

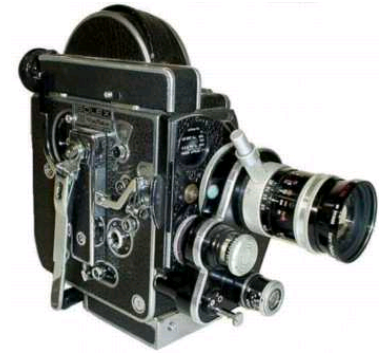
Compressive Sensing

A New Framework for Sparse Signal Acquisition and Processing

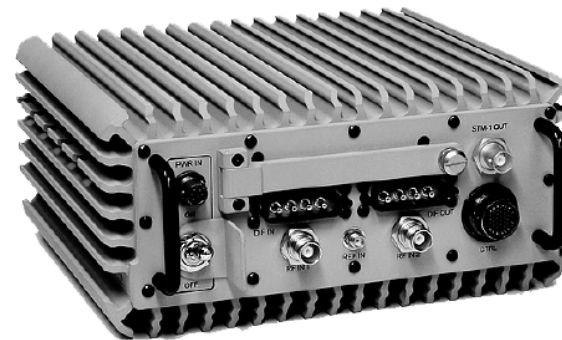
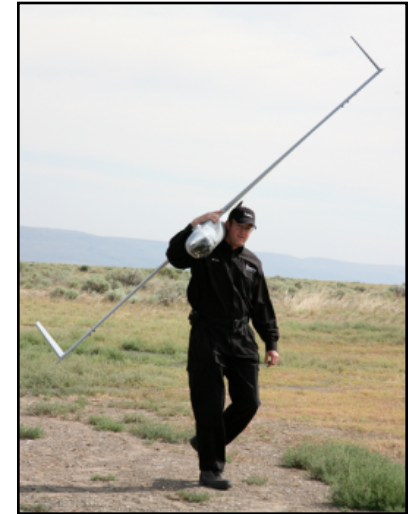
Richard Baraniuk

Rice University

RICE UNIVERSITY



Better, Stronger, Faster

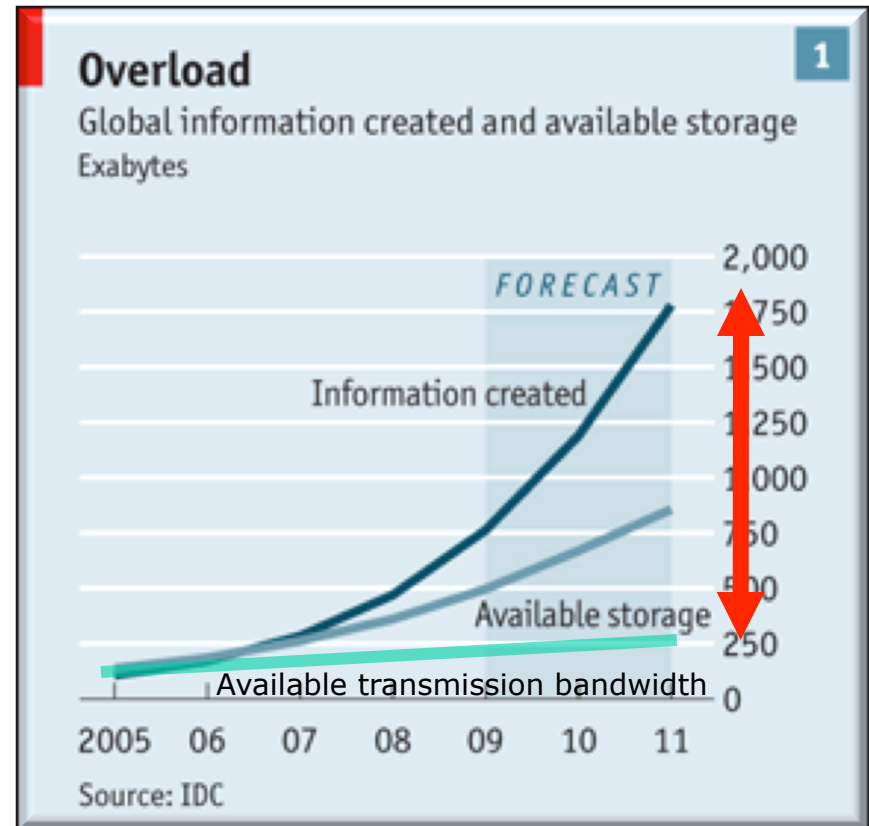


Accelerating Data Deluge

- **1250 billion gigabytes** generated in 2010

- # digital bits > # stars in the universe
- growing by a factor of 10 every 5 years

