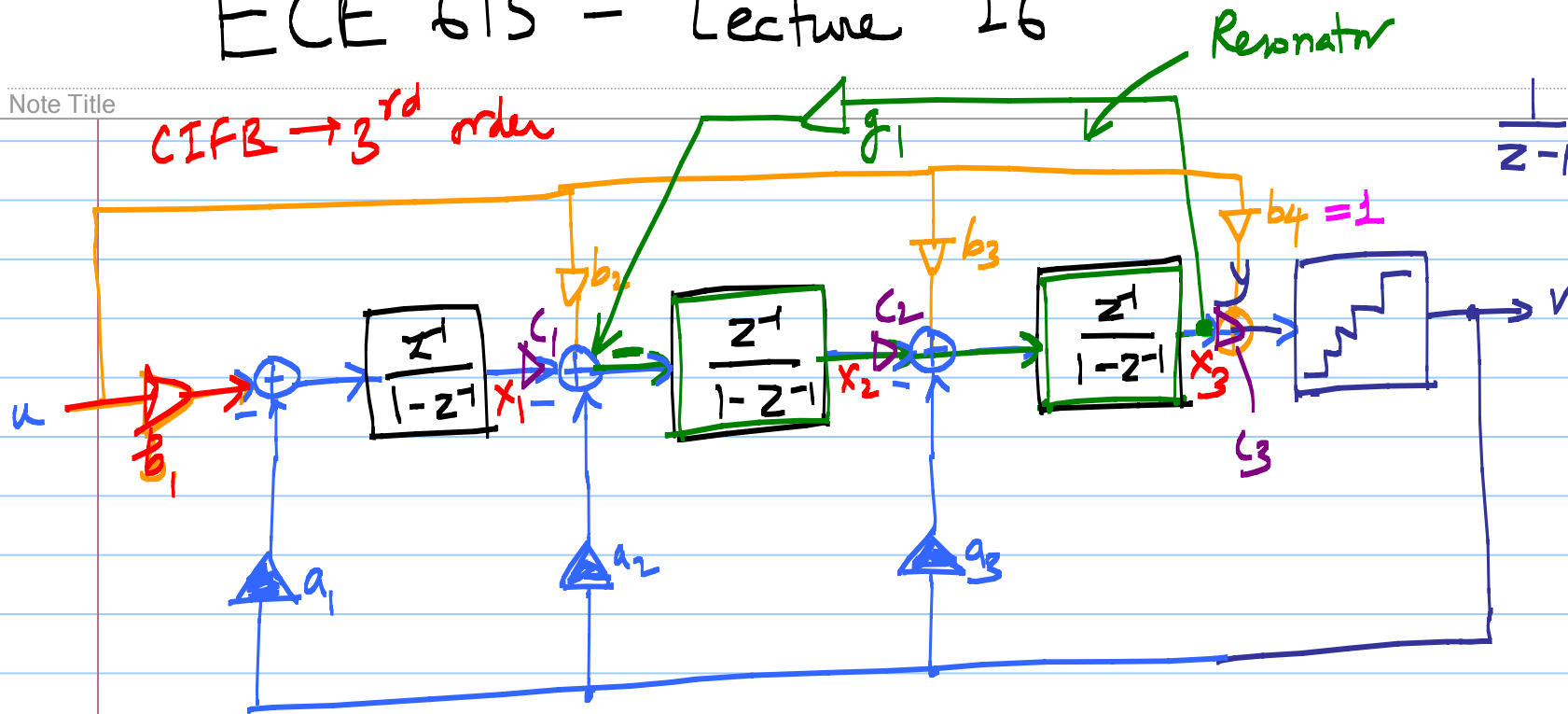


ECE 615 - Lecture 16

Note Title

10/29/2013

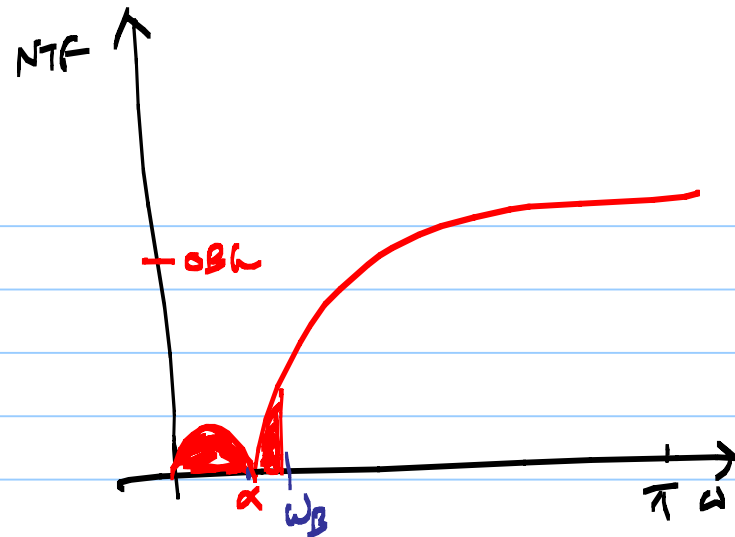
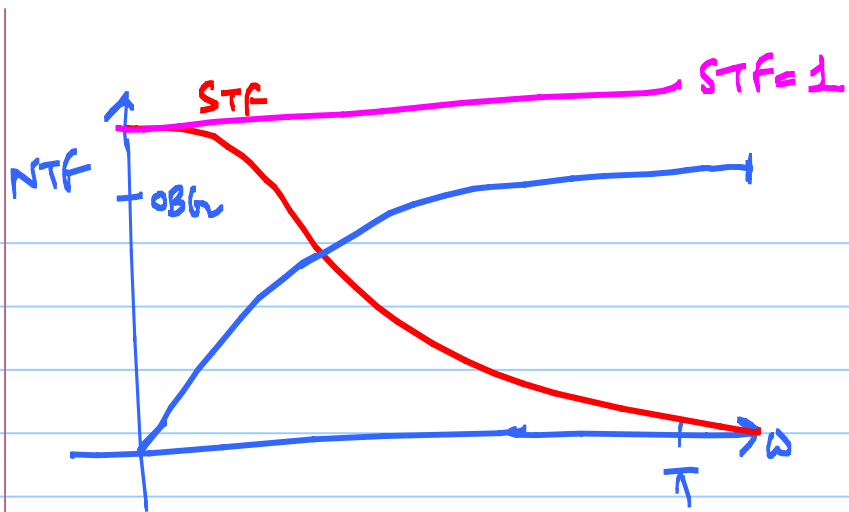


