

Zero Discussion

Note Title

4/9/2011

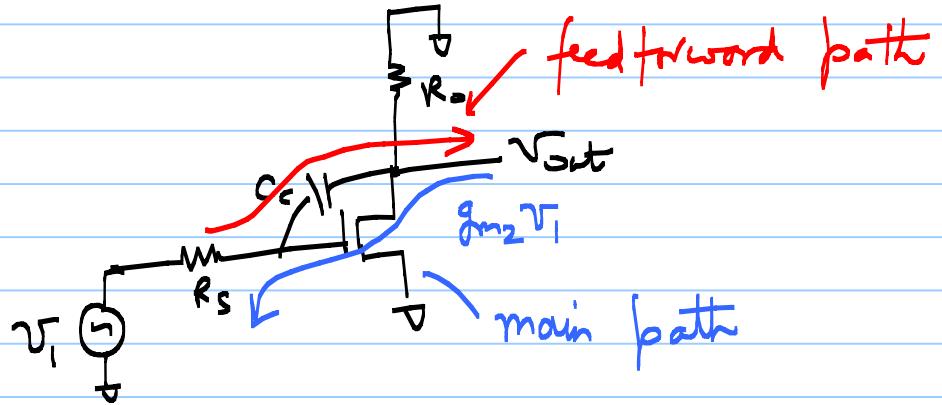
$$\omega_3 = + \frac{g_{m2}}{C_c}$$

↳ R HP zero

↳ arises due to the direct coupling of the input to the output through C_c .

↳ C_c provides a feedforward path that conducts the input signal to the output at very high frequencies

⇒ +20 dB/dec roll-up



→ RHP zero reduces phase at high frequencies as the signal through C_s adds to the main path output in the opposite phase.

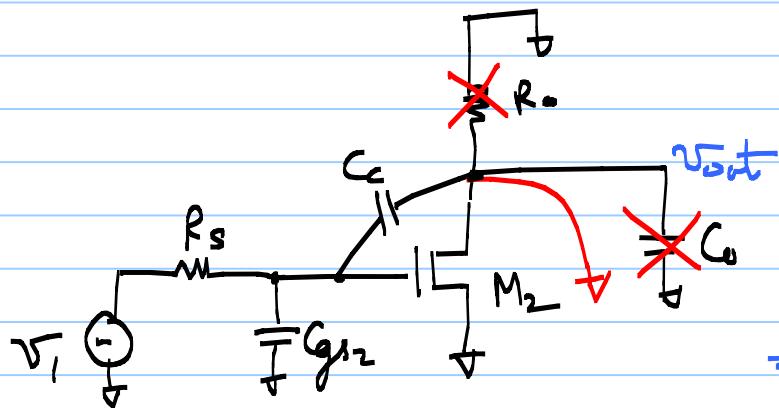
* Quick Calculation of zero :

$R_g + j\omega_g$
corresponds to $j\omega_g$

$$\text{at } s = s_g \Rightarrow \frac{V_{out}(s)}{V_{in}(s)} = 0$$

for a finite V_{in}

$$V_{out}(s_g) = 0$$



\Rightarrow the output can be shorted to ground
at this value of $s = s_g$, with no
current flowing through the short

\Rightarrow current through C_L & M_2 are equal and opposite

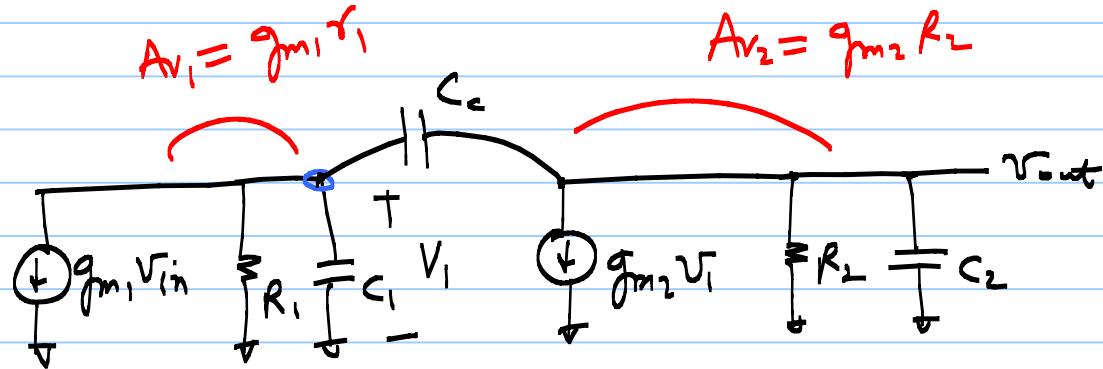
$$\Rightarrow V_1 C_C S_3 = g_{m2} V_1$$

$$\Rightarrow \boxed{S_3 = +\frac{g_{m1}}{C_g d_1}}$$

RHP zero.

