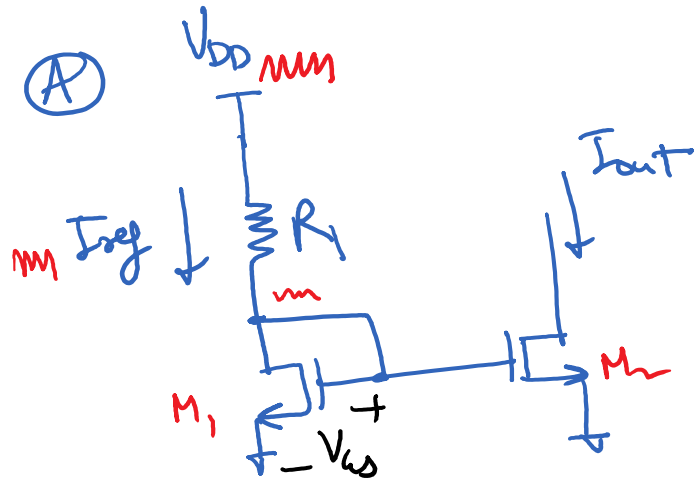
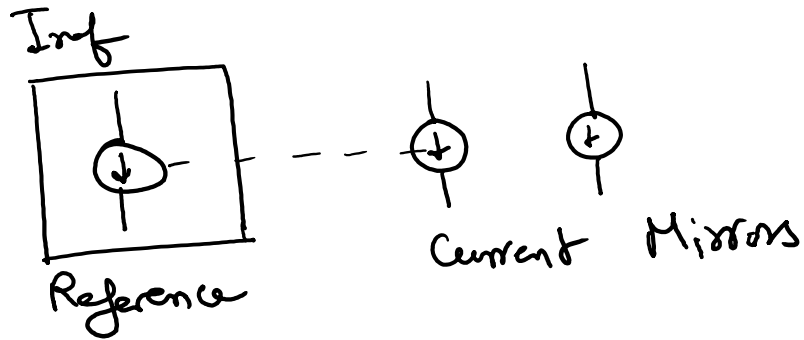


ECE 511 - Lecture 6.

Tuesday, September 11, 2018

11:00 AM



$$I_{ref} = \frac{V_{DD} - V_{gs}}{R_1} = \frac{\beta_n}{2} (V_{gs} - V_{thn})^2$$

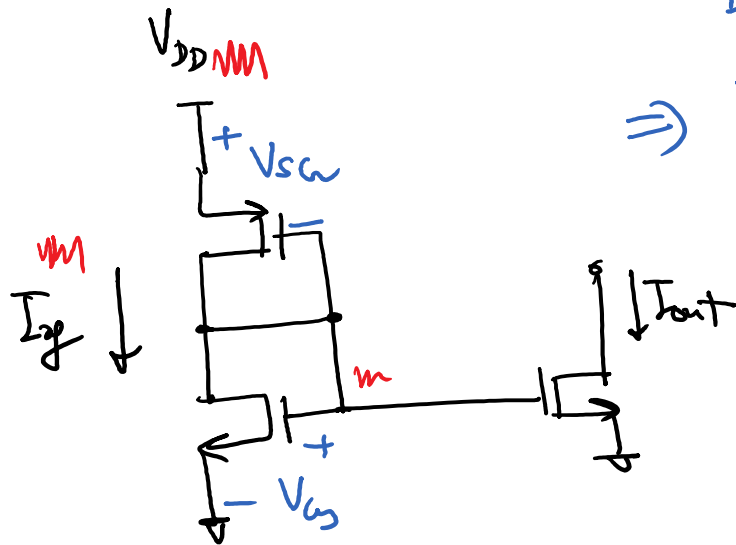
$$I_{ref} = f(V_{DD})$$

$$\Delta I_{out} = \frac{\Delta V_{DD}}{R_1 + \frac{1}{g_{m1}}} \cdot \frac{(W/L)_2}{(W/L)_1}$$

$$\beta_n = k_n \frac{W}{L}$$

$$V_{GS} = \sqrt{\frac{2I_D}{\beta}} + V_{THN}$$

③



$$V_{DD} = V_{scw} + V_{cs}$$

$$\Rightarrow V_{DD} = \sqrt{\frac{2I_{ref}}{\beta_p}} + V_{THP} + \sqrt{\frac{2I_{ref}}{\beta_n}} + V_{THN}$$

