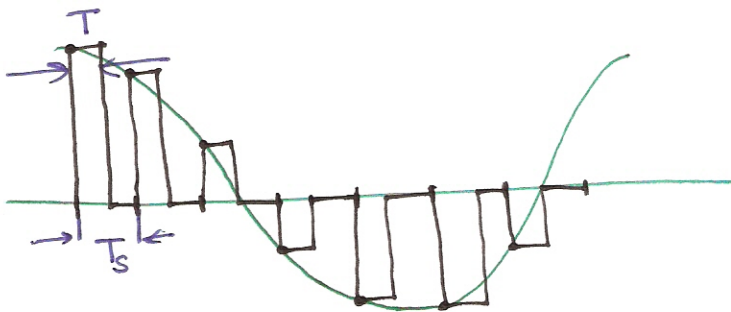
Sample and hold :



Also called zero-order hold (ZOH).
for NRZ pulse shape.

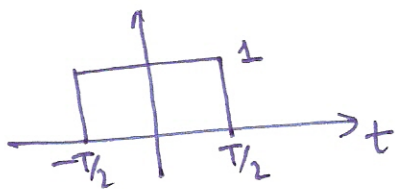
Ideal S/H \rightarrow aperture window is sufficiently narrow w.r.t T_s

Generalized S/H



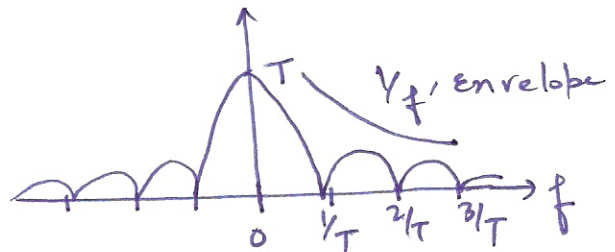
RZ pulse width $\rightarrow T$
sampling period $\rightarrow T_s$
 $0 < T \leq T_s$

* Recap on signals



$\text{rect}\left(\frac{t}{T}\right)$

$\longleftrightarrow f$



$T \text{sinc}(fT)$,

$\text{sinc}(x) = \frac{\sin(\pi x)}{\pi x}$

We have:

$$y(t) = \sum_{n=-\infty}^{\infty} x(nT_s) \cdot \text{srect}\left(\frac{t - T/2 - nT_s}{T}\right)$$

$$= \sum_{n=-\infty}^{\infty} [x(t) \cdot \delta(t - nT_s)] \otimes \text{srect}\left(\frac{t - T/2}{T}\right)$$

← Think Here!

$$= \left[x(t) \cdot \underbrace{\sum_{n=-\infty}^{\infty} \delta(t - nT_s)}_{p(t)} \right] \otimes \underbrace{\text{srect}\left(\frac{t - T/2}{T}\right)}_{h(t)}$$

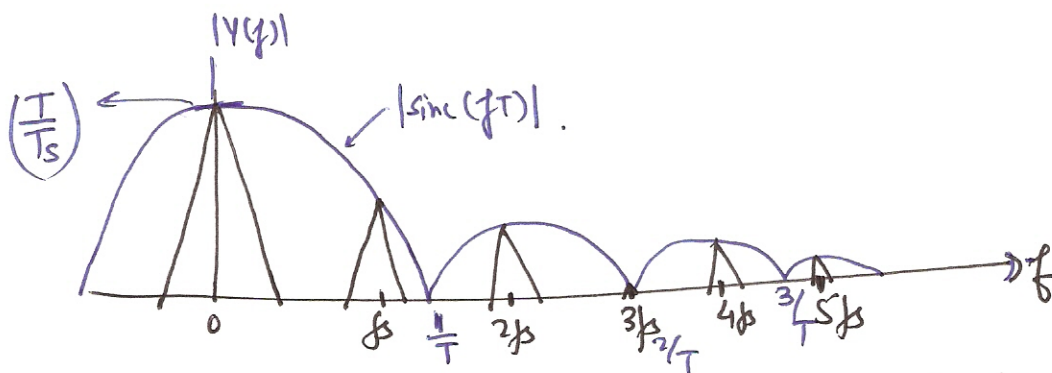
$$= [x(t) \cdot p(t)] \otimes h(t)$$

$$\Rightarrow Y(f) = [X(f) \otimes P(f)] \cdot H(f)$$

$$\begin{aligned} \rightarrow H(f) &= \mathcal{F}\left(\text{srect}\left(\frac{t - T/2}{T}\right)\right) \\ &= T \text{sinc}(fT) \cdot e^{-j\pi fT} \end{aligned}$$