- Total data **generated** > **total storage**

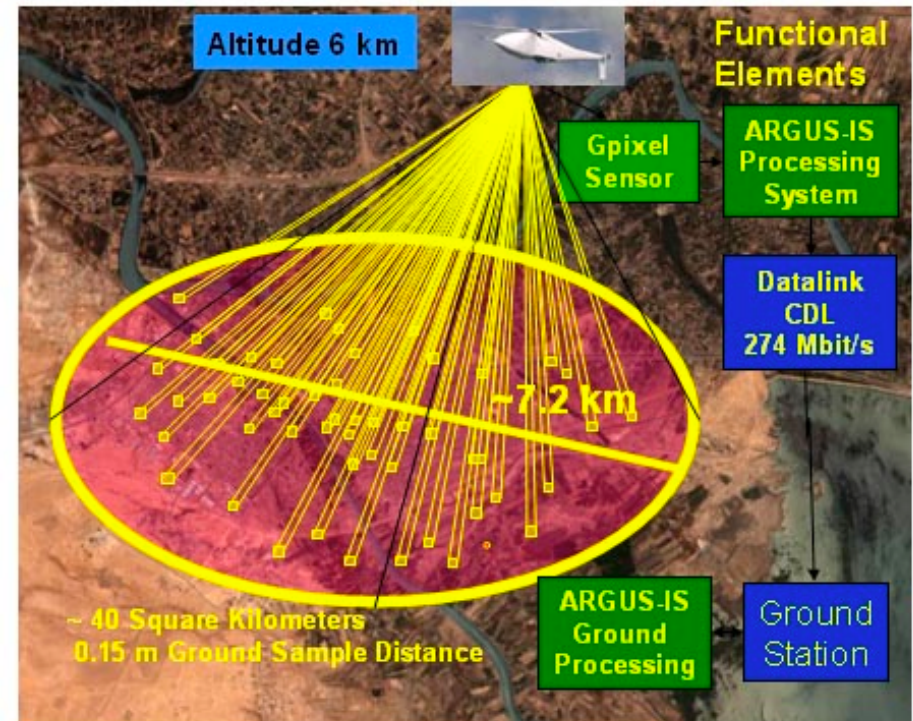


- Increases in **generation rate** >> increases in **transmission rate**

Case in Point: DARPA ARGUS-IS

- 1.8 Gpixel image sensor
 - video rate output:
770 Gbits/s
 - data rate input:
274 Mbits/s

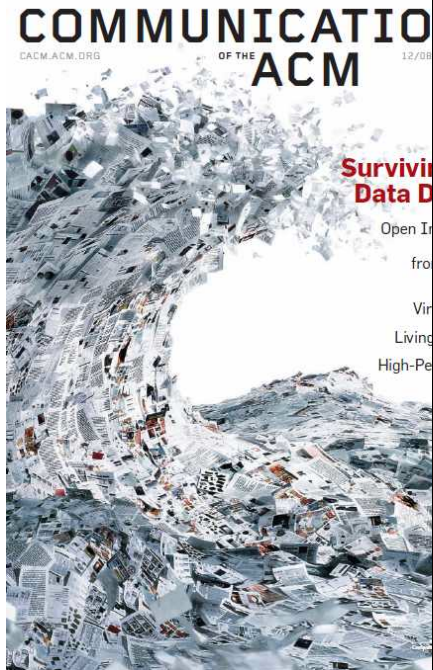
factor of **2800x**
way out of reach of
existing compression
technology



- **Reconnaissance without conscience**

- too much data to transmit to a ground station
- too much data to make effective real-time decisions

