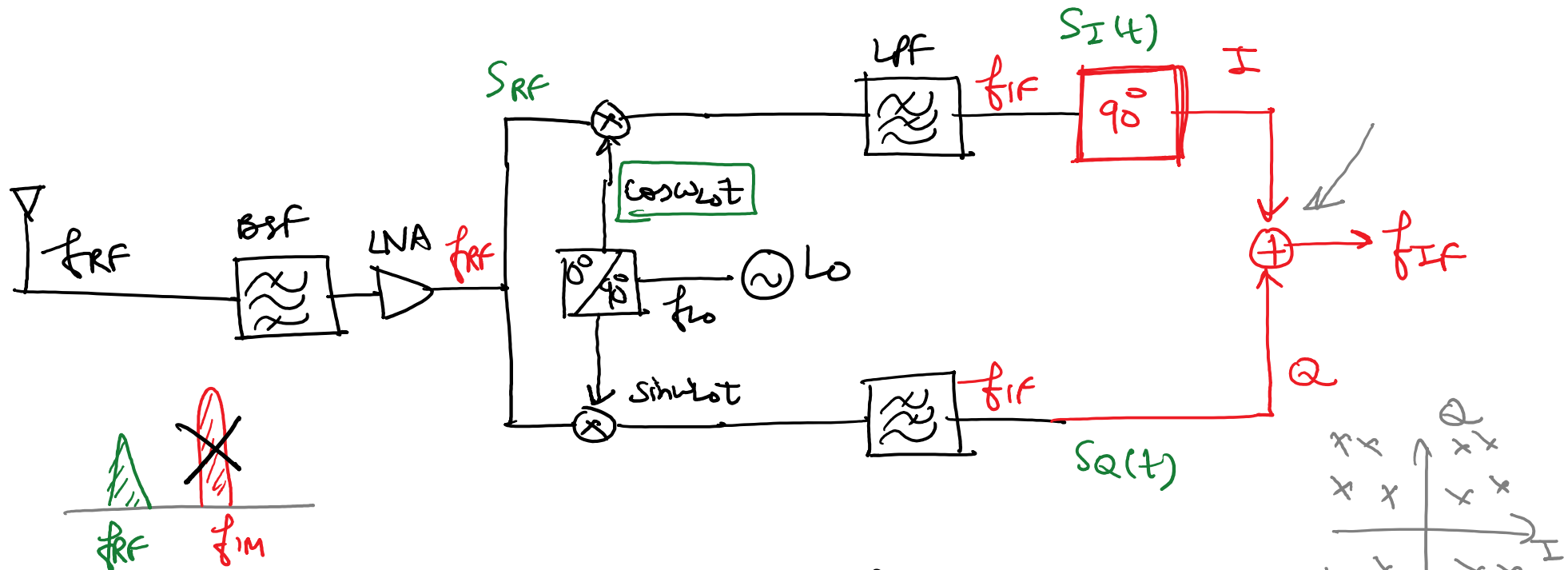


ECE 513- Lecture 5

Tuesday, September 4, 2018 9:20 AM

Hartley Image Reject Receivers



$$S_{RF} = A_{RF} \cos(\omega_{RF}t + \alpha_{RF})$$

$$S_{IM} = A_{IM} \cos(\omega_{IM}t + \alpha_{IM})$$

$$f_{RF} = f_{LO} - f_{IF} \quad \& \quad f_{IM} = f_{LO} + f_{IF}$$

$$\Rightarrow \omega_{RF} = \omega_{LO} - \omega_{IF} \quad \& \quad \omega_{IM} = \omega_{LO} + \omega_{IF}$$

$$S_I(t) = A_{RF} \cos(\omega_{RF}t + \alpha_{RF}) \cdot \cos(\omega_{LO}t) + A_{IM} \cos(\omega_{IM}t + \alpha_{IM}) \cdot \cos(\omega_{LO}t)$$

↓ multiply $\cos A \cos B$, LPF and arrange the terms

$$S_I(t) = \frac{A_{RF}}{2} \cos(\omega_{IF}t - \alpha_{RF}) + \frac{A_{IM}}{2} \cos(\omega_{IF}t + \alpha_{IM}) \longrightarrow \textcircled{1}$$