$$NTF(z) = \frac{(1-z^{-1})^3}{D(z)}$$

c_1, c_2, c_3

coefficient for
same scaling

\Rightarrow poles set the OBC



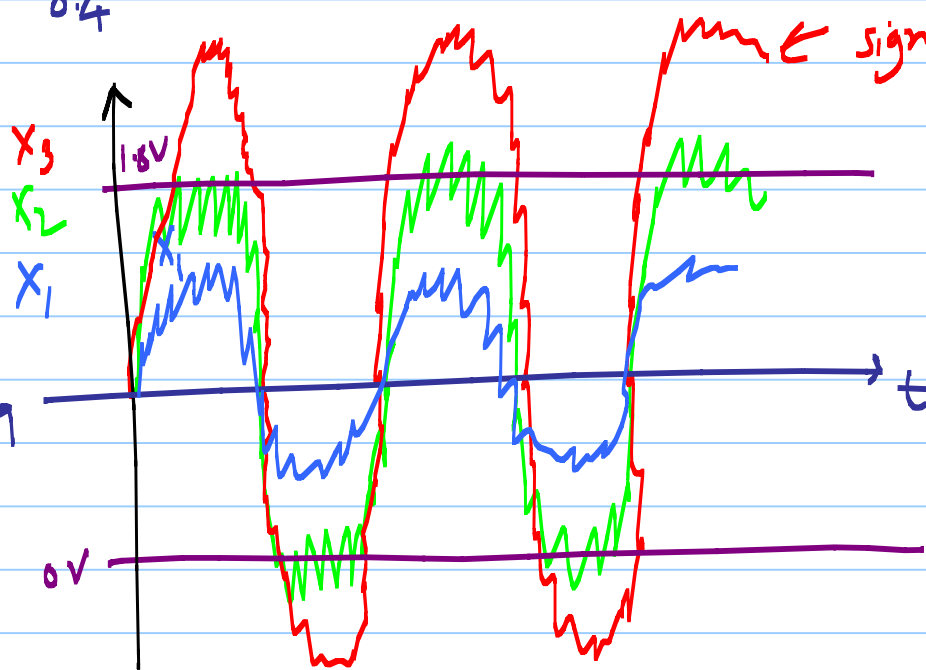
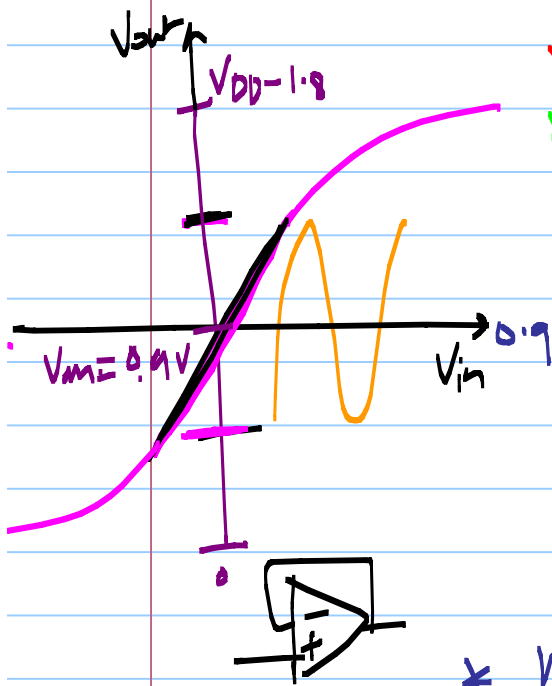
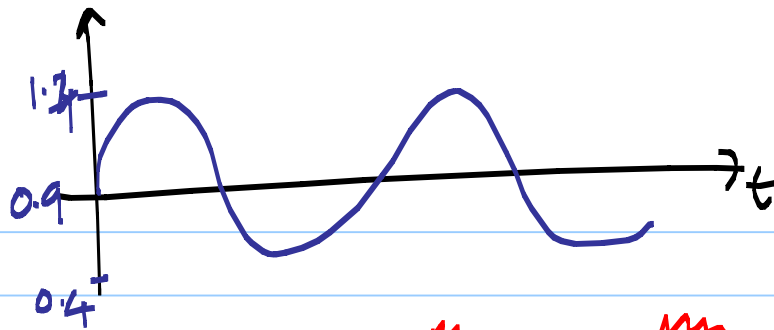
$$NTF(z) = \frac{(1 - z^{-1})(1 - e^{\pm j\alpha} z^{-1})}{D(z)}$$