Accelerating Data Deluge



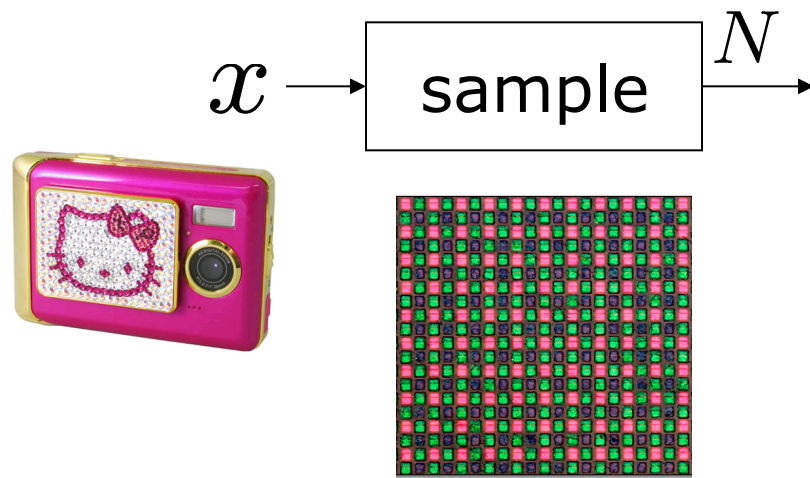
Today's Menu

- What's wrong with today's sensor systems?
why go to all the work to acquire massive amounts of multimedia data only to throw much/most of it away?
- One way out: dimensionality reduction (compressive sensing)
enables the design of radically new sensors and systems
- Theory: mathematics of sparsity
new nonlinear signal models and recovery algorithms
- Practice: compressive sensing in action
new cameras, imagers, ADCs, ...

