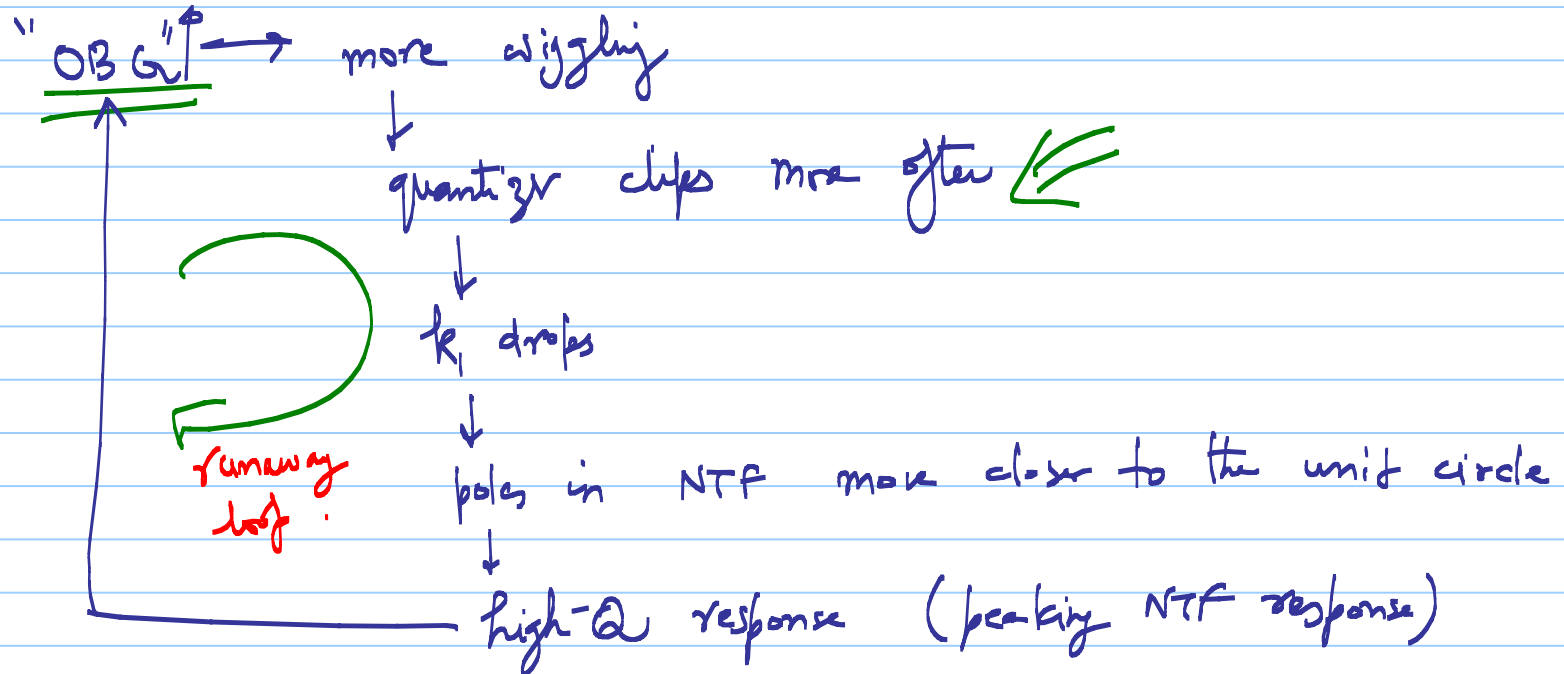


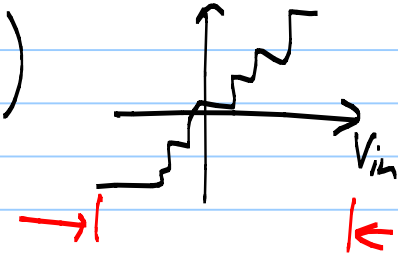
ECE 615- Lecture 10



Multibit $\Delta\Sigma$ Modulator (n -bit quantizer)