$$f_s = 1V$$

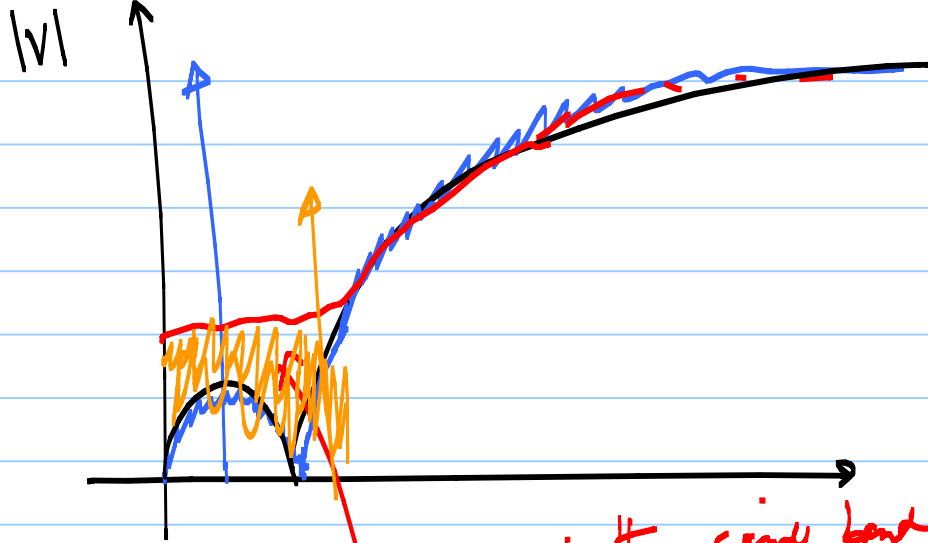
$$u = 0.5 f_s = 0.5V$$

$$V_{cm} = 0.9V$$

$$\frac{P_{in}}{f_s} = \frac{m}{N_{FFT}}$$

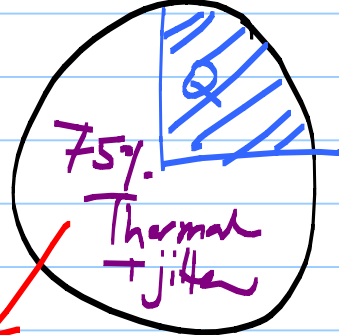


* We want to restrict the signals within the linear range of the integrator.