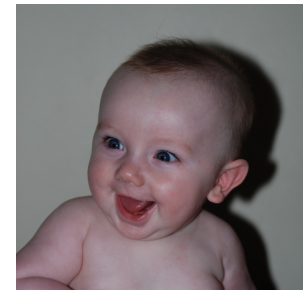
Sense by *Sampling*



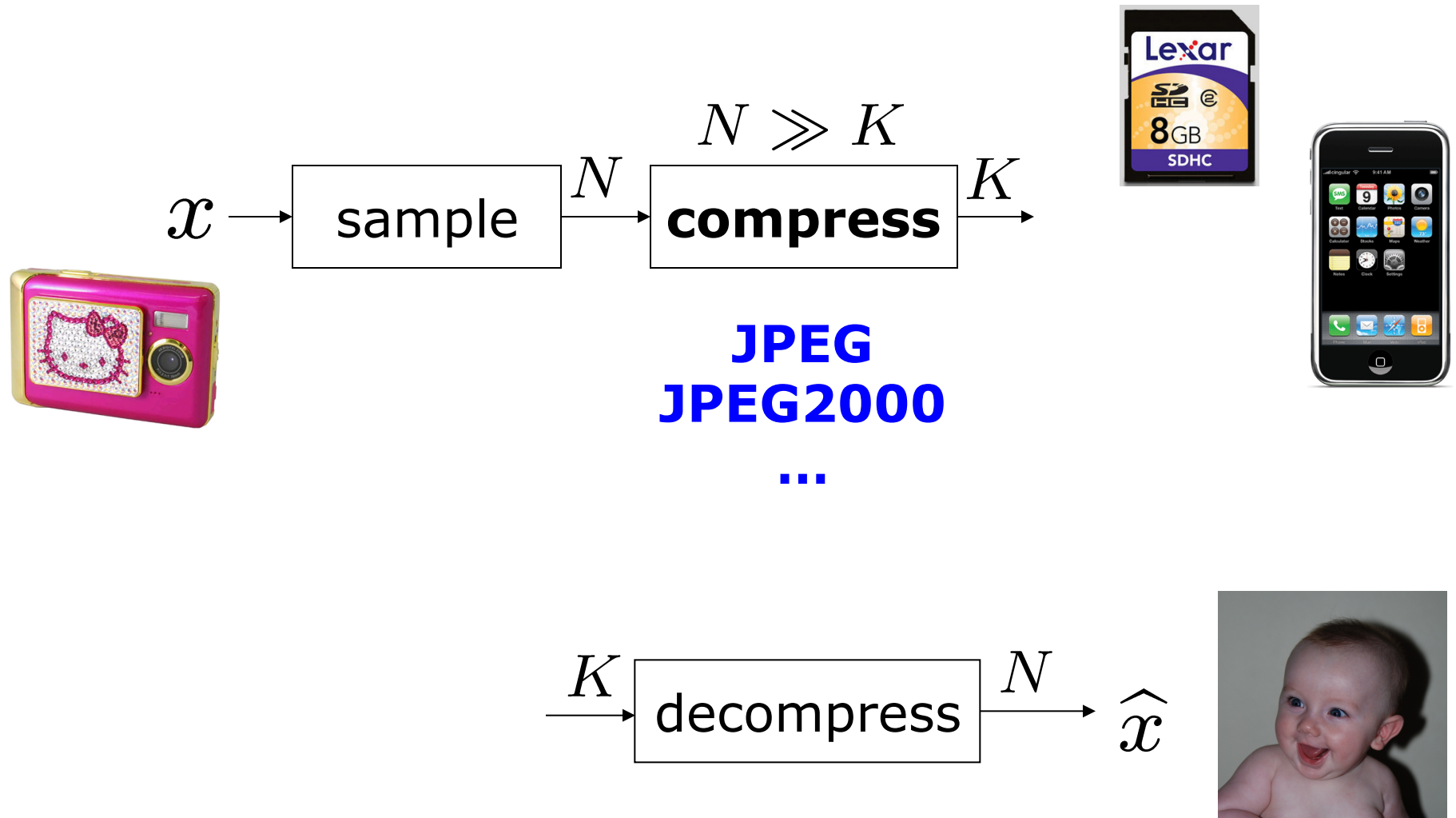
Sense by *Sampling*



**too
much
data!**

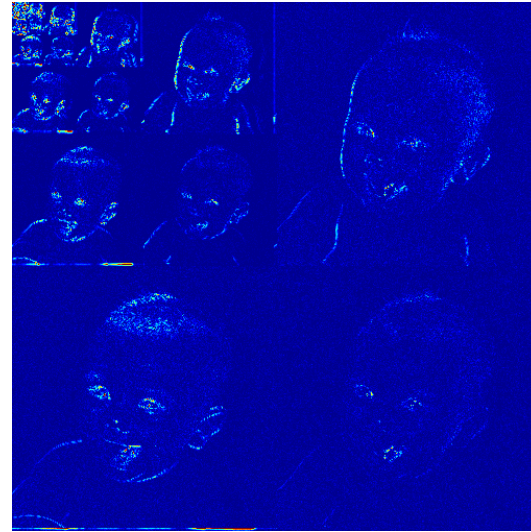


Sense then *Compress*



Sparsity

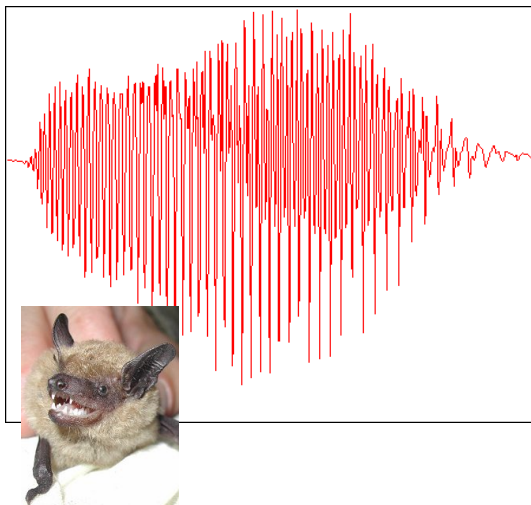
N
pixels



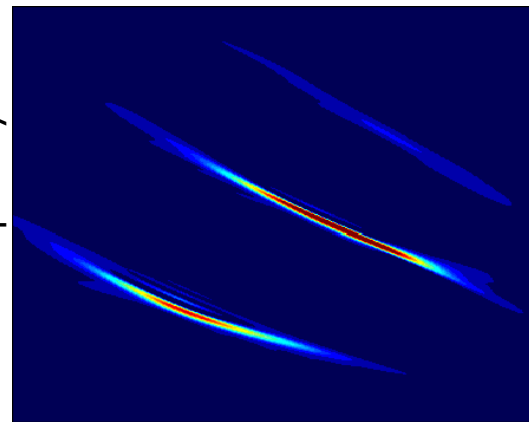
$K \ll N$
large
wavelet
coefficients

(blue = 0)