↓ 90° phase shift

$$S_{I90}(t) = \frac{A_{RF}}{2} \sin(\omega_{IF}t - \alpha_{RF}) + \frac{A_{IM}}{2} \sin(\omega_{IF}t + \alpha_{IM}) \longrightarrow \textcircled{2}$$

Similarly for the Q-path

$$S_Q(t) = A_{RF} \cos(\omega_{RF}t + \alpha_{RF}) \cdot \sin(\omega_{LO}t) + A_{IM} \cos(\omega_{IM}t + \alpha_{IM}) \cdot \sin(\omega_{LO}t)$$

↓ LPF and rearrange terms

$$S_Q(t) = \frac{A_{RF}}{2} \sin(\omega_{IF}t - \alpha_{RF}) - \frac{A_{IM}}{2} \sin(\omega_{IF}t + \alpha_{IM}) \longrightarrow \textcircled{3}$$

add ϵ_2 & ϵ_3

$$S_{IF}(t) = \underbrace{A_{RF}}_{\text{recovery}} \sin(\omega_{IF}t - \underbrace{\phi_{RF}}_{\text{recovery}})$$

→ recover original modulating bits

* Image Reject Rx are sensitive to

① Amplitude mismatch and phase error between the I & Q paths

→ monolithic integration

② If the LO phases ($\cos \omega_{LO}t$ & $\sin \omega_{LO}t$) are not in exact quadrature $\Rightarrow$ I & Q terms are not identical

→ fraction of image remains at f_{IF} .

Razavi 2nd ed. Chap-4

Image Reject Ratio (IRR)

If the LO waveforms are

$$(1+\varepsilon) \cos(\omega_{LO}t + \Delta\theta) \quad \& \quad \sin(\omega_{LO}t)$$

Amplitude mismatch
Phase error

Received & Image input signal

$$\Rightarrow A_{RF} \cos(\omega_{RF}t + \alpha_{RF}) + A_{IM} \cos(\omega_{IM}t + \alpha_{IM})$$

$$\left. \frac{P_{IM}}{P_{RF}} \right|_{out} = \frac{A_{IM}^2}{A_{RF}^2} \cdot \underbrace{\left(\frac{(1+\varepsilon)^2 - 2(1+\varepsilon)\cos\Delta\theta + 1}{(1+\varepsilon)^2 + 2(1+\varepsilon)\cos\Delta\theta + 1} \right)}_{IRR} \approx \frac{4}{\varepsilon^2 + \Delta\theta^2}$$

