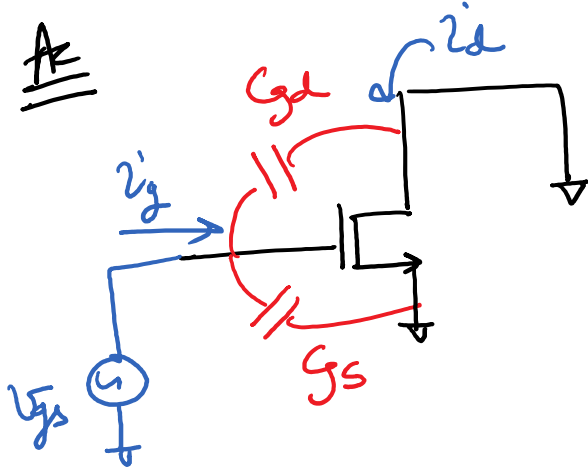


ECE 515- Lecture 4

Tuesday, September 4, 2018 11:00 AM

Transition frequency (f_T)



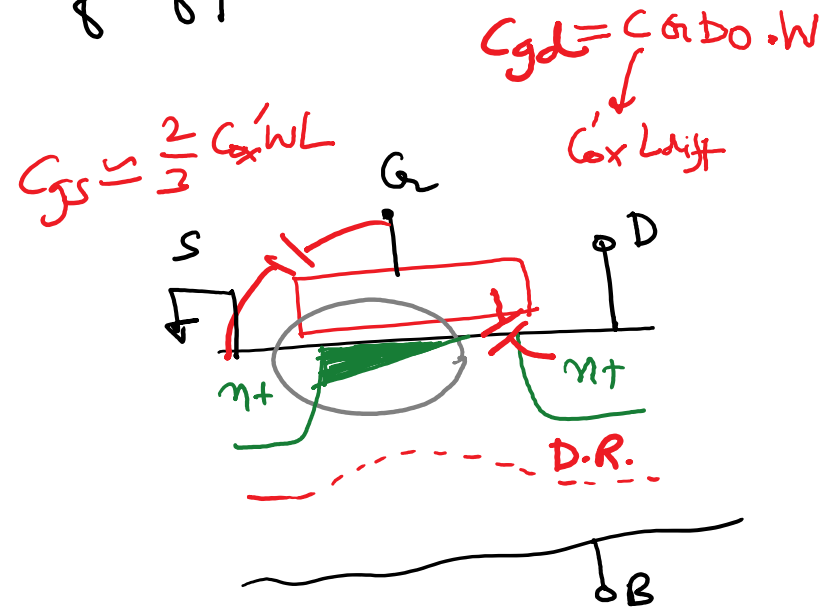
short circuit current gain

$$\left| \frac{i_d}{i_g} \right| = 1 \quad \text{at } f = f_T$$

$$\begin{aligned} i_d &= g_m v_{gs} \\ &= g_m i_g \cdot \frac{1}{s(C_{gd} + C_{gs})} \end{aligned}$$