* If we increase the # of quantizer levels (2^n)

\Rightarrow LSB size decreases ($\Delta = \frac{FS}{2^n}$)



\Rightarrow Variance of quantization noise at the quantizer input $NTF = (-z^{-1})^N$

$$\sigma_y^2 = \frac{\Delta^2}{12} \left(\sum_n h[n]^2 - 1 \right)$$

$$= \begin{cases} 5 \frac{\Delta^2}{12} & N=2 \\ 19 \frac{\Delta^2}{12} & N=3 \\ 69 \frac{\Delta^2}{12} & N=4 \end{cases}$$

$h[n] \leftrightarrow NTF(z)$

$$\Delta \downarrow 2 \Rightarrow \sigma_y^2 \downarrow 4$$

order of the modulator

* Maximum Stable Amplitude (MSA) \rightarrow u_{max}

MSA $<$ FS of the quantizer

$(-z^{-1})^2$

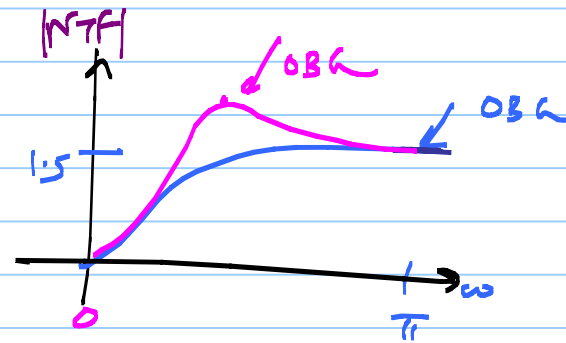
\Rightarrow NTF order, $N \uparrow \Rightarrow$ MSA \downarrow

$m \uparrow \Rightarrow \Delta \downarrow \Rightarrow$ MSA \uparrow

A good value for max stable amplitude \rightarrow MSA ≈ 0.8

* for a single-bit $\Delta\Sigma$ modulator.

$$OBC \leq 1.5 \quad \leftarrow \text{Lee's Criterion}$$



$$OBC = \max_{\omega} |NTF(e^{j\omega})|$$

$$= \text{infinity norm} \Rightarrow \|NTF\|_{\infty}$$

$$\begin{aligned} \text{NTF}(z) = (1-\bar{z}^{-1}) &\Rightarrow \text{OBQ} = 2 \\ (1-\bar{z}^{-1})^2 &\Rightarrow \text{OBQ} = 4 \end{aligned}$$

Single-bit $\Delta\Sigma \Rightarrow$ easy circuit-level design (low-power)