$$\Rightarrow |H(f)| = T |\text{sinc}(fT)|$$

$$\Rightarrow Y(f) = \left(\frac{T}{T_s} \sum_{k=-\infty}^{\infty} X(f - kf_s) \right) \cdot \underbrace{\text{sinc}(fT)}_{\text{sinc distortion!}} \cdot e^{-j\pi fT}$$

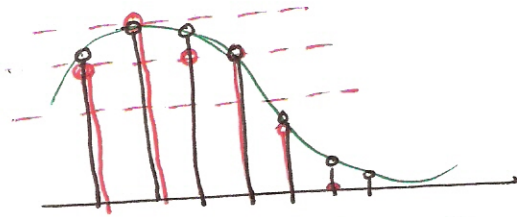
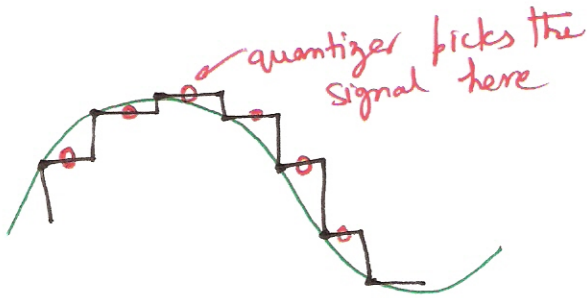


The replicas are weighted by the sinc response.

for $T = T_s \Rightarrow \Sigma \text{H} \rightarrow$ worst sine distortion
for $\frac{T}{T_s} \rightarrow 0$, sine distortion vanishes but the output signal power of the S/H diminishes.

* Is the S/H's sine distortion a problem in an ADC with a S/H in the front-end ??

Ans \rightarrow No!



In an ADC, the quantizer senses the output of the front-end S/H only during the hold mode.

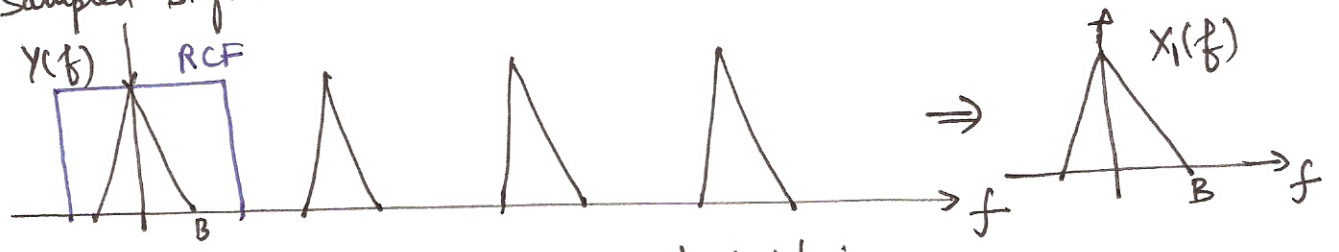
\rightarrow the quantized value only corresponds to the sampled points on the input

\Rightarrow Not an issue in the ADC!

BUT, the sine distortion is an issue in a scenario where the sampler is followed by a DAC. (post-processing a DAC's output).