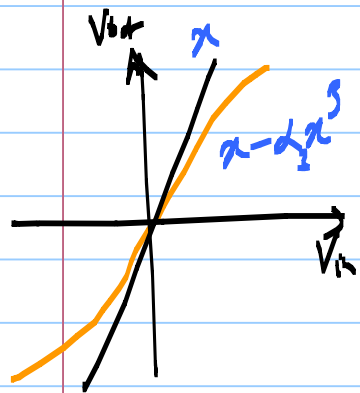
Sufficient dithering
 ↳ thermal noise
 of the circuit
 for dithering

Input-referred
 in-band noise



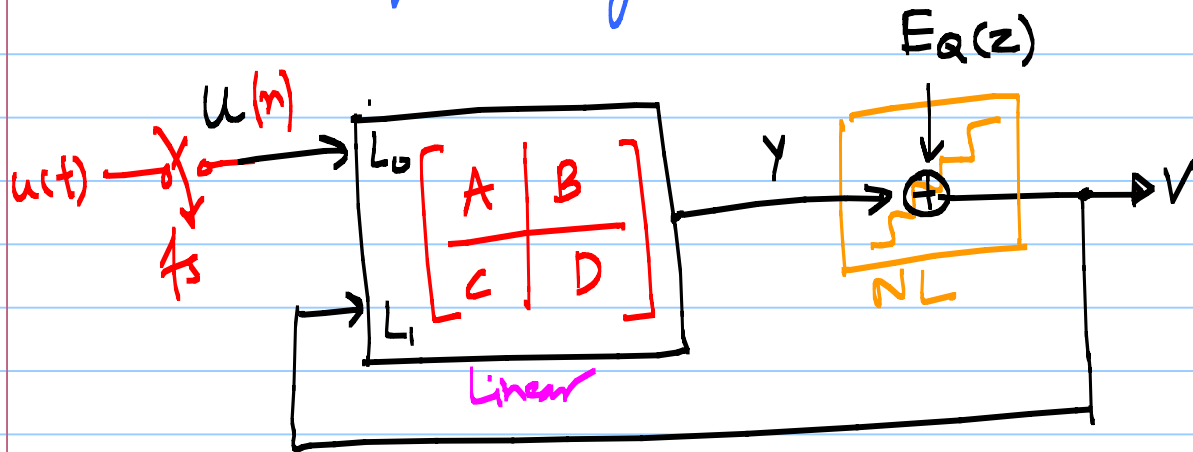
Notch in the signal band
 gets filled up with
 added Q-noise

$$\propto \sqrt{\frac{kT}{C}}$$



Dynamic Range Scaling (DRS)

"Linear System Theory"



* The loop filter can be described as a linear system

$$x[n+1] = Ax[n] + B \begin{bmatrix} u[n] \\ v[n] \end{bmatrix}$$

↑ next state ↑ previous

$x \rightarrow$ states
integrator outputs

$$x = \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

3x1 vector

$$y[n] = Cx[n] + D \begin{bmatrix} u[n] \\ v[n] \end{bmatrix}$$

for M^{th} -order modulator, $x[n] \in \mathbb{R}^{M \times 1} \rightarrow$ state vector

$A \in \mathbb{R}^{M \times M} \rightarrow$ interconnections
within the loop filter

$B \in \mathbb{R}^{M \times L} \Rightarrow$ describes how $u(n)$ & $v(n)$ are applied to the loop-filter

$C \in \mathbb{R}^{1 \times M}$
 $D \in \mathbb{R}^{1 \times L}$ } \Rightarrow compute $y(n)$ from the states and
direct path from inputs to $y(n)$