\hookrightarrow but can't use aggressive NTF

$$\text{OBQ} \leq 1.5$$

Derivation skipped

$$\max u(n) \leq M+2 - \|h\|_1$$

of levels

2-norm of the NTF

$$\sum_n |h[n]|$$

Ex $M=16$ levels, $\text{NTF}(z) = (1-\bar{z}^{-1})^3 \Rightarrow \|h\|_1 = 8$

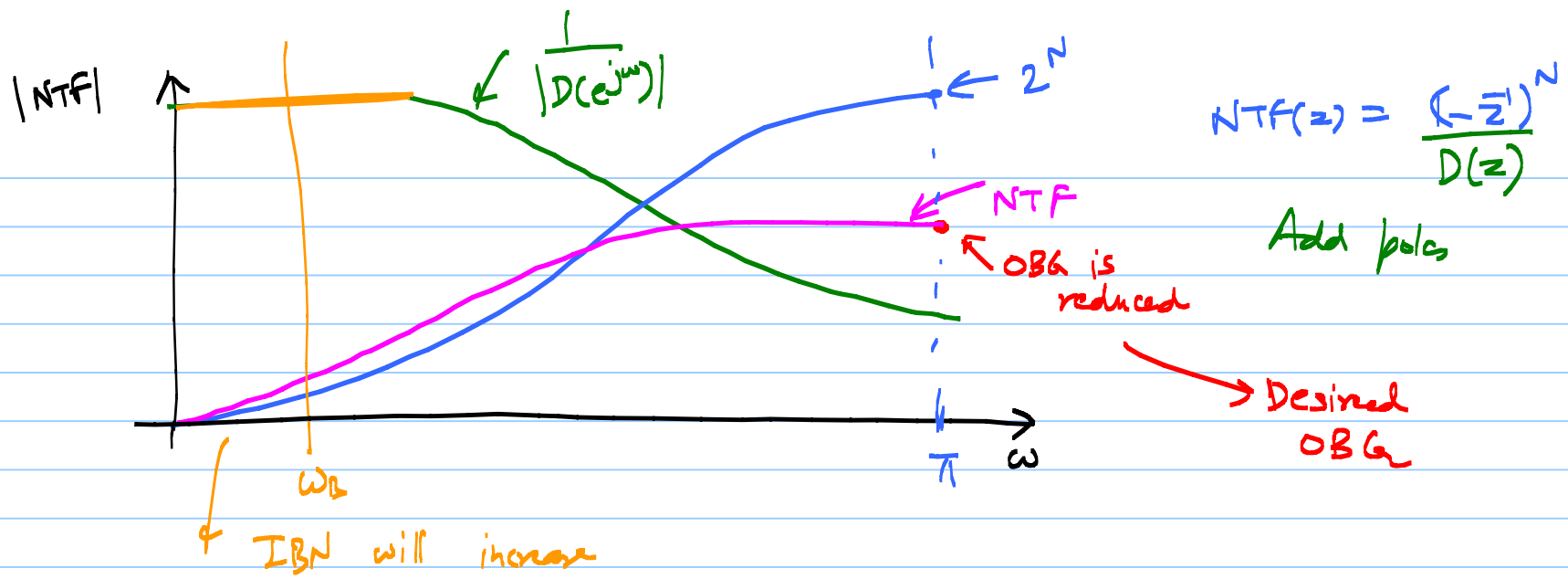
$$\Rightarrow \frac{\max u(n)}{FS} = \underline{\underline{62.5\%}}$$

$$\boxed{\text{OBQ} = 8}$$

Systematic NTF Design procedure

* for higher-order NTF, ~~$(1-z^{-1})^N$~~ MSA reduces drastically as $OBG = 2^N$

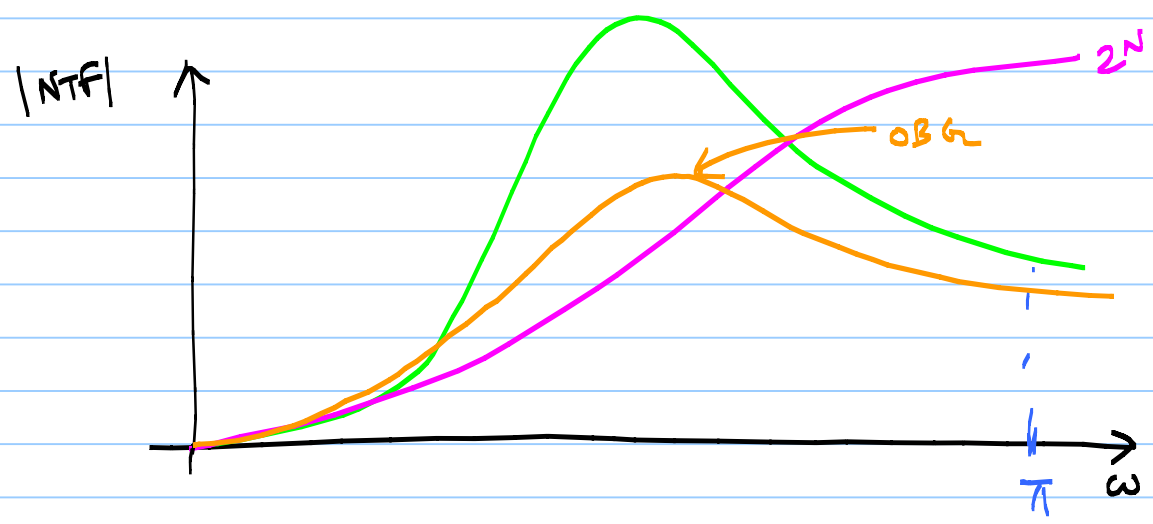
* We want the MSA to be 75% - 80% of the FS