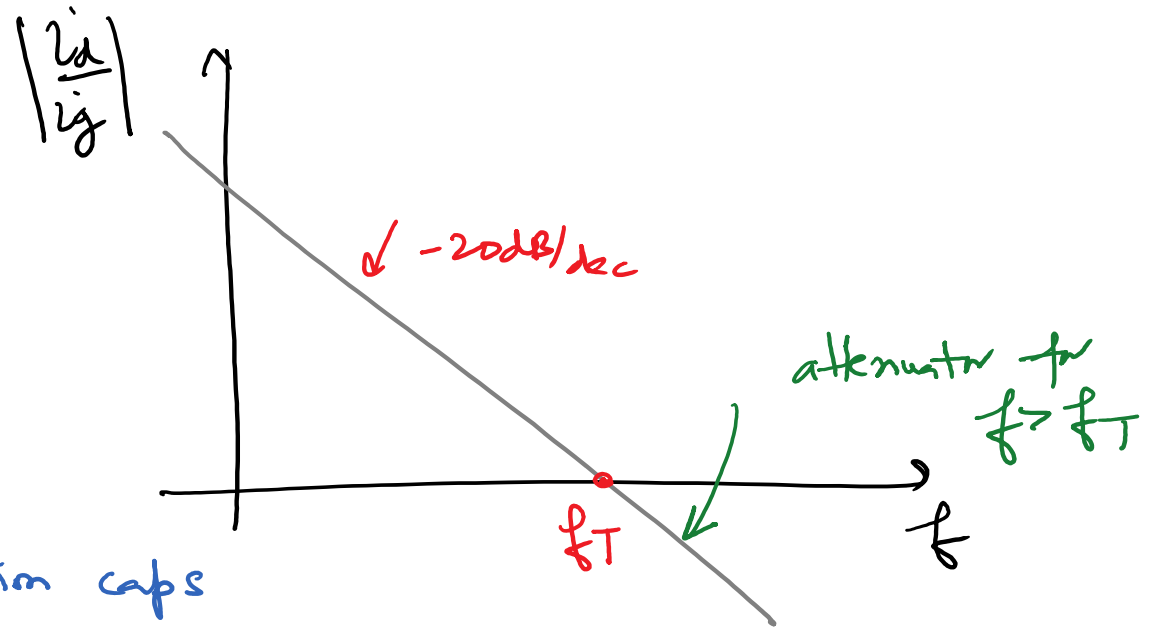
$$\Rightarrow \left| \frac{i_d}{i_g} \right| = \left| \frac{g_m}{2\pi f (C_{gd} + C_{gs})} \right| = 1$$

$$f_T = \frac{g_m}{2\pi (C_{gd} + C_{gs})} \approx \frac{g_m}{2\pi C_{gs}}, \quad \because C_{gs} \gg C_{gd}$$



$$f_T = \frac{g_m}{2\pi C_{gs}}$$

↳ ignores junction caps
and is defined such that ω_c is excluded



$$C_{gs} = \frac{2}{3} C_{ox} WL$$

$$f_T = \frac{g_m}{2\pi C_{gs}} = \frac{\mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{TH})}{2\pi \cdot \frac{2}{3} \cdot C_{ox} \cdot WL} = \frac{3\mu_n}{4\pi} \cdot \frac{V_{OV}}{L^2}$$

$$\Rightarrow \boxed{f_T \propto \frac{V_{OV}}{L^2}}$$

for high speed use minimum channel length
use large overdrive voltage

maximum gain that can be extracted from a transistor

$$A_{vo} = g_m r_o$$

$$= \sqrt{2\beta I_D} \cdot \frac{1}{\lambda I_D} = \frac{\sqrt{2 k_p n \frac{W}{L}}}{\lambda} \cdot \frac{1}{\sqrt{I_D}} \rightarrow I_D \uparrow \Rightarrow A_{vo} \downarrow$$

$$\Rightarrow g_m r_o = \frac{k_p n \frac{W}{L} \cdot V_{ov}}{\lambda \cdot \frac{k_p n \frac{W}{L} \cdot V_{ov}^2}} = \frac{2}{\lambda \cdot V_{ov}} \propto \frac{L}{V_{ov}}$$

$$\lambda \propto \frac{1}{L}$$

$$\boxed{g_m r_o \propto \frac{L}{V_{ov}}}$$

\Rightarrow open circuit gain \downarrow
 $L \downarrow$ or $V_{ov} \uparrow$

CMOS

In short-channel process

$$f_T \propto \frac{V_{ov}}{L}$$

$$g_{m,r_o} \propto \frac{L}{V_{ov}}$$

\therefore Devices are not square law in short channel CMOS

$$g_{FT} = g_{m,r_o} \cdot f_T \propto \frac{V_{ov}}{L} \cdot \frac{L}{V_{ov}} \propto \text{const.}$$

Trade off between

$g_{m,r_o} \leftarrow f_T$

$$V_{ov} \uparrow \Rightarrow f_T \uparrow + g_{m,r_o} \downarrow$$

With CMOS scaling $\Rightarrow L_{min} \downarrow$ in every generation
 $f_T \uparrow$ but $g_{m,r_o} \downarrow$