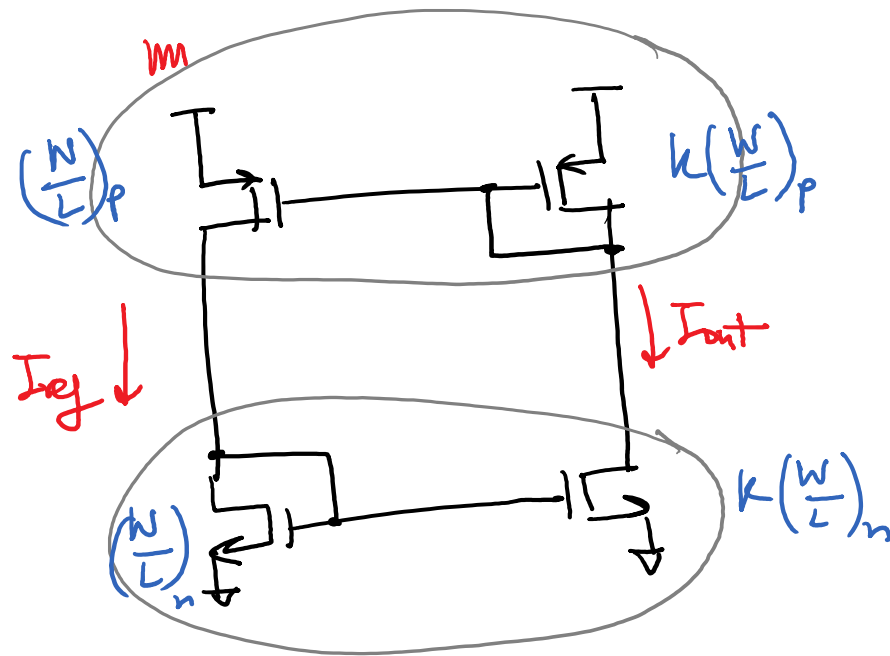
$$(V_{DD} - V_{THN} - |V_{THP}|)^2 = I_{ref} \left(\sqrt{\frac{2}{\beta_p}} + \sqrt{\frac{2}{\beta_n}} \right)^2$$

$$I_{ref} = \frac{(V_{DD} - V_{THN} - |V_{THP}|)^2}{\left(\sqrt{\frac{2}{\beta_p}} + \sqrt{\frac{2}{\beta_n}} \right)^2}$$

$V_{DD} < V_{THN} + |V_{THP}|$
 $I_{ref} \Rightarrow$
 Trivial
 Case

$$A_{I_{ref}} = \frac{\Delta V_{DD}}{\frac{1}{g_{mn}} + \frac{1}{g_{mp}}}$$

In order to get rid of V_{DD} sensitivity, the circuit must bias itself.
 $\rightarrow I_{ref}$ must be derived from I_{out}