ratio of powers out

for $\varepsilon \ll 1$
 $\Delta\theta \ll 1 \text{ rad}$

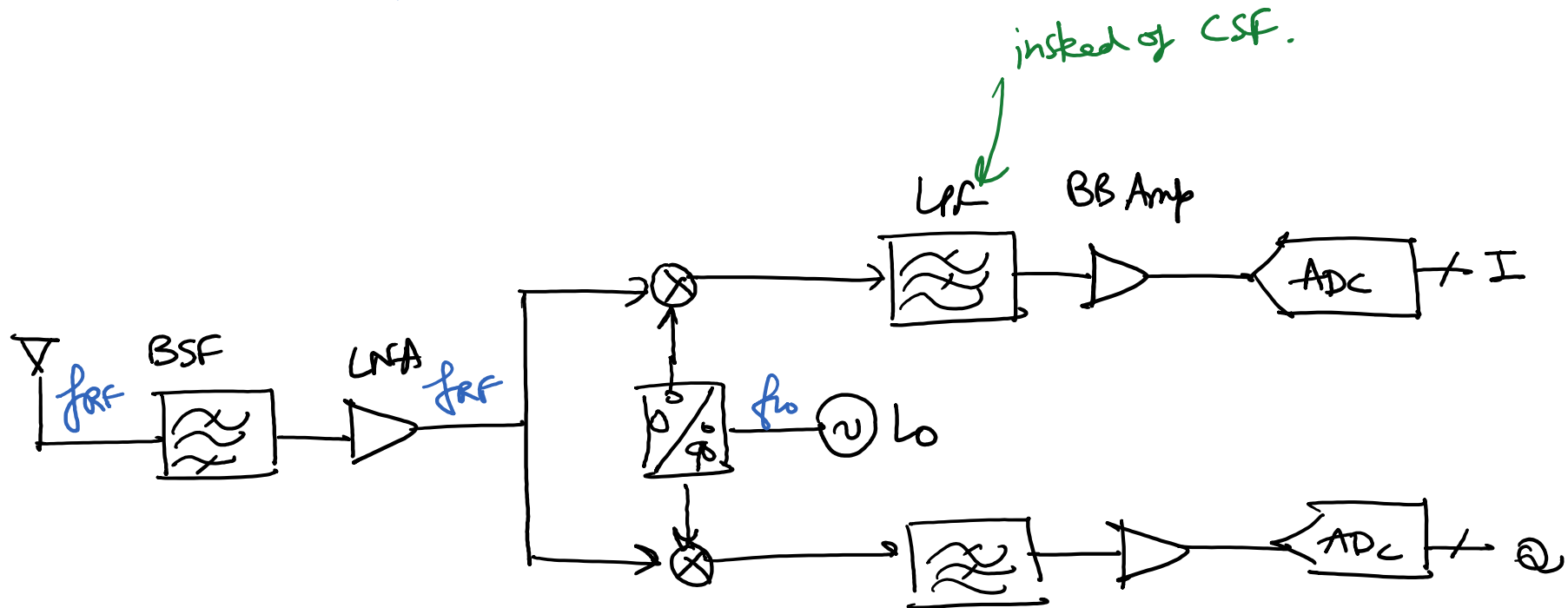
at the output

Weaver Architecture $\Leftarrow$ similar to Hartley Receiver
 $\uparrow$ Read from the Textbook

Direct Conversion (Zero-IF) Receiver

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↳ Colpitts, 1924



$$f_{RF} = f_{LO} \Rightarrow f_{IF} = 0$$

↳ the modulated carrier is downconverted to baseband
↳ special case of heterodyne RX

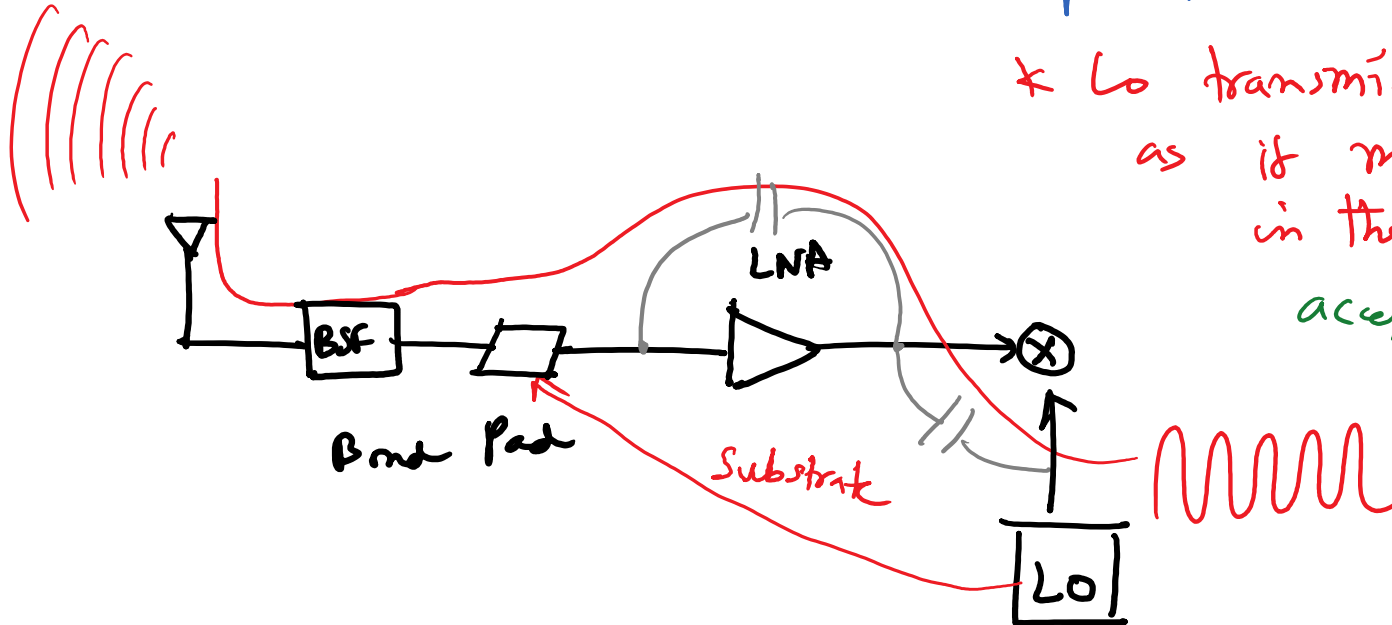
(+) Image problem disappears $\Rightarrow$ Signal is its own image
 $\hookrightarrow$ IRF is no longer needed
 $\hookrightarrow$ CSF is replaced by a LPF