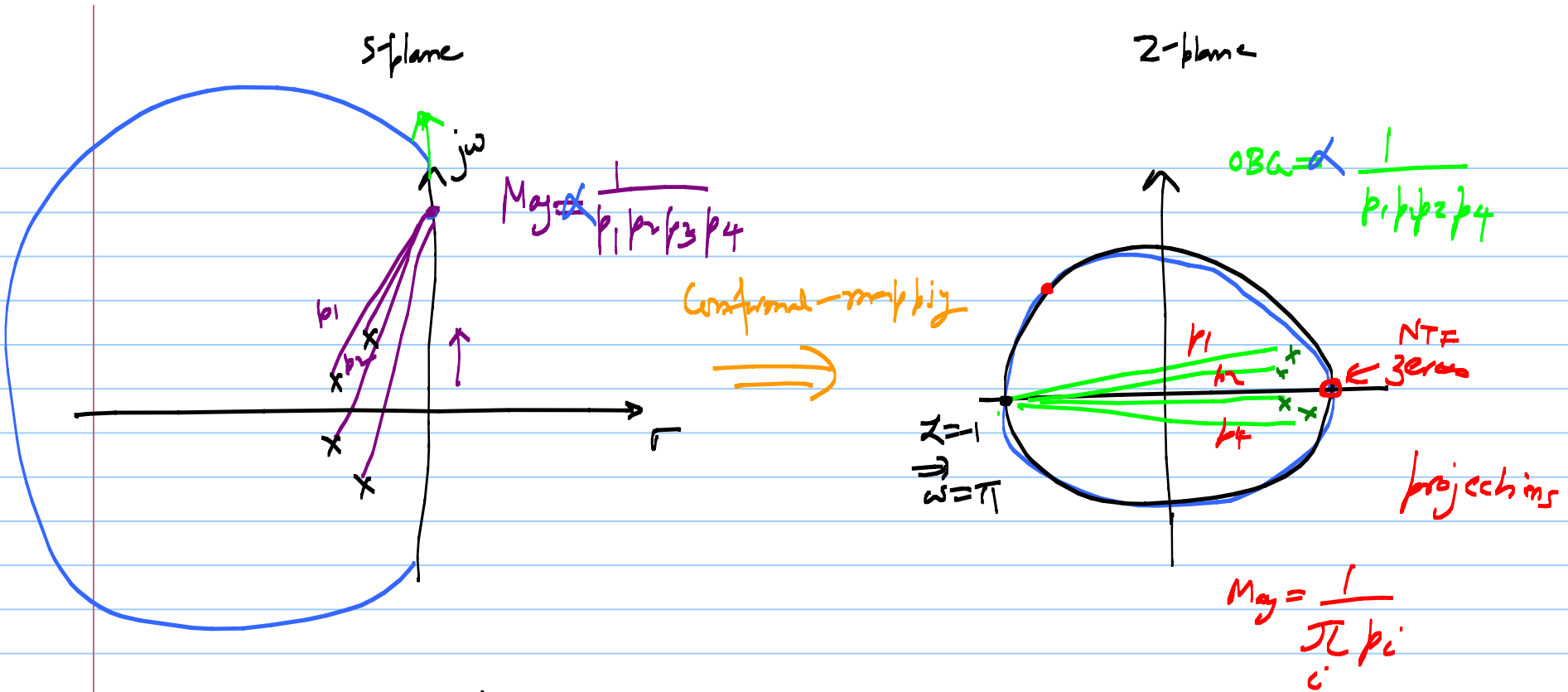


for a given order \Rightarrow

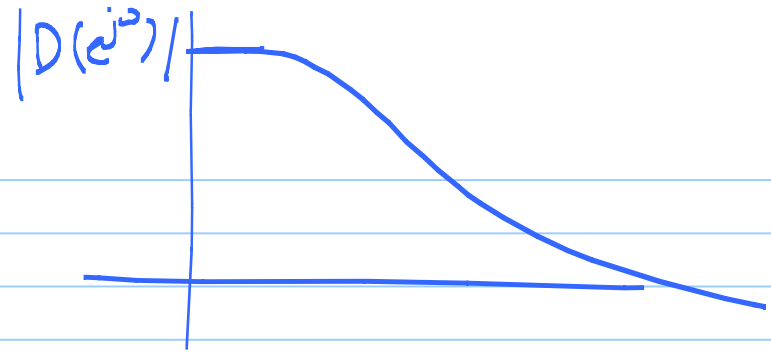
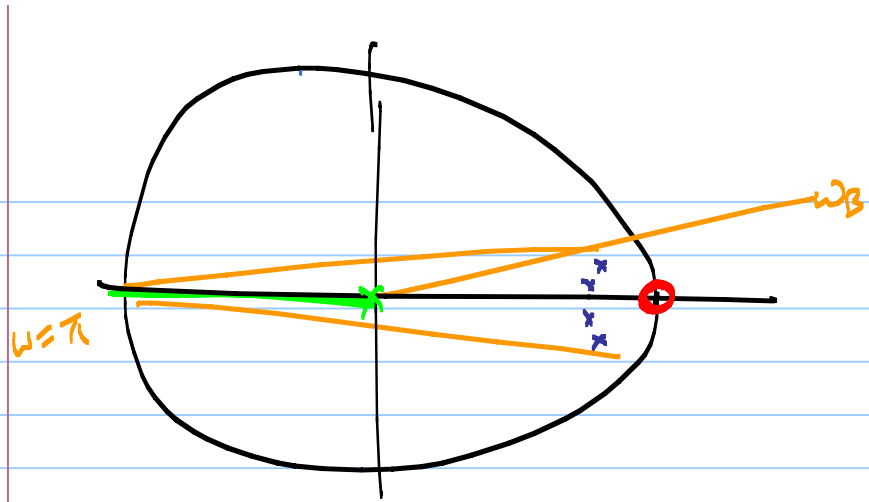
$OBG \leftrightarrow SQNR$
 $Stability \leftrightarrow performance$

$$\frac{1}{D(z)} \leftarrow \text{LPF}$$





* gain at $\omega = \pi$ is lower as the poles are farther away from $z = -1$



* Recall the realizability condition (delay free loops)

$h[n]$

$$h[0] = 1 \Rightarrow \text{NTF}(\infty) = 1$$

$$\text{If } \text{NTF}(z) = \frac{(1 - z^{-1})^4}{a_0 + a_1 z^{-1} + \dots + a_q z^{-q}} = \frac{N(z)}{D(z)}$$