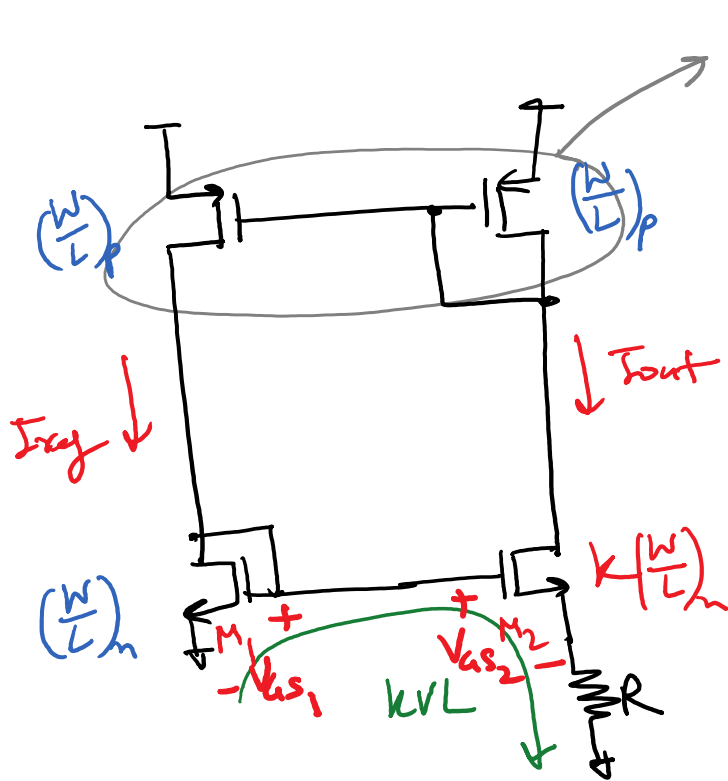


top mirror copies I_{out} to I_{ref}
 bottom mirror copies I_{ref} + I_{out} !

$$I_{out} = k I_{ref} \quad \left(\text{assuming } \lambda = 0 \right)$$

I_{ref} & I_{out} are independent of V_{DD} .

How do we set the values of I_{ref} & I_{out} ?



$$I_{out} = I_{ref}$$

$$V_{gs1} = V_{gs2} + I_{out} R$$

$$\sqrt{\frac{2I_{out}}{k\mu_n(W/L)_1}} + V_{thn1} = \sqrt{\frac{2I_{out}}{k\mu_n(W/L)_1}} + V_{thn2} + I_{out} R$$

ignoring body-effect: $V_{thn1} = V_{thn2}$

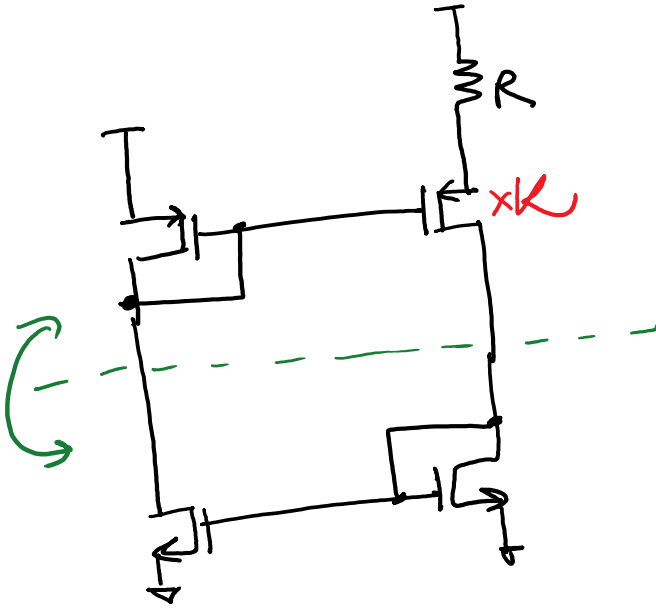
$$\sqrt{\frac{2I_{out}}{k\mu_n(W/L)}} \left(1 - \frac{1}{\sqrt{k}}\right) = I_{out} \cdot R$$

$$I_{out} = 0, \quad \frac{2}{k\mu_n(W/L)R^2} \cdot \left(1 - \frac{1}{\sqrt{k}}\right)^2 = I_{ref}$$

Independent of supply voltage

Beta Multiplier Reference (BMR)

→ independent of V_{DD} ($\lambda=0$)
not independent of Temp.