Reconstruction:

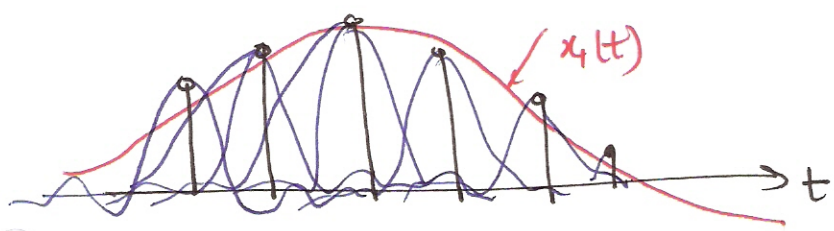
Sampled signal



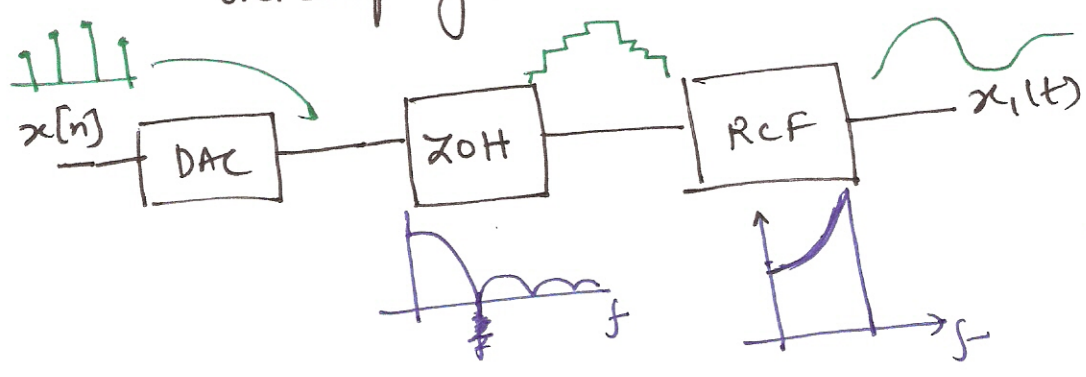
in time domain its a sine interpolation

$$Y(f) \cdot \text{rect}\left(\frac{f}{B}\right) \xleftrightarrow{f^{-1}} y(t) \otimes \text{sinc}(tB)$$

Using, $\text{Sinc}(tB) \xleftrightarrow{f} \text{rect}\left(\frac{f}{B}\right)$
Duality property

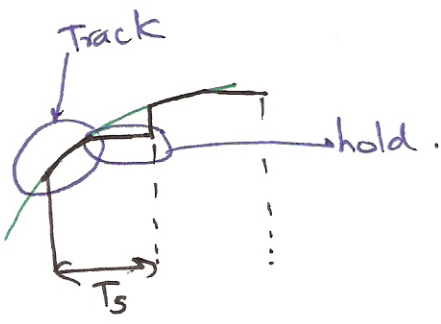
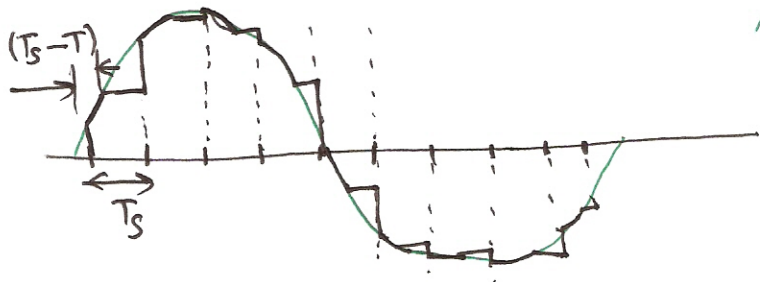


Again, the requirements on the RCF are relaxed by oversampling.



sinc^{-1} response shape is used in the RCF to compensate for the sinc distortion in the ZOH

Track and hold (T/H).



$y(t)$ follows $x(t)$ during the track (ata acquisition) phase and is held during the hold phase

At high speeds (100 MHz \rightarrow 10 GHz)
 the aperture time increases w.r.t the sample period
 \Rightarrow distinction between S/H and T/H disappears at such speeds.
 Ex. 10 GHz ADC all employ T/H's in the front-end.