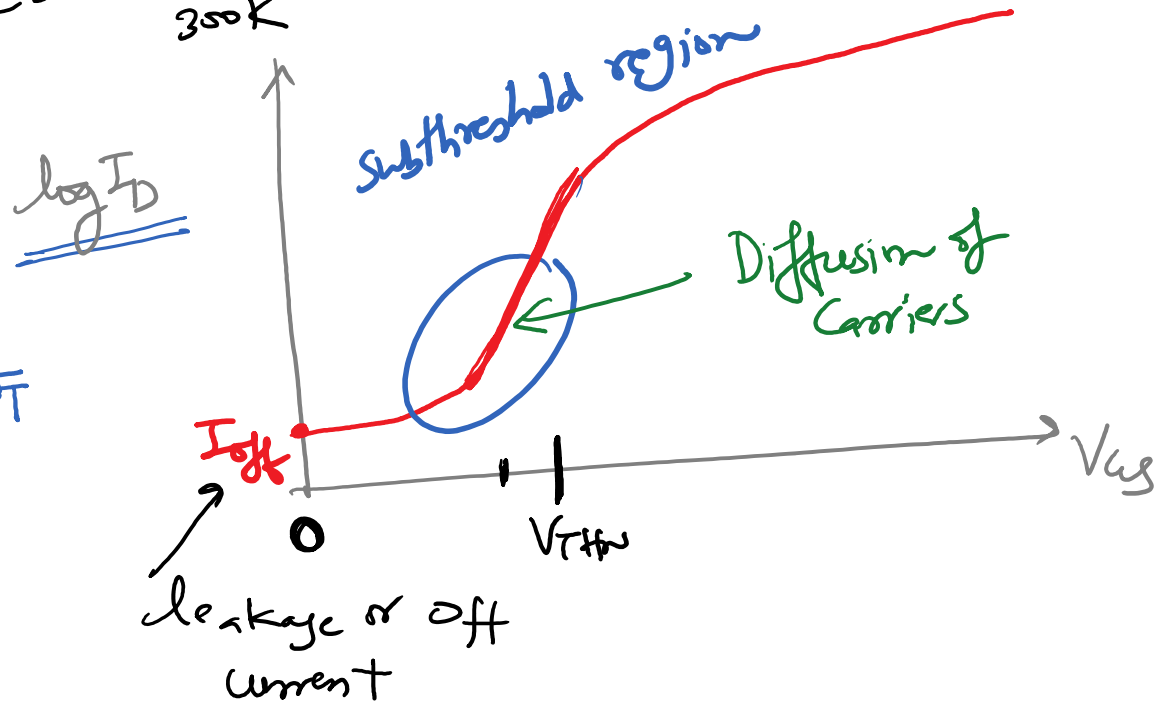
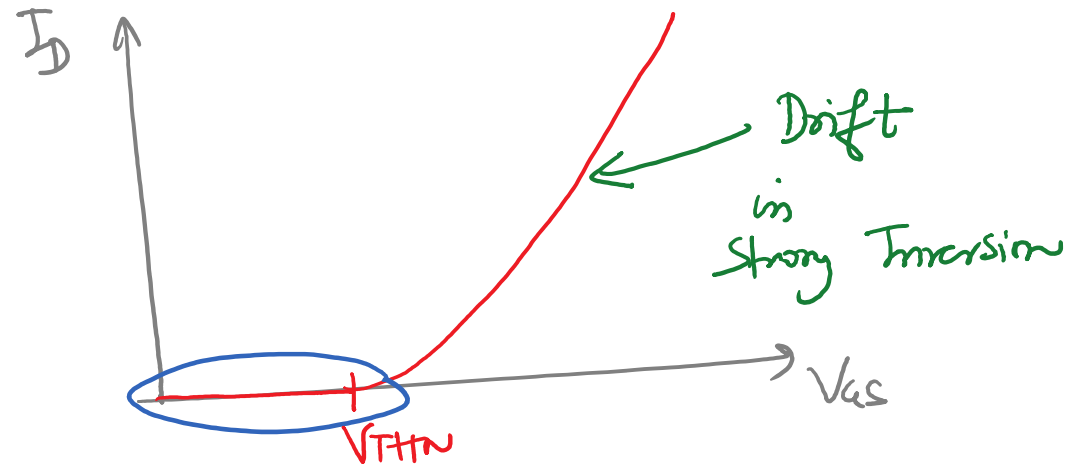
Subthreshold g_m

$$g_m \text{ Sub-}V_T \rightarrow e^{kn \log} \rightarrow kn$$

$$I_D = I_{D0} \cdot \frac{W}{L} e^{\frac{V_{GS} - V_{THN}}{nV_T}} \quad \frac{kT}{q} = 26 \text{ mV} @ RT_{300K}$$

