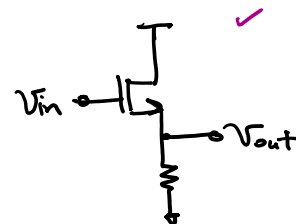
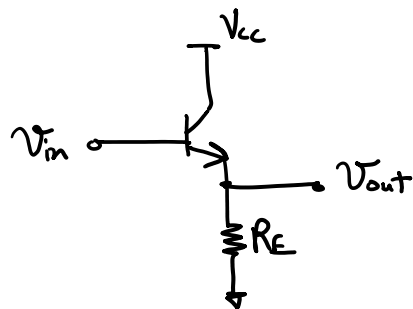


ICE 310 - Lecture 33

Wednesday, April 18, 2018 10:27 AM

Emitter follower (Common Collector)

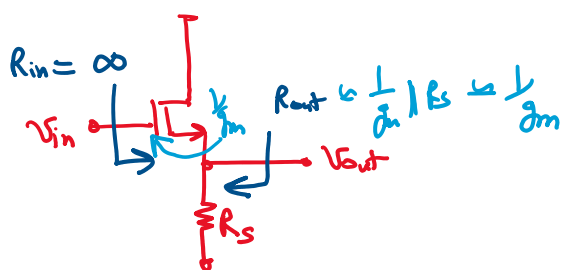
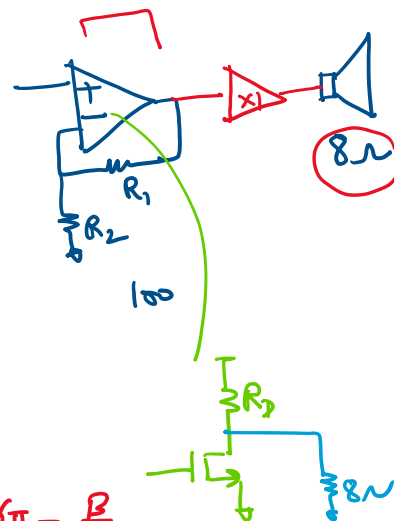


$$\frac{V_{out}}{V_{in}} = \frac{1}{1 + \frac{r_{\pi}}{(\beta+1)} \cdot \frac{1}{R_E}}$$

$$\approx \frac{R_E}{R_E + \frac{1}{g_m}}$$

$$\frac{r_{\pi}}{\beta+1} \approx \frac{r_{\pi}}{\beta} = \frac{1}{g_m}$$

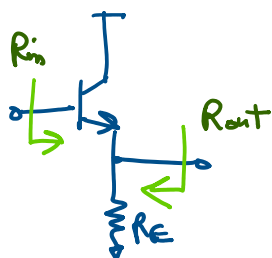
$$\therefore r_{\pi} = \frac{\beta}{g_m}$$



$$R_{in} \rightarrow \infty$$

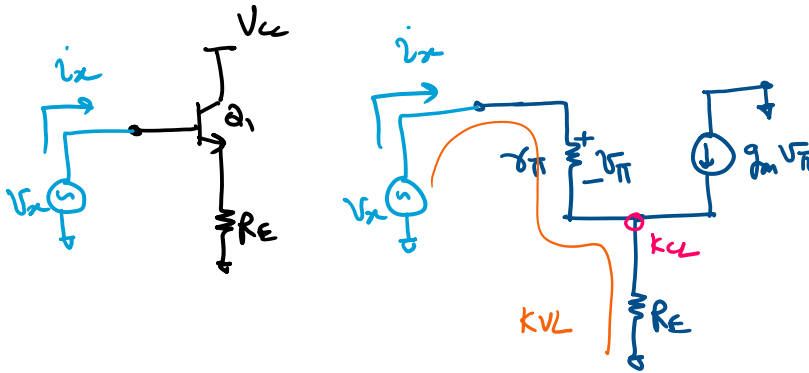
$$R_{out} \approx \frac{1}{g_m} \rightarrow 0$$