M=3

$$A = R^{3 \times 3}$$

$$\dot{x} = Ax$$
$$x(n+1) = \Delta t(n)$$

	\dot{x}_1	\dot{x}_2	\dot{x}_3
x_1		c_1	
x_2			c_2
x_3		$-g_1$	

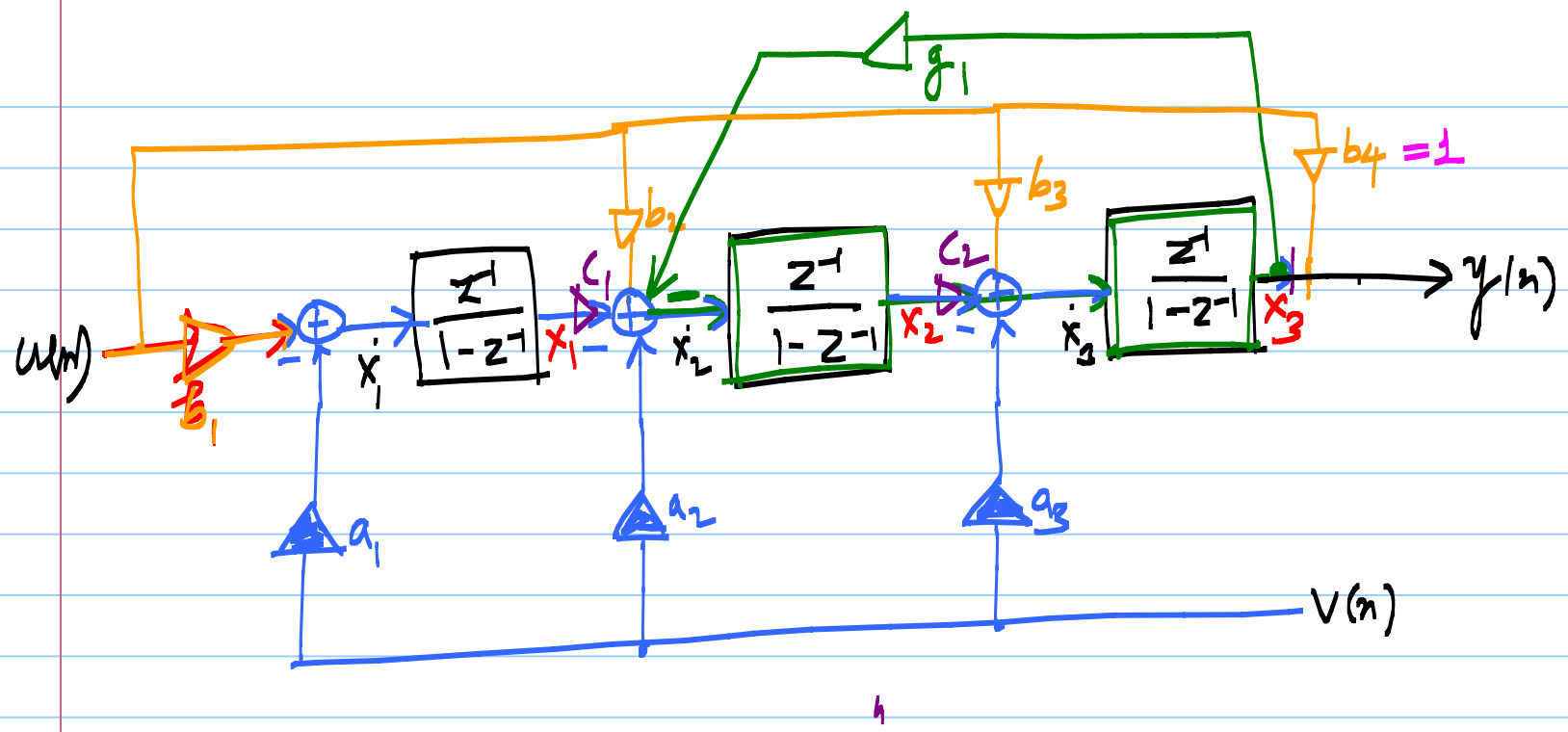
$$\dot{x}_2 = x_1 \cdot c_1$$

$$\dot{x}_3 = x_2 \cdot c_2$$

$$B = \left[\begin{array}{c|c} b_1 & -a_1 \\ b_2 & -a_2 \\ b_3 & -a_3 \end{array} \right]$$