$$g_m = \left. \frac{\partial I_D}{\partial V_{GS}} \right|_{V_{GS}} = I_{D0} \cdot \frac{W}{L} \cdot e^{\frac{V_{GS} - V_{THN}}{nV_T}} \cdot \frac{1}{nV_T}$$

$$g_m = \frac{I_D}{nV_T}$$



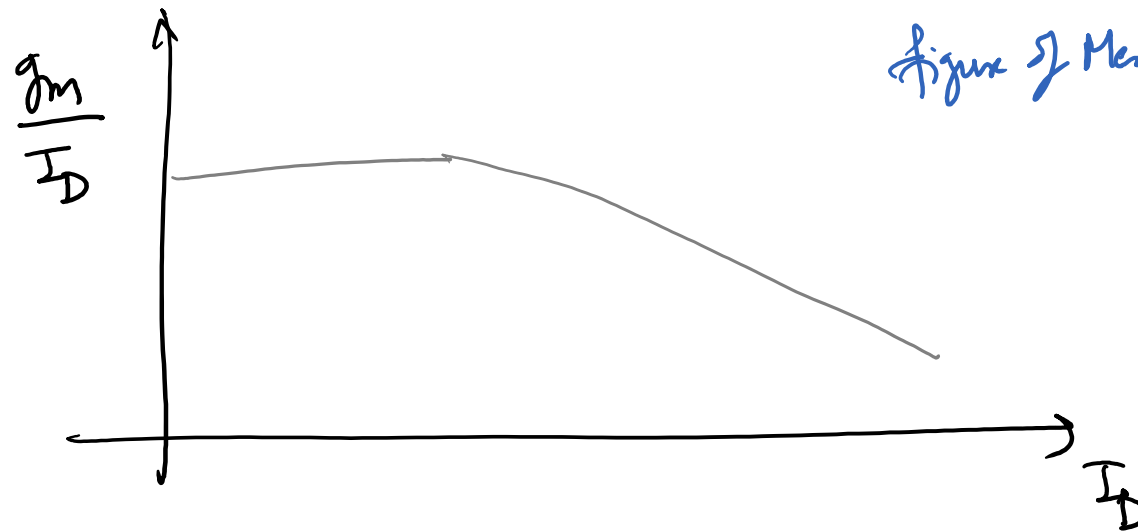
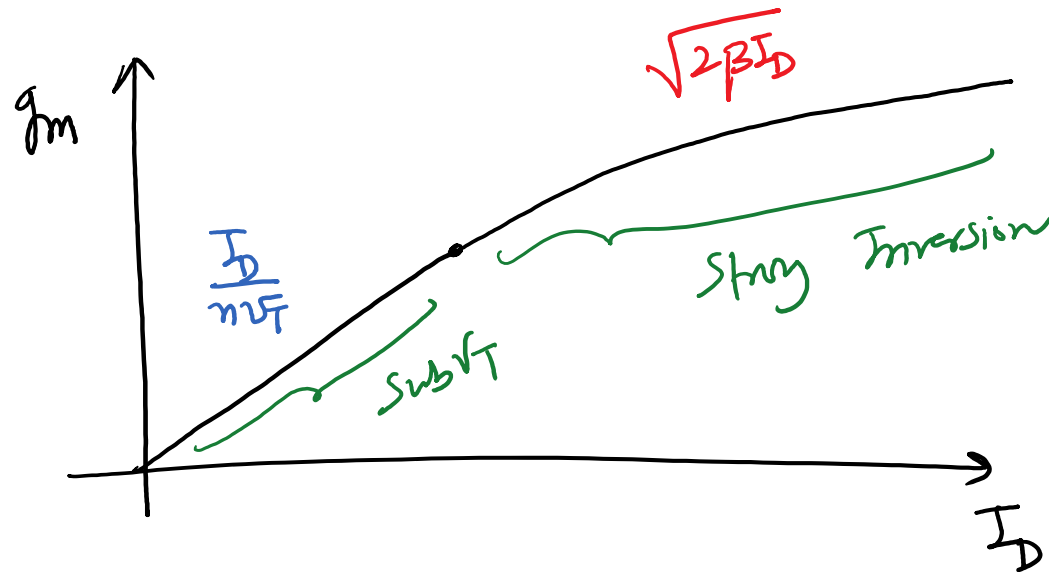
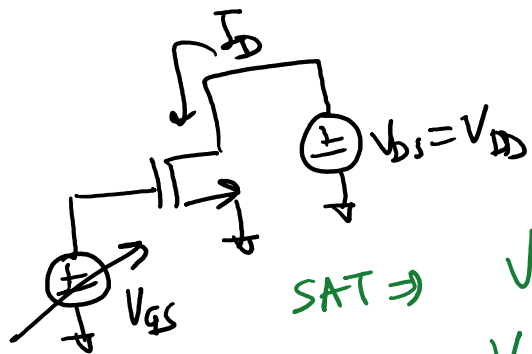