Need to derive R_{in} & R_{out} of Emitter follower circuit



$$V_A \rightarrow \infty$$

$$r_o \rightarrow \infty$$



$$V_x = V_{\pi} + (i_x + g_m V_{\pi}) R_E \quad \text{--- ①}$$

$$V_{\pi} = i_x r_{\pi} \quad \text{--- ②}$$

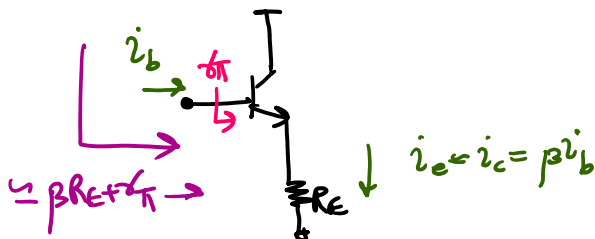
$$\therefore r_{\pi} = \frac{\beta}{g_m}$$

$$\Rightarrow g_m r_{\pi} = \beta$$

$$\Rightarrow V_x = i_x r_{\pi} + (i_x + \underbrace{g_m r_{\pi}}_{\beta} i_x) R_E$$

$$\Rightarrow V_x = i_x r_{\pi} + (\beta + 1) i_x R_E$$

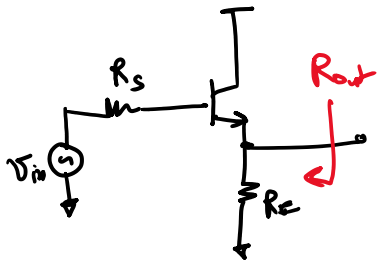
$$\Rightarrow \boxed{R_{in} = \frac{V_x}{i_x} = r_{\pi} + (\beta + 1) R_E}$$



Emitter follower "transforms" the load resistor, R_E , to a much larger value $(\beta + 1) R_E + r_{\pi}$

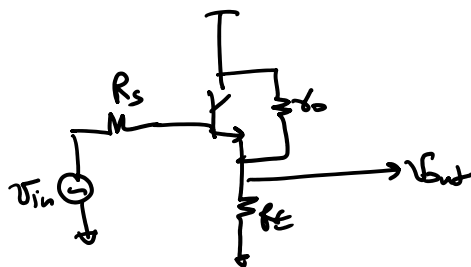
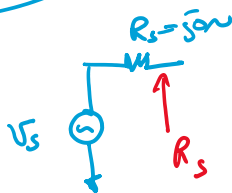
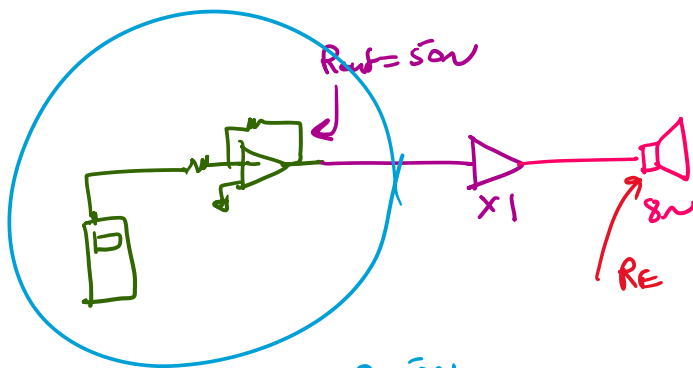
$\rightarrow$ forms an efficient voltage buffer by driving $R_{in} \rightarrow \infty$

by doing $K_{in} \rightarrow \infty$



$$R_{out} = \left(\frac{R_s}{\beta+1} + \frac{1}{g_m} \right) \parallel R_E \approx \frac{R_s}{\beta+1} + \frac{1}{g_m}$$

Does a good job of providing a low R_{out} $\rightarrow$ good voltage buffer.