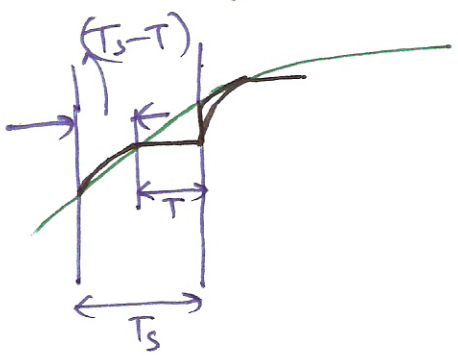
Analysis of T/H

$$y(t) = y_{\text{Track}}(t) + y_{\text{Hold}}(t)$$

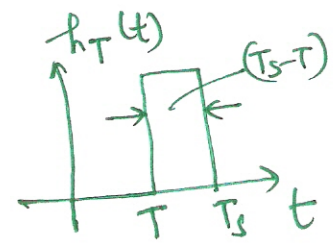
← summation of two responses signals

- * $y_{\text{Hold}}(t)$ is same as in the RZ S/H response
- * need to find $y_{\text{Track}}(t)$

* Note that $y_{\text{Track}}(t) = x(t) \cdot \left[h_{\text{T}}(t) \otimes \sum_{n=-\infty}^{\infty} \delta(t - nT_s) \right]$.



hold lasts for 'T'
 track for '(Ts - T)'



$$\Rightarrow h_{\text{T}}(t) = \text{rect} \left(\frac{t - \left(\frac{T+T_s}{2}\right)}{(T_s - T)} \right)$$

REMAINING IS A HW PROBLEM!

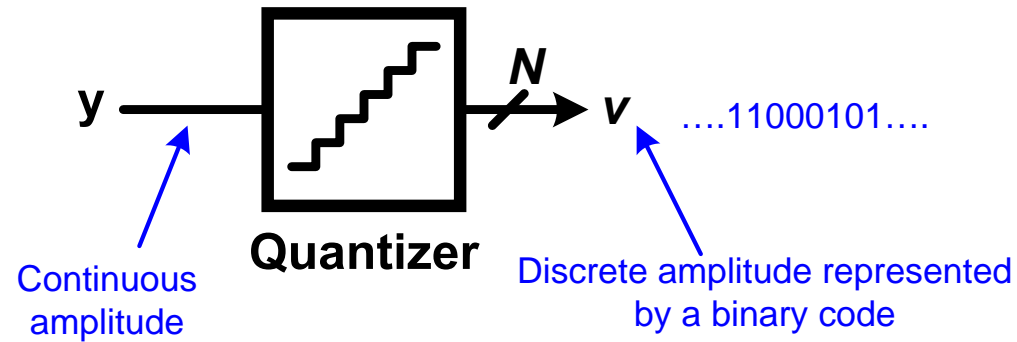
ECE 697 Delta-Sigma Converters Design

Lecture#2 Slides

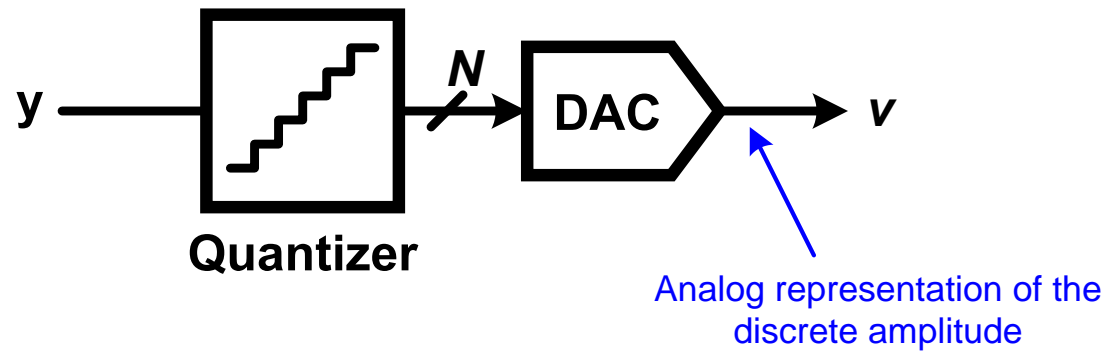
Vishal Saxena

(vishalsaxena@u.boisestate.edu)