Transconductance of M_1 : $\sqrt{2\beta I_{ref}}$

$$g_{m1} = \sqrt{2\beta I_{ref}}$$

$$= \sqrt{2 \cancel{k\mu_n \left(\frac{W}{L}\right)_1} \cdot \frac{2}{\cancel{k\mu_n \left(\frac{W}{L}\right)_1} \cdot \frac{1}{R_2} \left(1 - \frac{1}{\sqrt{k}}\right)^2}$$

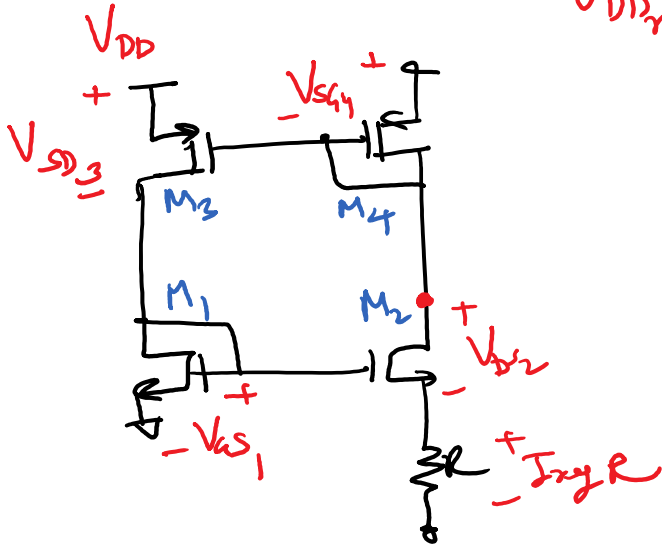
$$\Rightarrow g_m = \frac{2}{R} \left(1 - \frac{1}{\sqrt{k}}\right) \leftarrow \text{indep of } V_{DD} \text{ \& } V_{TH} \text{ etc}$$

\uparrow varies with temp

\Rightarrow Constant- g_m biasing circuit

for $k=4$, $g_m = \frac{1}{R}$

$$V_{DD_{min}} = \max \{ V_{GS1} + V_{SD3, sat}, V_{SG4} + V_{DS, sat2} + I_{avg} \cdot R \}$$



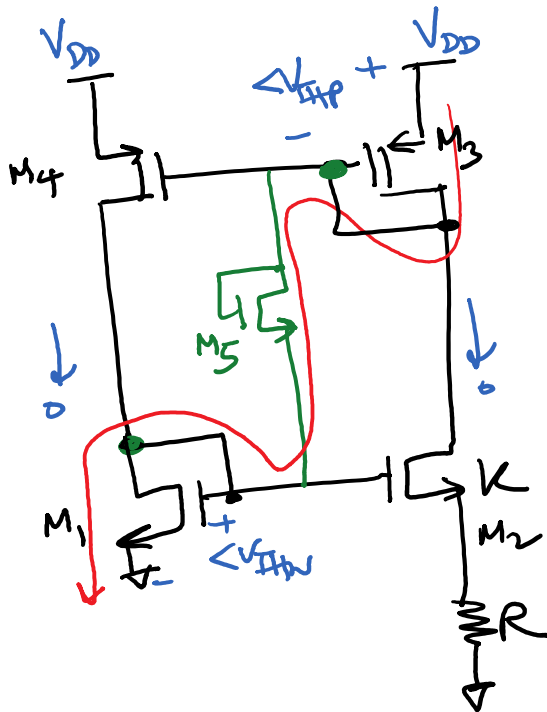
Start-up problem
of all transistors
↳ they

carry zero current when V_{DD} is turned on
may remain off indefinitely.

∴ $I_{out} = 0$ is also an admissible solution.

↳ degenerate bias condition ($I_{ref} = I_{out} = 0$)

↳ need to force the devices to
turn-on.