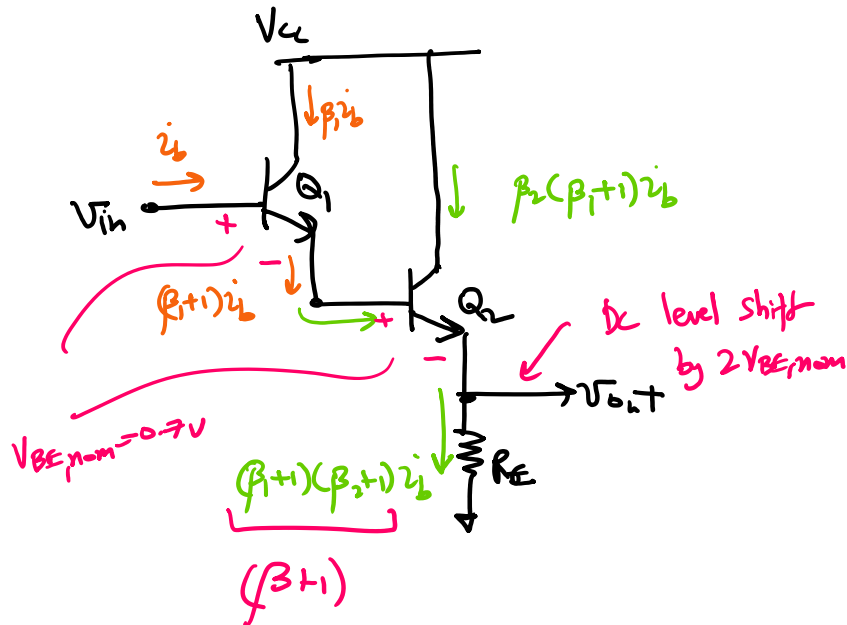


$$A_v = \frac{R_E \parallel r_o}{R_E \parallel r_o + \frac{R_s}{\beta+1} + \frac{1}{g_m}}$$

$$R_{in} = r_{\pi} + (\beta+1)(R_E \parallel r_o)$$

$$R_{out} = \left(\frac{R_s}{\beta+1} + \frac{1}{g_m} \right) \parallel R_E \parallel r_o$$

Darlington Pair $\equiv$ Two Emitter Followers in Cascade



$$\beta + 1 = (\beta_1 + 1)(\beta_2 + 1)$$

$$\beta \approx \beta_1 / \beta_2 = \beta_o^2$$

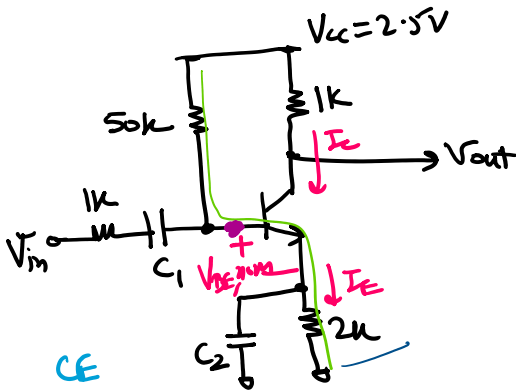
$$\text{If } \beta_1 = \beta_2 = \beta_o$$

$$\text{Current gain} = \beta_1 / \beta_2$$

$$I_B \ll I_E$$

Excellent Voltage Buffer

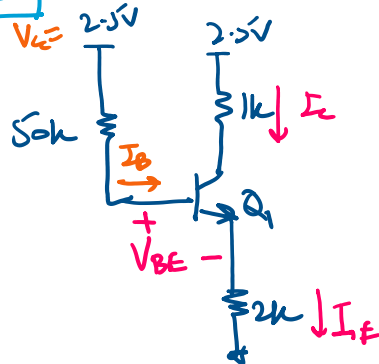
Example 5.52:



$I_B = 8 \times 10^{-16} A$, $\beta = 100$, $V_A = \infty$, $\delta_o \rightarrow \infty$ Caps are very large

(Do not ignore base current)