\swarrow HPF
 \swarrow LPF

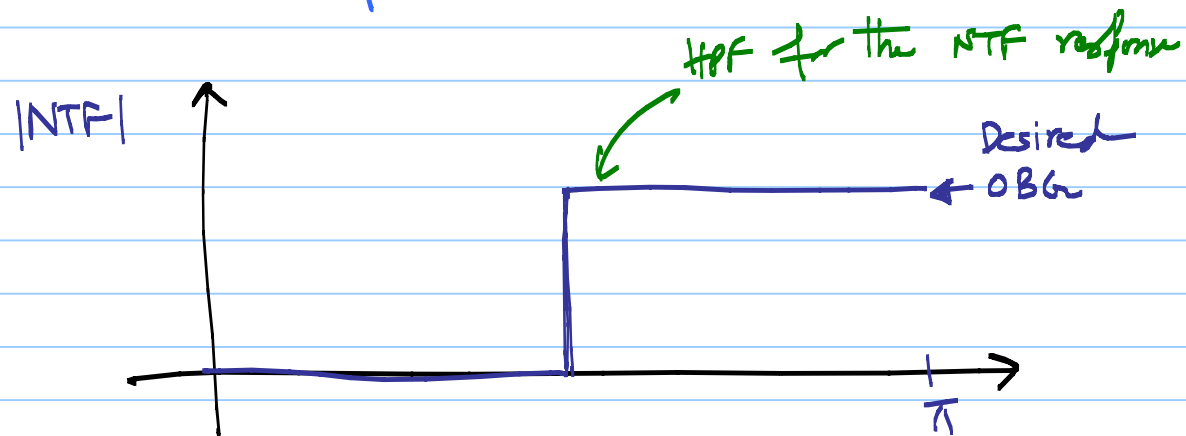
$$\text{NTF}(z = \infty) = 1$$

$$\text{NTF}(z^{-1} = 0) = 1 \Rightarrow$$

$$a_0 = 1$$

* Which pole configuration to use?

Approximation problem



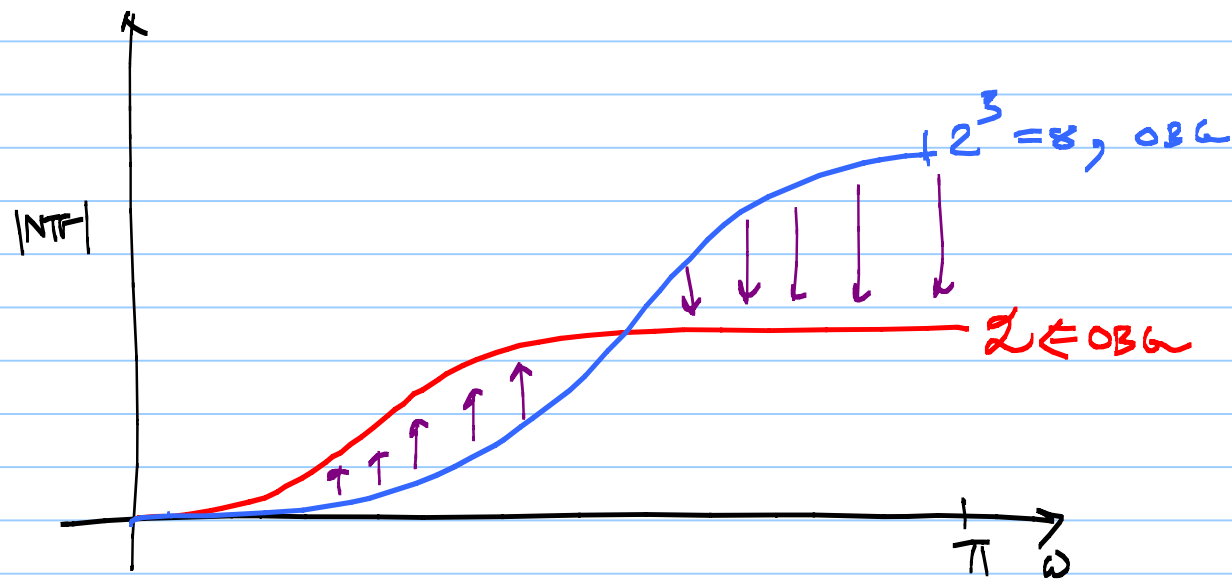
* Approximate the ideal brickwall HPF response with a polynomial NTF(z).