N
wideband
signal
samples



frequency

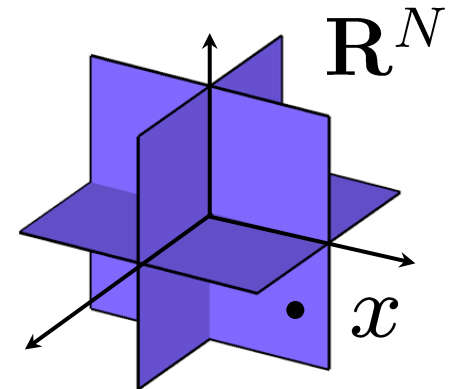


time

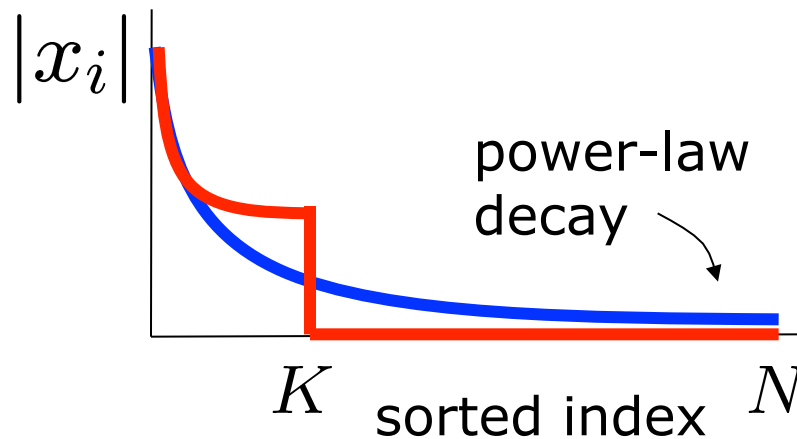
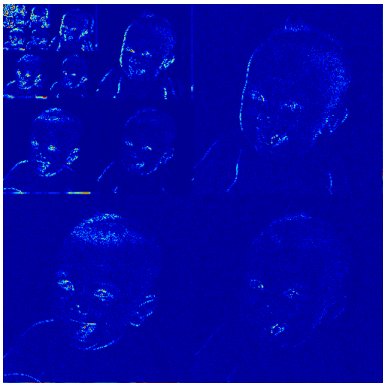
$K \ll N$
large
Gabor (TF)
coefficients

Concise Signal Structure

- **Sparse** signal: only K out of N coordinates nonzero
 - model: union of K -dimensional subspaces

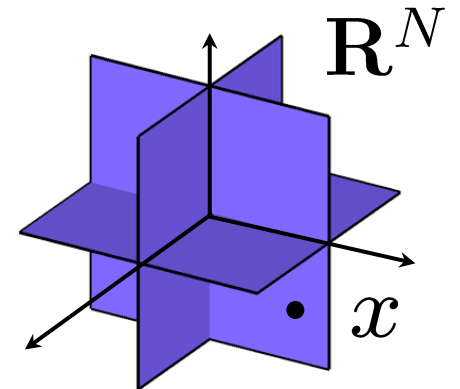


- **Compressible** signal: sorted coordinates decay rapidly with power-law

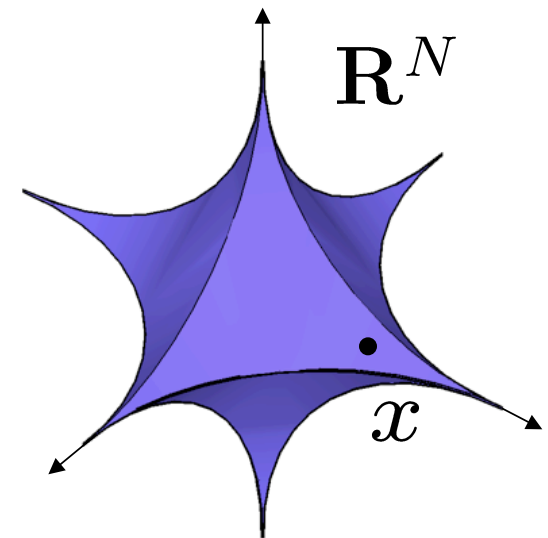
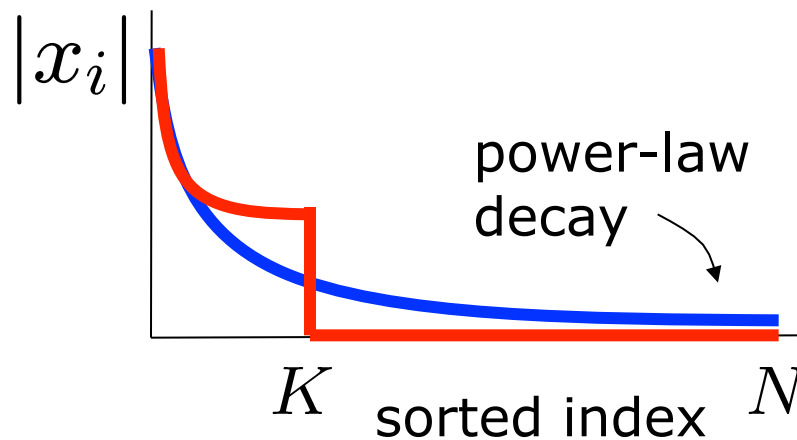


Concise Signal Structure

- **Sparse** signal: only K out of N coordinates nonzero
 - model: union of K -dimensional subspaces