M₅ is providing current path from M₃ to M₁ so that they snap to the desired solution for I_{avg}.

* But M₅ should be off during normal operation

for M₅ to turn-off

$$V_{THN1} + V_{THN5} + V_{THN3} < V_{DD} \text{ to turn on } M5 \rightarrow \textcircled{1}$$

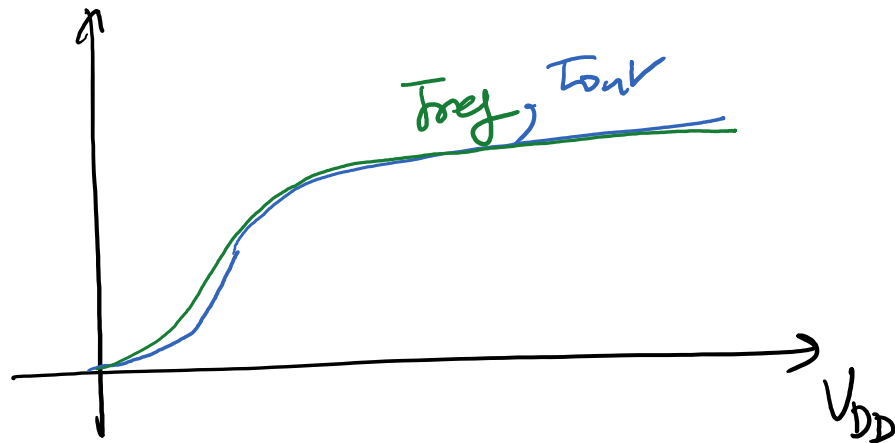
$$V_{G1} + V_{THN5} + V_{SG3} > V_{DD} \text{ to turn off } M5 \rightarrow \textcircled{2}$$

$$V_{DD} - (V_{G1} + V_{SG3}) < V_{THN5}$$

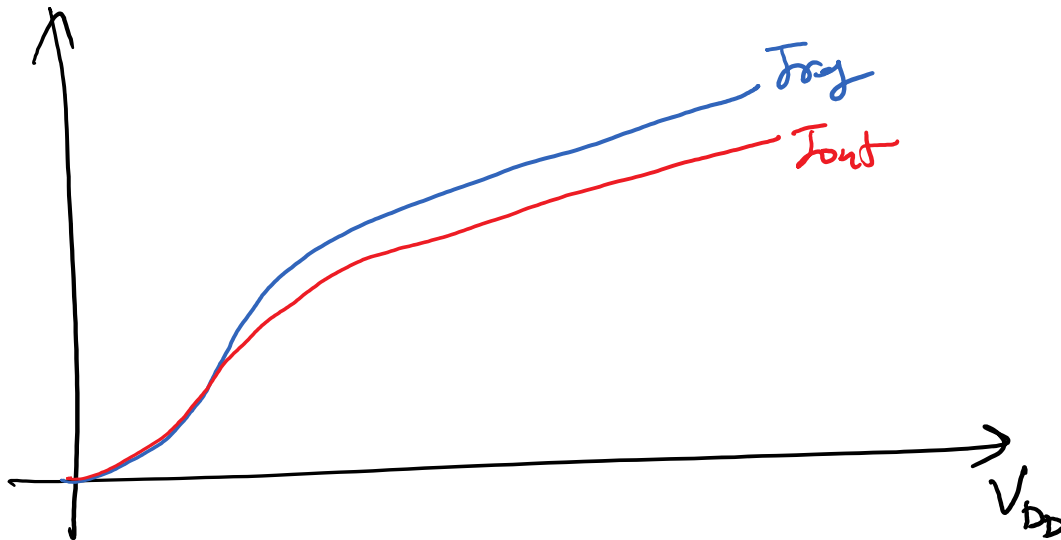
Not very robust idea.

normal state

In long-L process:



Short-channel process:



Not very accurate
and supply independent
due to CCM.