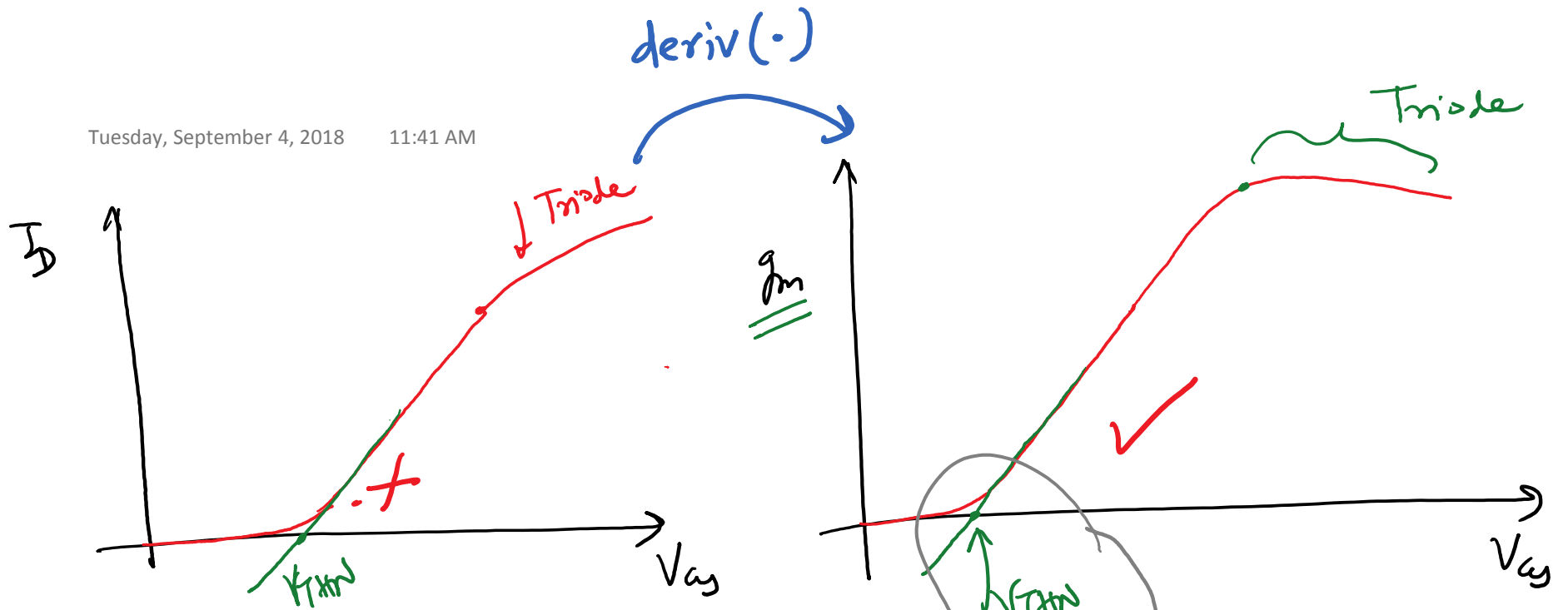


figure of Merit = $\frac{g_m}{I_D}$

In subVT \rightarrow we get
higher $\frac{g_m}{I_D}$ but
 f_T will be small
 $f_T < 1 \text{ MHz}$



SAT \Rightarrow

$$V_{DS} > V_{GS} - V_{THN}$$

$$V_{GS} < V_{DS} + V_{THN}$$

$V_{GS} = V_{THN}$ where $g_m = 0$.