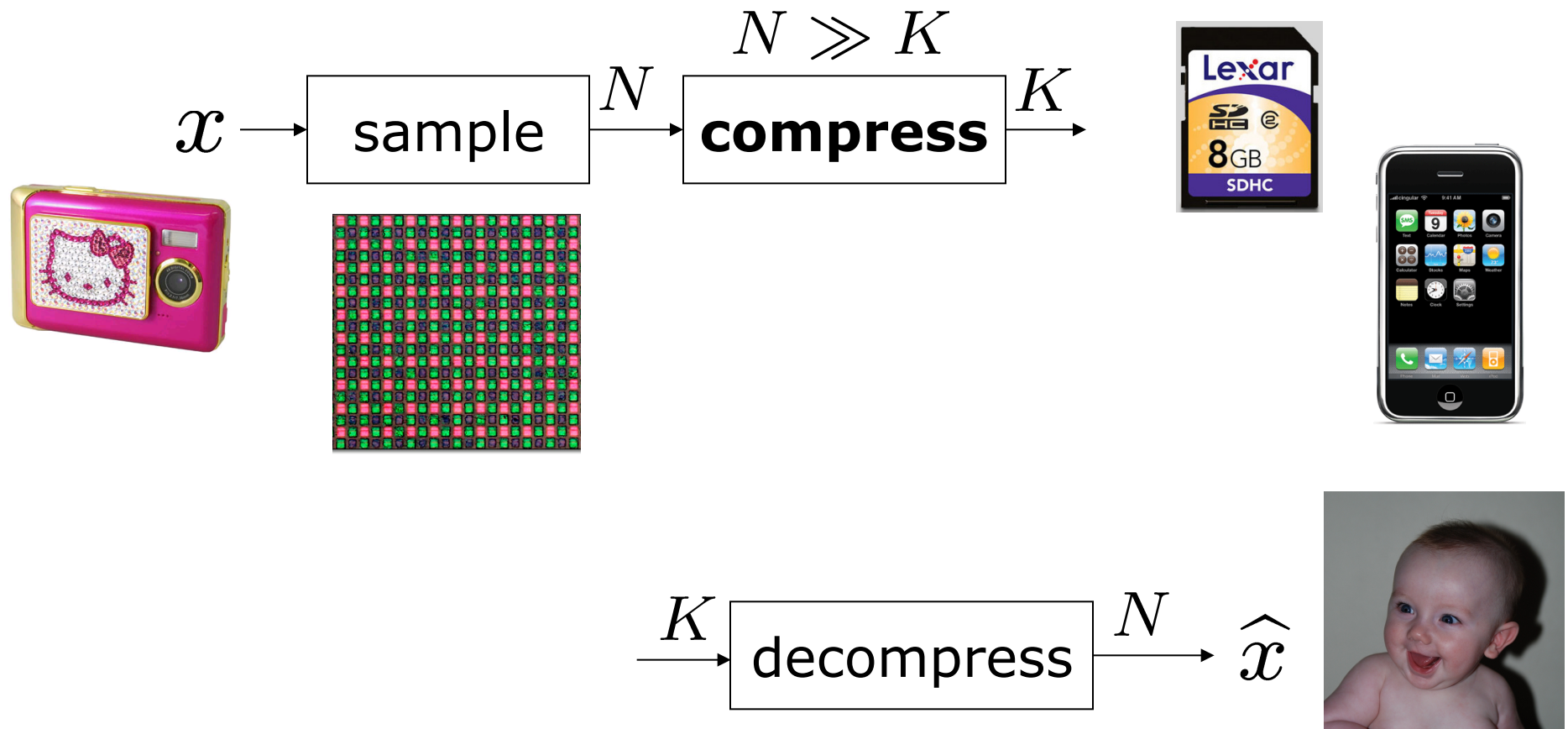


- **Compressible** signal: sorted coordinates decay rapidly with power-law
 - model: ℓ_p ball: $\|x\|_p^p = \sum_i |x_i|^p \leq 1, p \leq 1$

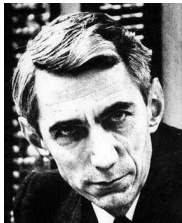


What's Wrong with this Picture?

- **Why go to all the work to acquire N samples only to discard all but K pieces of data?**