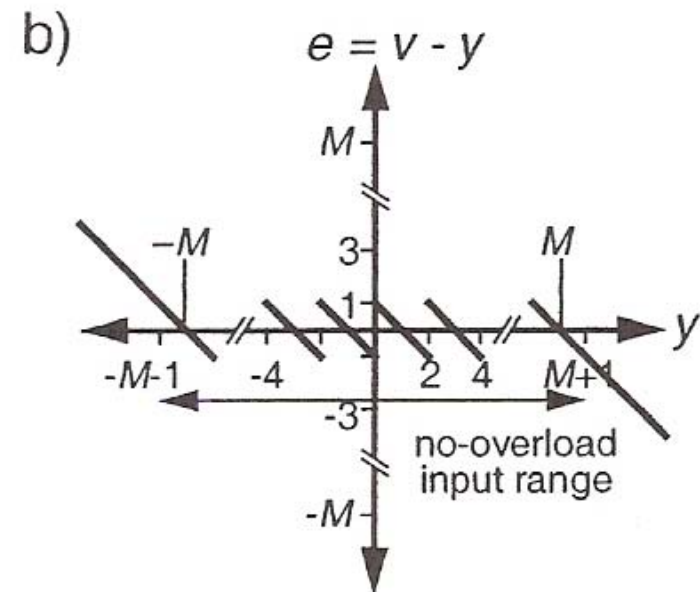
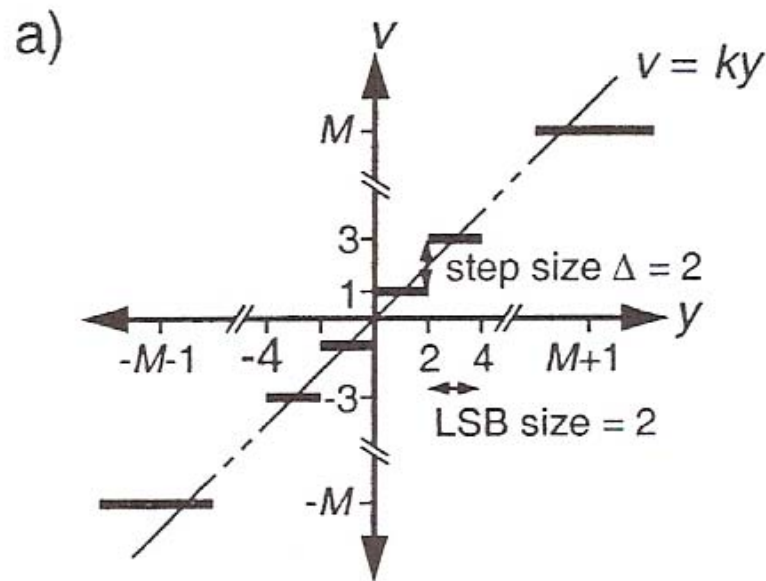
Quantizer



Modeling

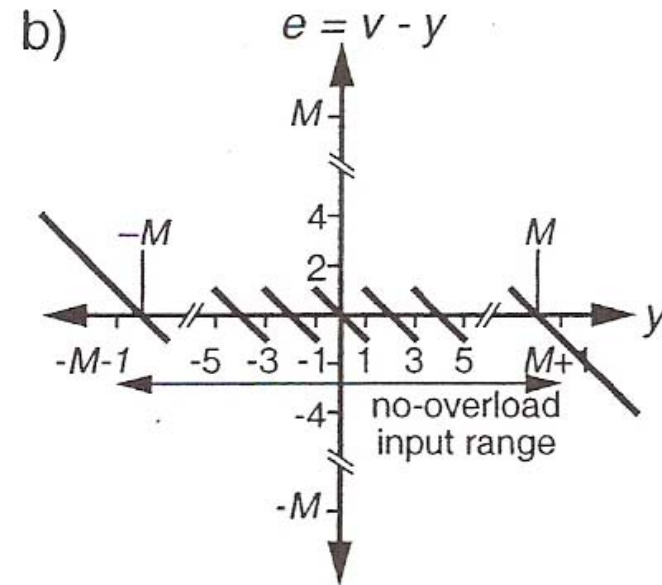
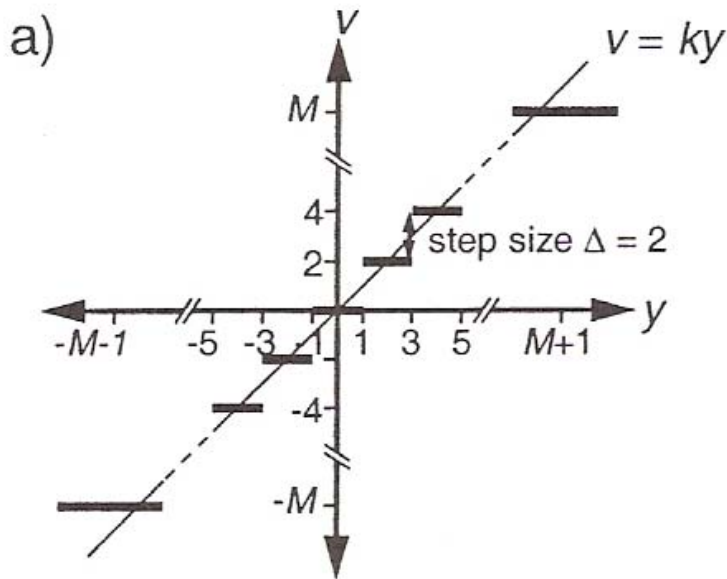