$$C = [0 \quad 0 \quad c_3]$$

$$D = [b_4 \quad 0]$$



find $L(z)$ from $\begin{bmatrix} A & B \\ C & D \end{bmatrix}$

$$L_0(z) = \frac{Y(z)}{U(z)} \Big|_{V(z)=0}$$

$$L_1(z) = \frac{Y(z)}{V(z)} \Big|_{U(z)=0}$$

CT in Linear Systems course

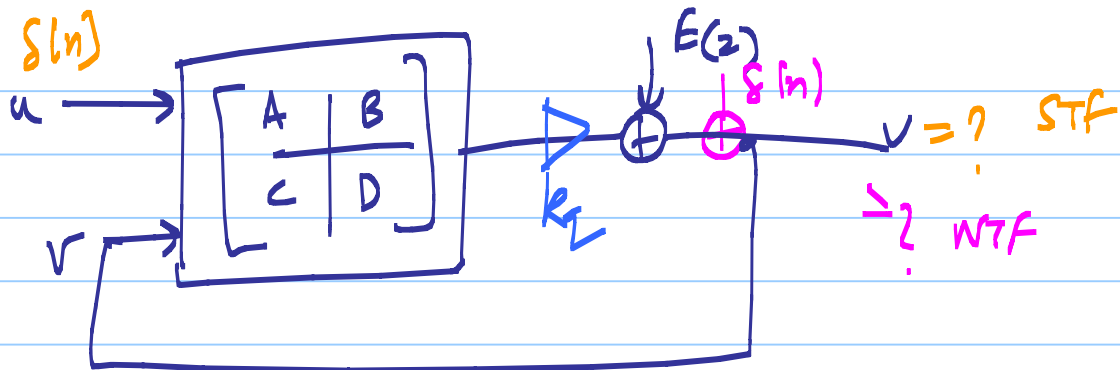
$$L(s) = C(sI - A)^{-1}B + D$$

DT

$$\begin{bmatrix} L_0(z) \\ L_1(z) \end{bmatrix} = C(z^{-1}I_{m \times m} - A)^{-1}B + D$$

→ $NTF(z) = \frac{1}{1 + L_1(z)}$ ✓

$STF(z) = \frac{L_0(z)}{1 + L_1(z)}$ ✓



for the closed-loop system

$$B = [B_1 \quad B_2]$$

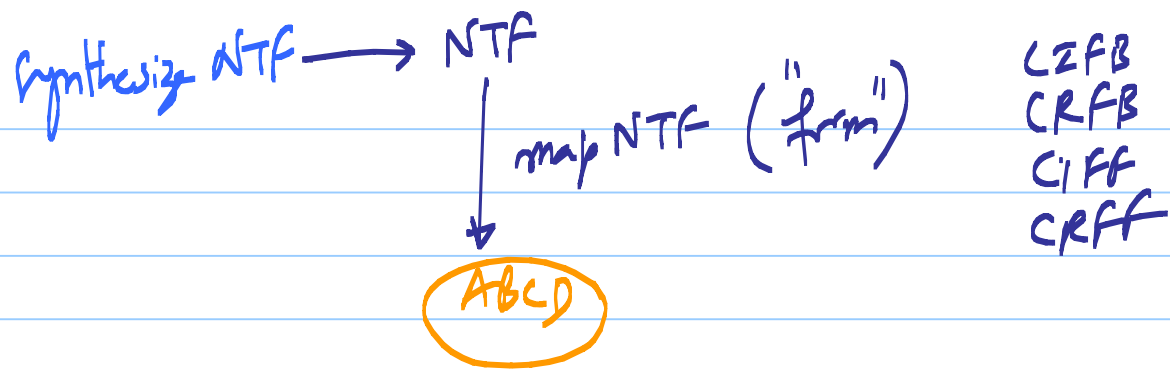
$$A_{cl} = A + k_L B_2 C$$

$$B_{cl} = [B_1 + k_L B_2 D_1 \quad B_2]$$

$$C_{cl} = k_L C$$

$$D_u = [k_2 D_1 \quad 1]$$

$$\begin{bmatrix} \text{STF}(z) \\ \text{NTF}(z) \end{bmatrix} = C_u (z^{-1}I - A_u)^{-1} B_u + D_u$$



$$X_{lim} = 0.5V$$

$$d_{max} = \begin{bmatrix} 1 \\ 2 \\ 3 \end{bmatrix}$$

\uparrow found by simulation

$$X_{\text{new, max}} = S \cdot X_{\text{max}} \rightarrow \begin{bmatrix} 1 \\ 2 \\ 3 \end{bmatrix}$$

$$S = \begin{bmatrix} 1/2 & 0 & 0 \\ 0 & 1/2 & 0 \\ 0 & 0 & \dots & 1/x_m \end{bmatrix} = \begin{bmatrix} 1/2 & 0 & 0 \\ 0 & 1/4 & 0 \\ 0 & 0 & 1/6 \end{bmatrix}$$