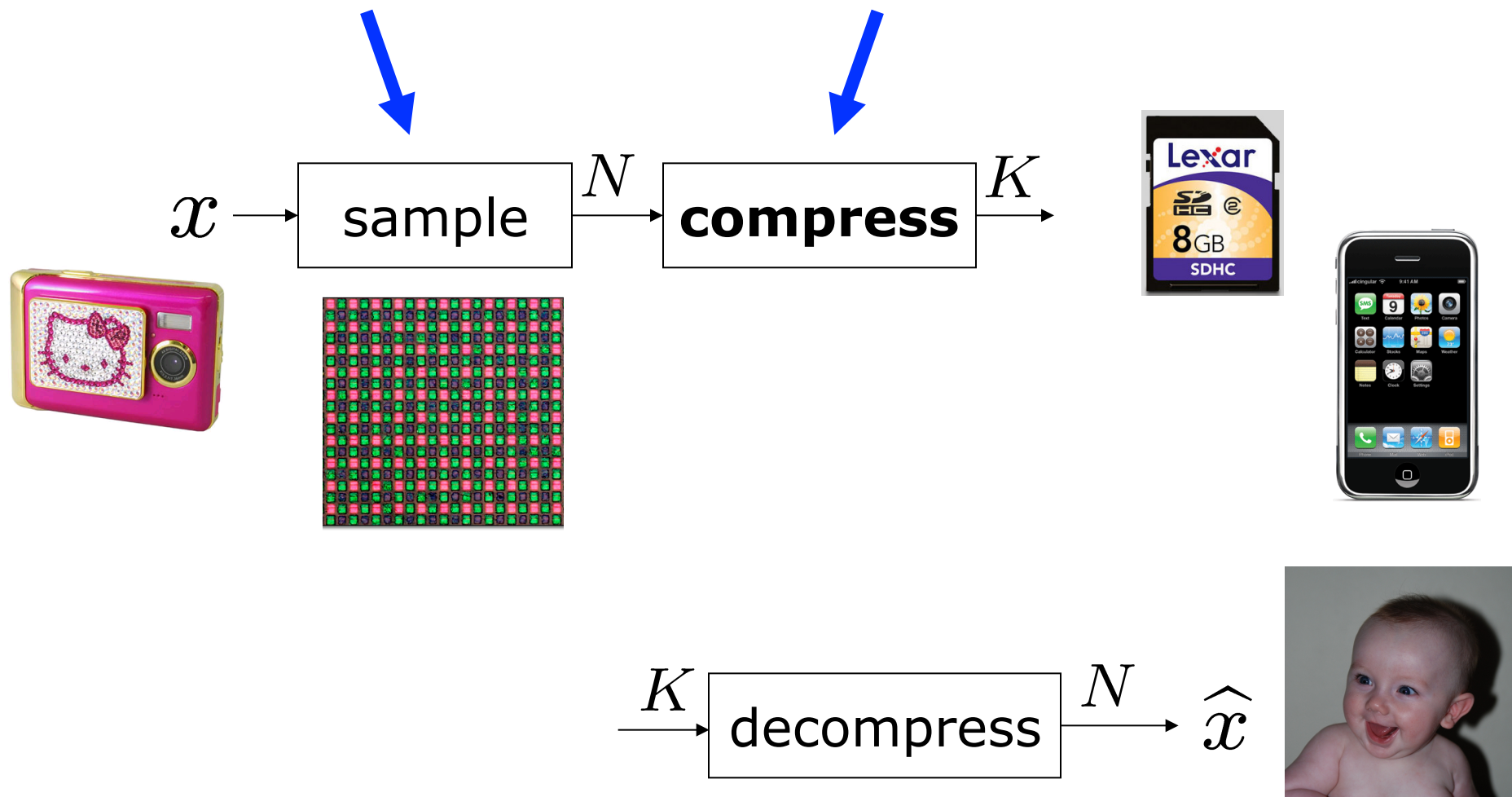


What's Wrong with this Picture?



linear processing
linear signal model
(bandlimited subspace)

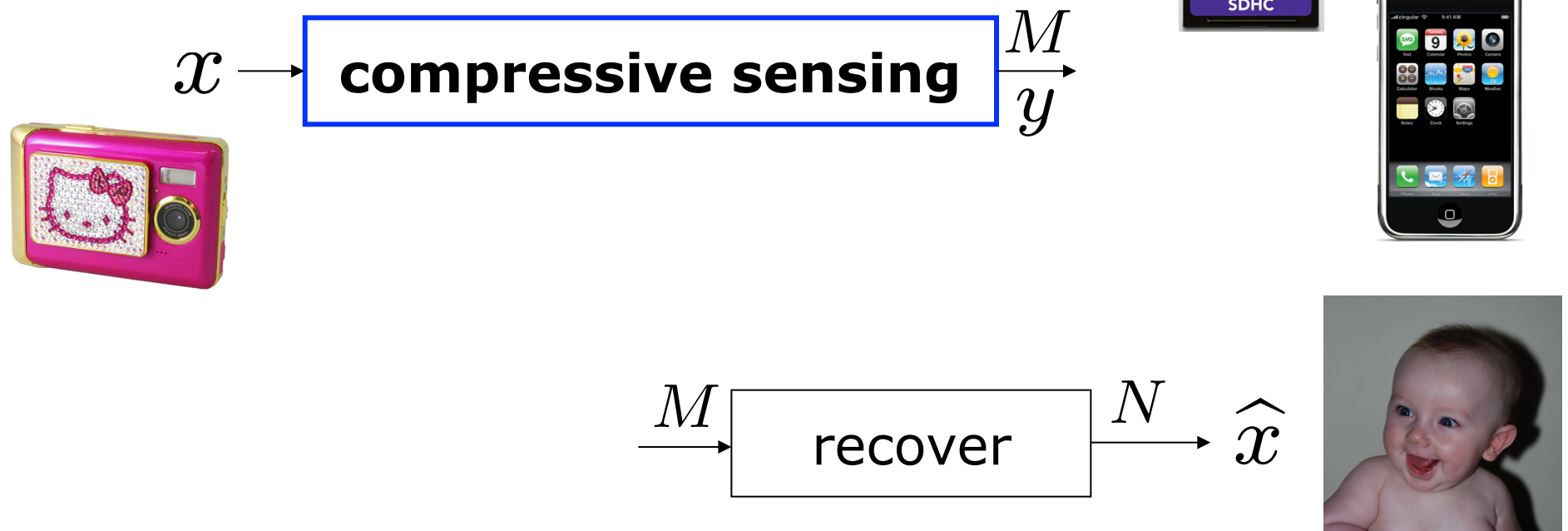
nonlinear processing
nonlinear signal model
(union of subspaces)