Issues with Zero-IF Rx's

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10:10 AM

LO leakage:



* LO couples to the antenna thru parasitic capacitance & substrate

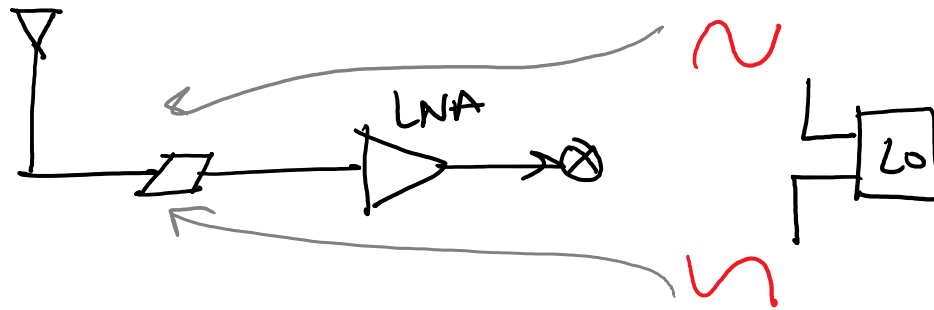
* LO transmission is undesirable as it may affect other Rx in the band

acceptable value is

-50 to -70 dBm
measured at the antenna

* LO leakage is suppressed by the BSF in heterodyne architecture ($f_{RF} = f_{LO} - f_{IF}$)