"Scaling matrix"

Linear transformation of the state-space

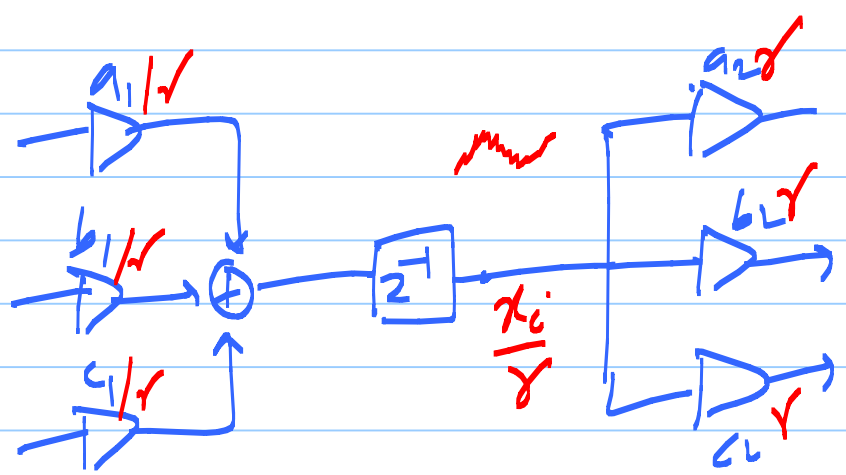
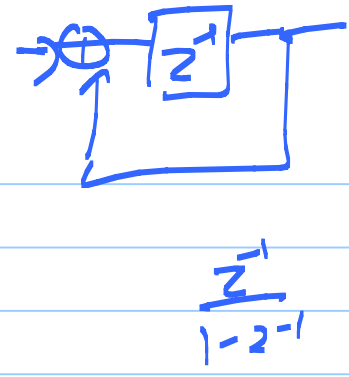
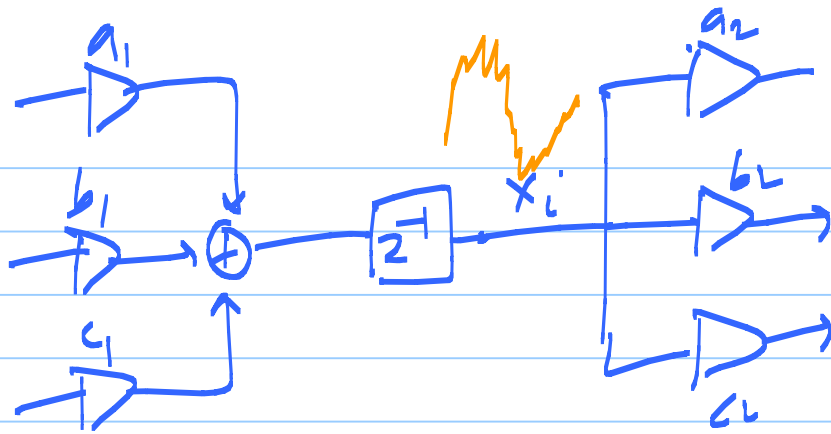
$$x_s = Sx$$

for $x_s = Sx \Leftarrow$ Dynamic Range Scaling

$$ABCD_s = \left[\begin{array}{c|c} SAS^{-1} & SB \\ \hline CS^{-1} & D \end{array} \right] = \left[\begin{array}{c|c} A_1 & B_1 \\ \hline C_1 & D_1 \end{array} \right]$$

scale ABCD matrix

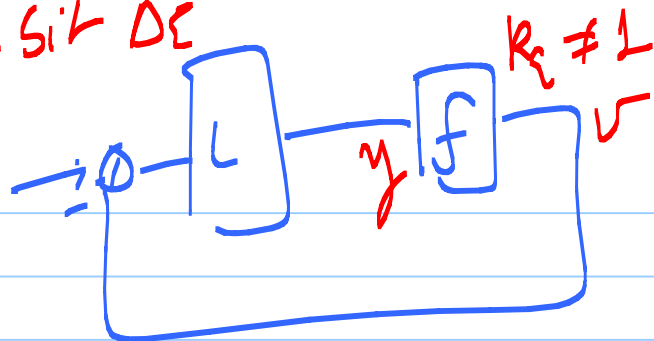
scale ABCD



DRS

$c \neq 1$

for single-site DE



$$k_2 = \frac{E[y]}{E[y^2]}$$