MOSFET Temperature Effects

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$$I_D = \frac{\mu_n C_{ox}}{2} \frac{W}{L} (V_{GS} - V_{THN})^2$$

How does I_D change with T .

$$\mu_n = \mu_n(T)$$

$$V_{THN} = V_{THN}(T)$$

$$V_{THN} = -V_{ms} - 2V_{fp} + \frac{Q_{do}' - Q_{ss}'}{C_{ox}'}$$

\downarrow
 $\frac{kT}{q} \ln \frac{N_{D,poly}}{N_A}$

$$\Rightarrow \frac{\partial V_{THN}}{\partial T} \approx -\frac{k}{q} \ln \left(\frac{N_{D,poly}}{N_A} \right)$$

$$\boxed{\frac{\partial V_{THN}}{\partial T} \approx -1 \text{ mV}/^\circ\text{C}}$$

$$T \uparrow \Rightarrow V_{THN} \downarrow$$

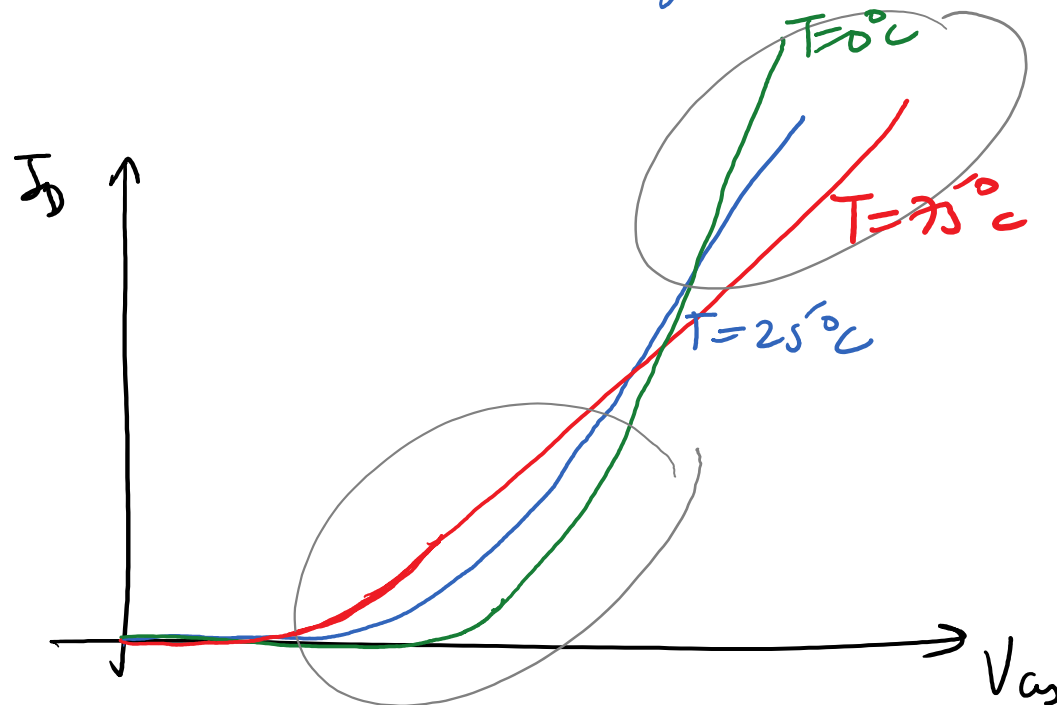
$$R_{ln} = \mu_n C_{ox}'$$

└ mobility

$T \uparrow$ scattering

$$v_{a \downarrow} \Rightarrow \mu_n \downarrow$$

$$\mu(T) \propto T^{-3/2}$$



$T=75^\circ\text{C}$

$T=0^\circ\text{C}$

$I_D \uparrow$ for small V_{ov}
 but $I_D \downarrow$ for larger
 $V_{GS} = V_{DD}$

Really depends upon the application.

$$\frac{\partial I_D}{\partial T} \propto \frac{(V_{GS} - V_{THN})}{2} \frac{\partial k_n}{\partial T} - k_n \cdot \frac{\partial V_{THN}}{\partial T}$$

from dominates at higher V_{GS} .