DC picture



$$I_B = \frac{V_{CC} - V_{BE} - I_E \times 2k}{50k} \quad I_E \approx I_C$$

$$I_C = \beta I_B = \frac{\beta (V_{CC} - V_{BE} - I_C \times 2k)}{50k}$$

$$I_C = \frac{100 (2.5 - V_{BE})}{50k + (\beta + 1) 2k}$$

$$I_C = \frac{100 (2.5 - V_{BE})}{252k}$$

$I_C \approx I_B e^{V_{BE}/V_T}$

Instead of iterative solution, use $V_{BE, nom} = 0.7V$

0.7V

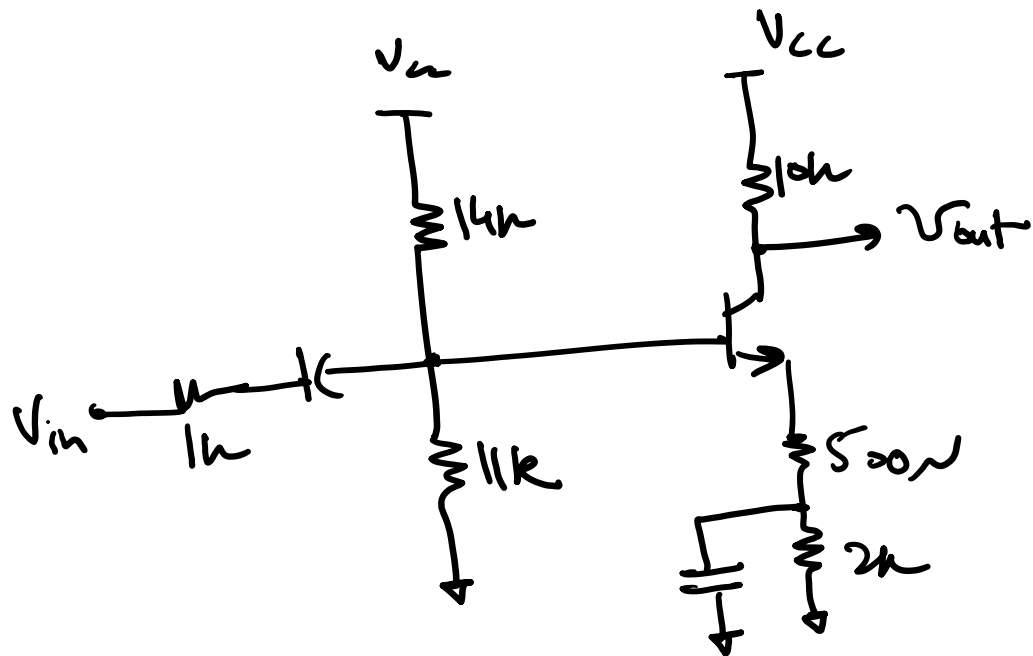
$$I_c = \frac{100 \times (2.5 - V_{BE, \text{nom}})}{252 \text{ k}} = 0.714 \text{ mA}$$

0.7V

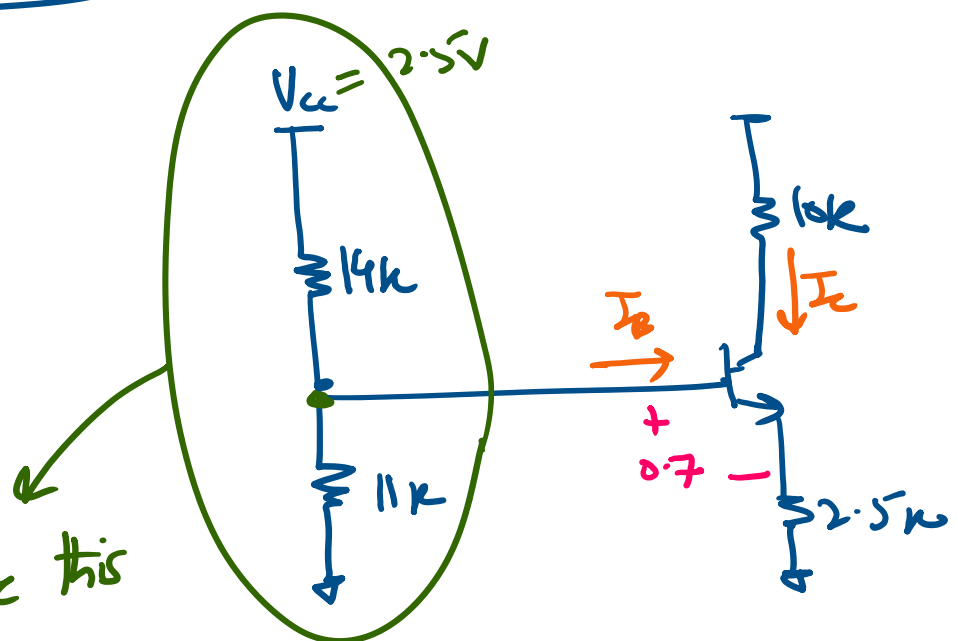
$$g_m = \frac{I_c}{V_T} = \frac{0.714 \text{ mA}}{26 \text{ mV}} = 27.5 \frac{\text{mA}}{\text{V}}$$

$$r_{\pi} = \frac{\beta}{g_m} = 364 \text{ k}\Omega$$

Problem 5.52 (c)



DC picture:



Thevenize this
part of the circuit

part of the circuit