Mid-Rise Quantizer (even number of levels)



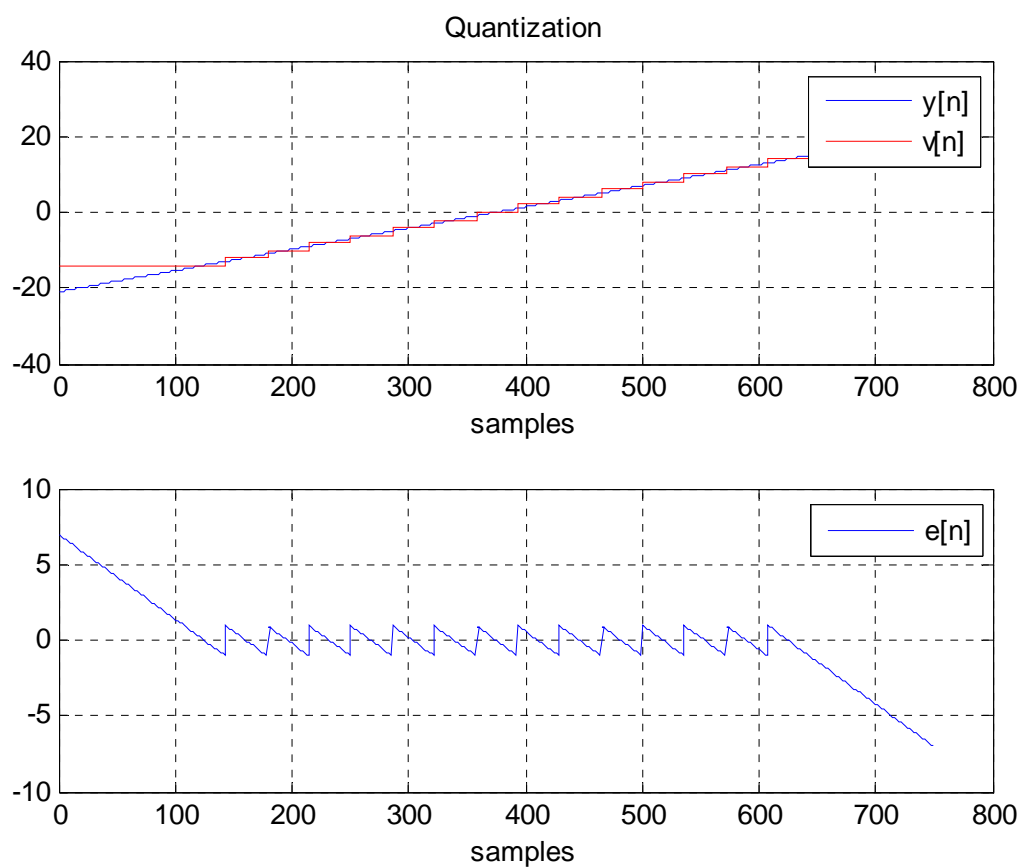
- ❑ Step rising at $y=0$ (mid-rise).
- ❑ In this figure (DSM toolbox model), $\text{LSB} = \Delta = 2$
- ❑ $M = \text{Number of steps}$, (M is odd here)
 - ✓ Number of levels ($n\text{Lev}$) = $M+1$, (even)
- ❑ Input thresholds: $0, \pm 2, \dots, \pm(M-1)$.
- ❑ Output levels: $\pm 1, \pm 3, \dots, \pm M$.