\Rightarrow if the polarity of the current through C_C is inverted
then it could lead to an LHP zero.

* Quick Calculation of f_m (or ω_m) from the circuit.

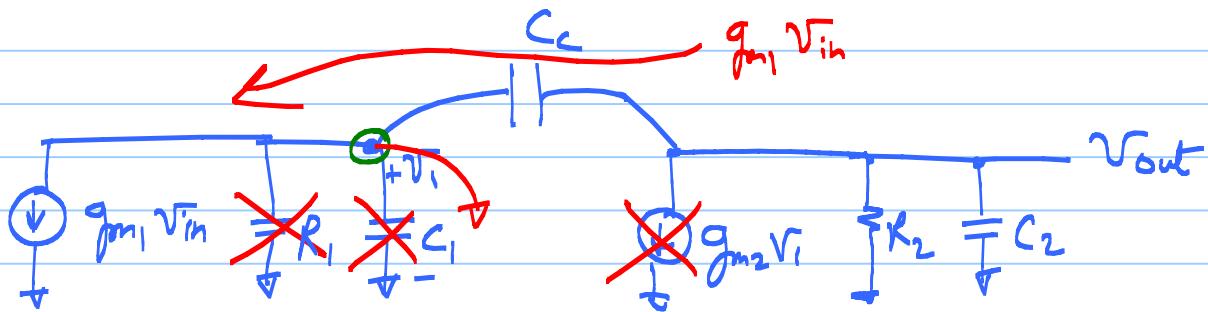


* When the amplifier is operated in feedback, the output of the first stage is given by $v_1 = \frac{v_{out}}{A_{v2}}$

* if A_{v2} is large then $v_1 \approx 0$

\Rightarrow all the current $g_{m1}v_{in}$ flows to the output through C_c .

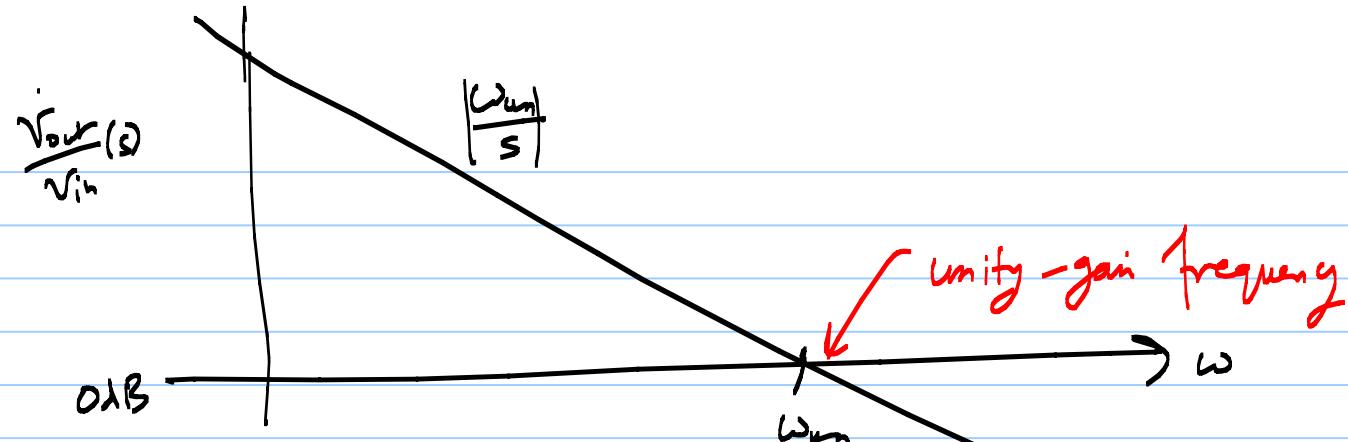
Equivalent circuit \Rightarrow



\Rightarrow
KCL @ 0 \Rightarrow

$$g_{m1}v_{in} = sC_c v_{out}$$

$$\Rightarrow \frac{v_{out}}{v_{in}} \approx \frac{g_{m1}}{sC_c} = \frac{\omega_m}{s} \quad \text{where } \omega_m = \frac{g_{m1}}{C_c}$$



$$\Rightarrow f_m = \frac{\omega_m}{2\pi} = \frac{g_m}{2\pi C_c}$$

- * Alternatively we can also derive form as follows :
- * Dominant pole response

$$H(s) \approx \frac{A_v}{1 + s/\omega_{p_1}}$$

$$\approx \frac{A_v \cdot \omega_{p_1}}{s} \quad \text{for} \quad \left| \frac{s}{\omega_{p_1}} \right| \gg 1$$

$$= \frac{\omega_m}{s}$$

$$\Rightarrow \omega_m = A_v \cdot \omega_{p_1}$$

$$= g_{m1} f_1 g_{h2} f_2 \cdot \frac{1}{\beta_2 R_L \beta_1 C_L} = \boxed{\frac{g_{m1}}{C_L}}$$