① Butterworth

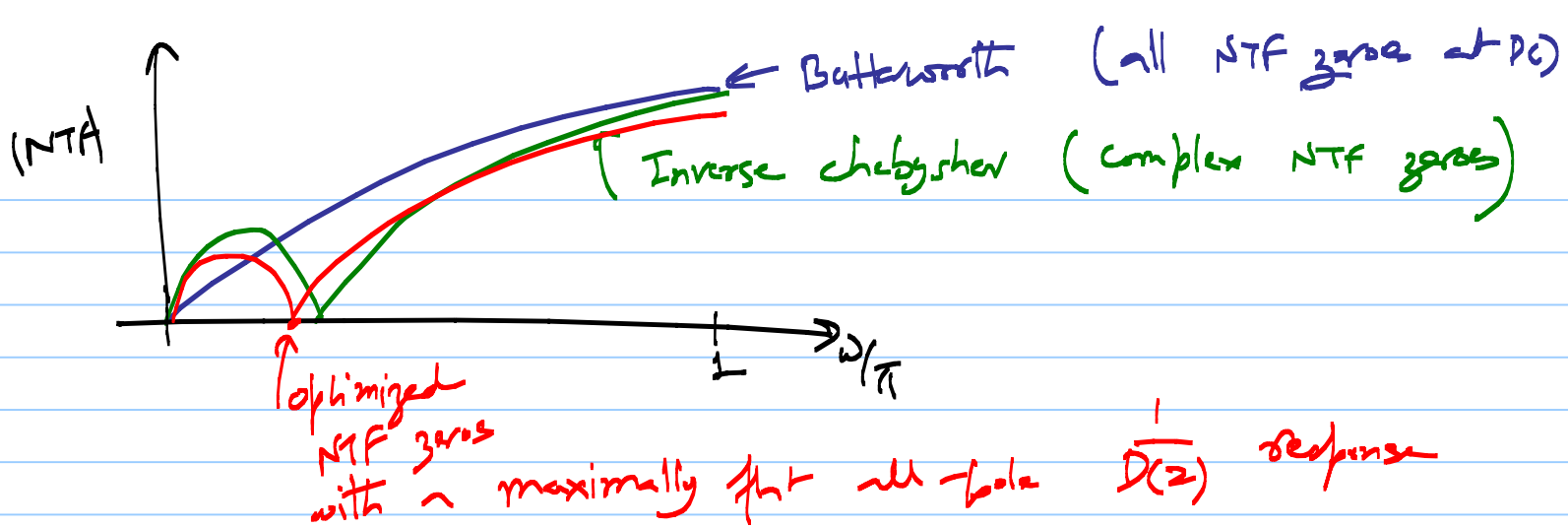
② Inverse Chebyshev

③ Maximally flat all-pole transfer function, etc.

* properly choose poles to reduce OBG
 \Rightarrow MSA \uparrow \Rightarrow stability is improved
 \hookrightarrow higher IBN



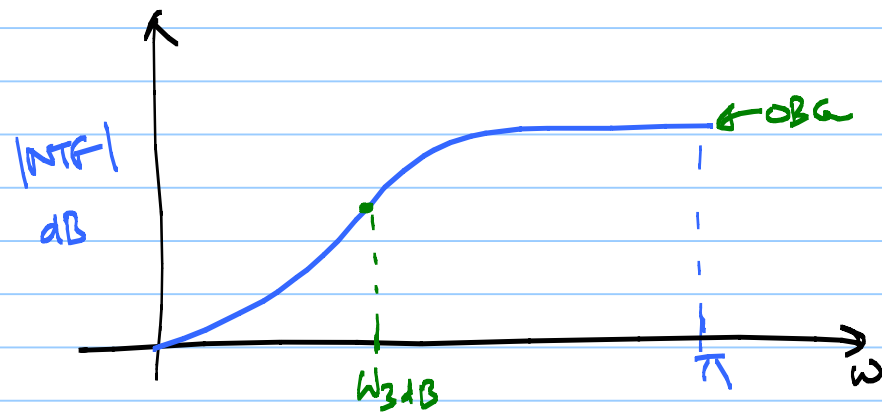
$$\frac{(1-z^{-1})^3}{D(z)}$$



* MATLAB Functions

{ butter
 cheby2
 maxflat

* Let's choose a Butterworth NTF response



$$\frac{(-z^{-1})^3}{a_0 + a_1 z^{-1} + \dots + a_3 z^{-3}}$$

3rd-order Butterworth response

We need $a_0 = 1$