* LO leakage is avoided by symmetric layout

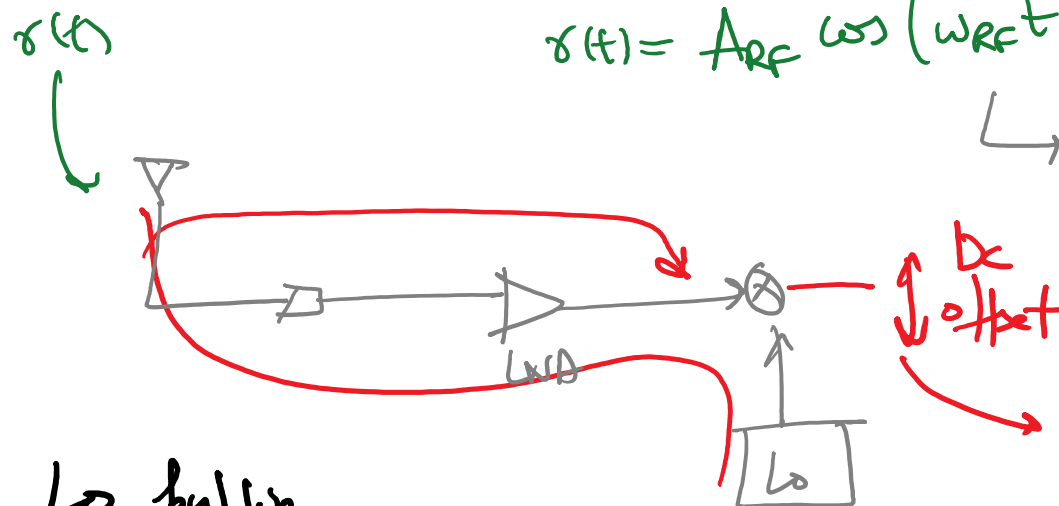


* DC offset due to LO self-mixing

$$s(t) = A_{RF} \cos(\omega_{RF}t + \phi_{RF}) \cdot \cos(\omega_{LO}t)$$

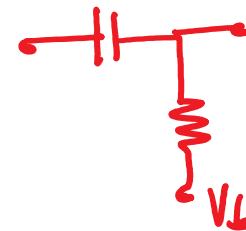
↳ DC terms

$$\omega^2 g = \frac{1 + \cos 2\alpha}{2}$$



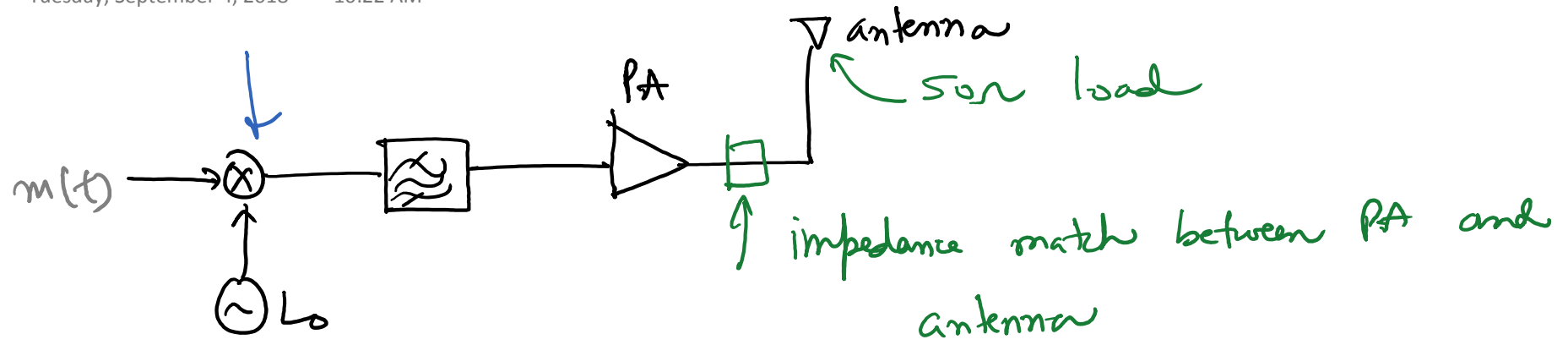
* LO pulling

* Even-order nonlinearities.



RF Transmitter Architectures

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* PA is a key component in the transmitter

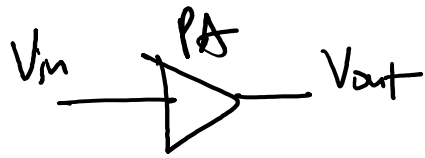
The modulation of the carrier is performed:

- ① Linearly by frequency translation using a mixer
- ② By directly modulating the carrier amplitude, frequency, phase or pulse width

↳ Direct modulation is implemented either in linear or non-linear fashion.

* Depending upon the type of modulation

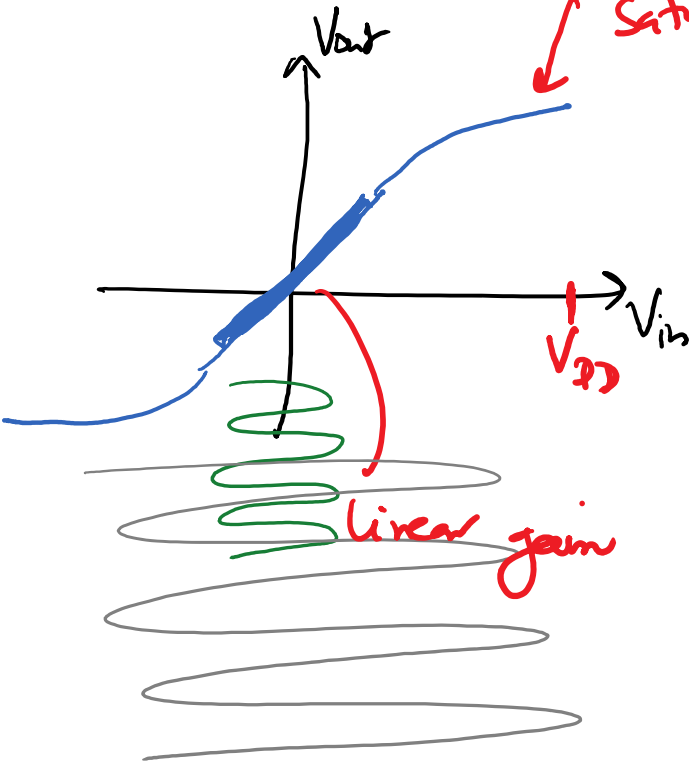
↳ PA operates in linear or saturated regime