Mid-Tread Quantizer (odd number of levels)



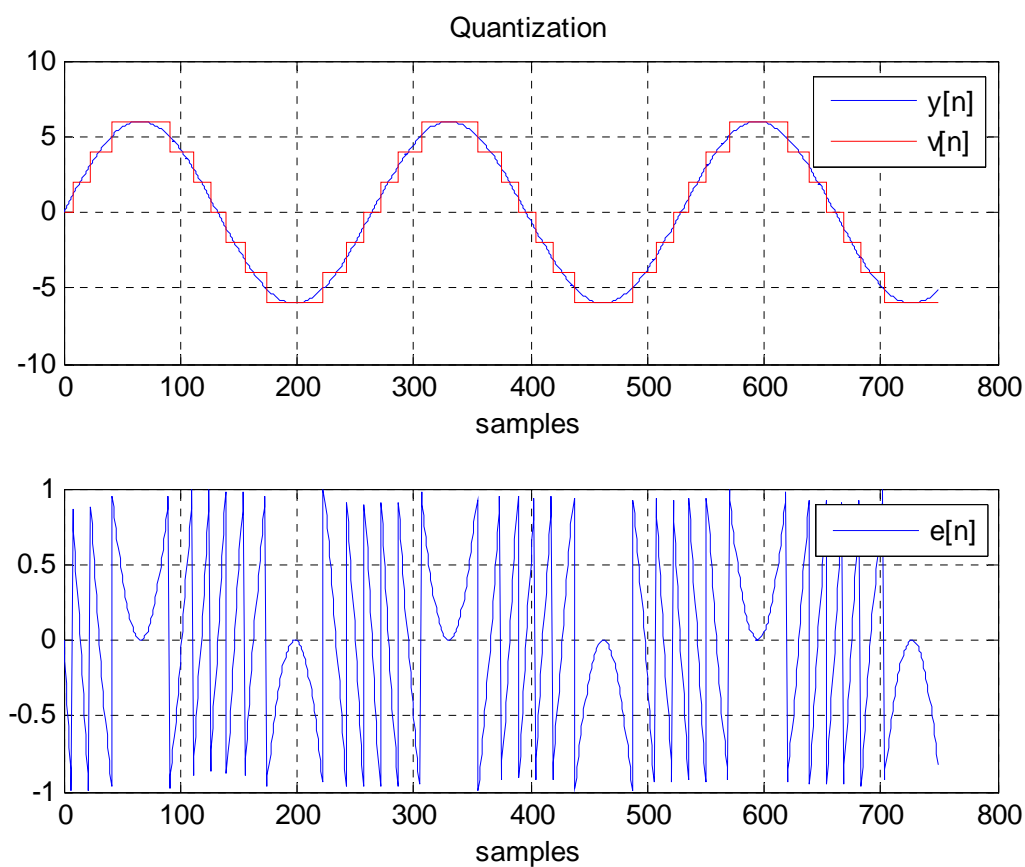
- ❑ Flat part of the step at $y=0$ (mid-tread).
- ❑ Here, $\text{LSB} = \Delta = 2$
- ❑ $M = \text{Number of steps}$, (M is even here)
 - ✓ Number of levels ($n\text{Lev}$) = $M+1$, (odd)
- ❑ Input thresholds: $0, \pm 2, \dots, \pm(M-1)$.
- ❑ Output levels: $0, \pm 2, \pm 4, \dots, \pm M$.

Quantizer characteristics : Slow ramp input



File: Quantizer_ramp_input.m

Quantizer characteristics : Sine input



File: Quantizer_sine_input.m