gain is
Saturated

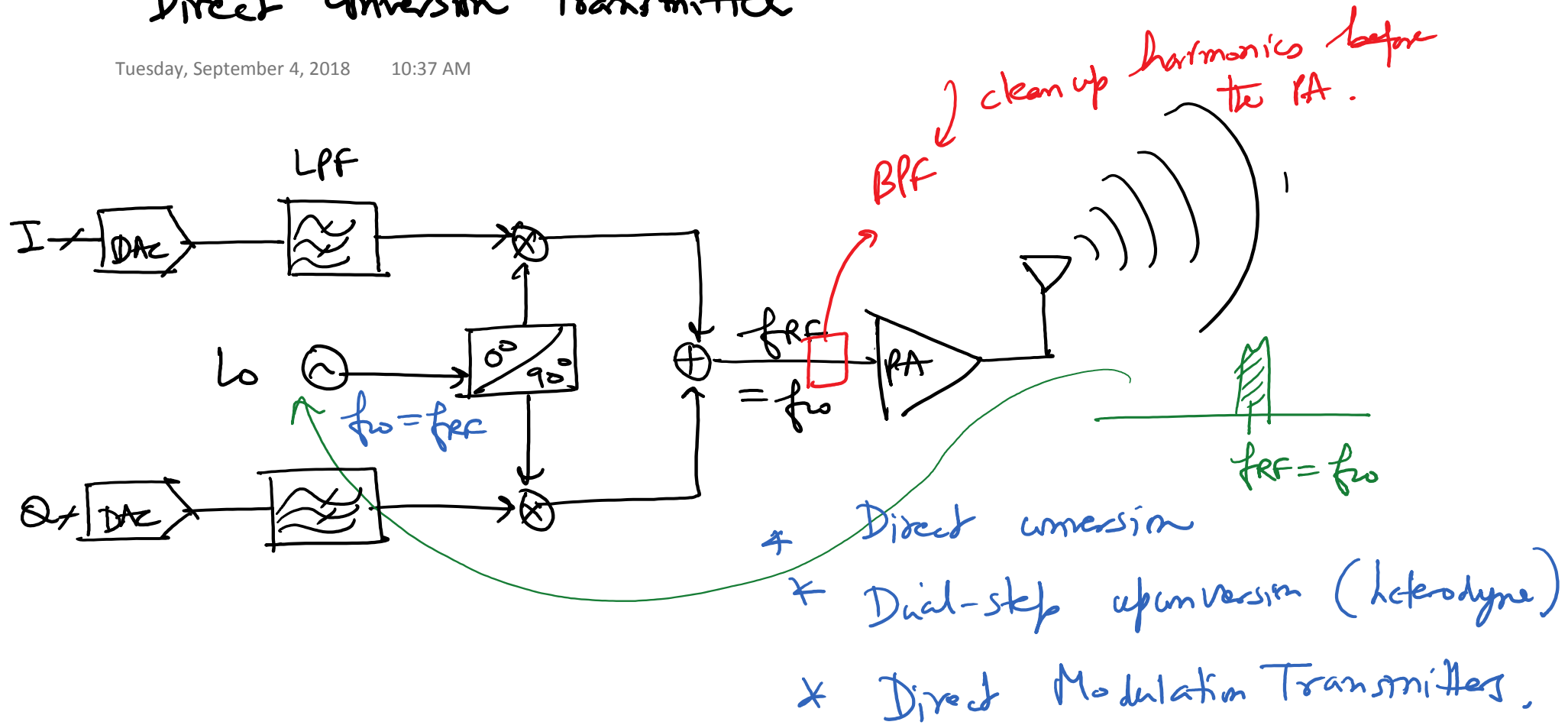
$$P_T \Rightarrow \frac{V_{sgn}^2}{Z_0}$$

has ramifications for the
efficiency and power consumption
of the entire TX.



Direct Conversion Transmitter

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- (-) Stringent matching between I & Q paths
- (-) PA signal "pulls" the oscillator and changes its frequency $\Rightarrow$ Need good LO isolation/buffering

[READ from the Book]