Compressive Sensing

- Directly acquire “**compressed**” data via dimensionality reduction
- Replace samples by more general “measurements”

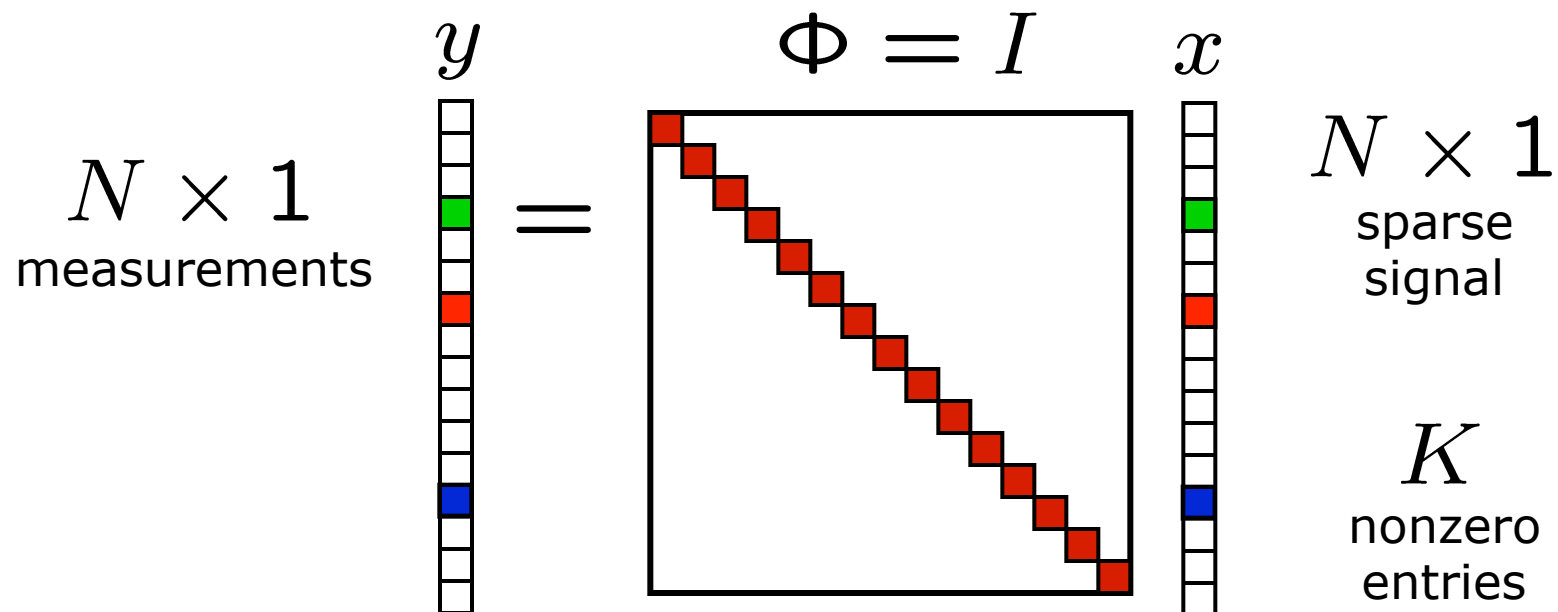
$$K \approx \underline{M} \ll N$$



Sampling

- Signal x is K -sparse in basis/dictionary Ψ
 - WLOG assume sparse in space domain $\Psi = I$

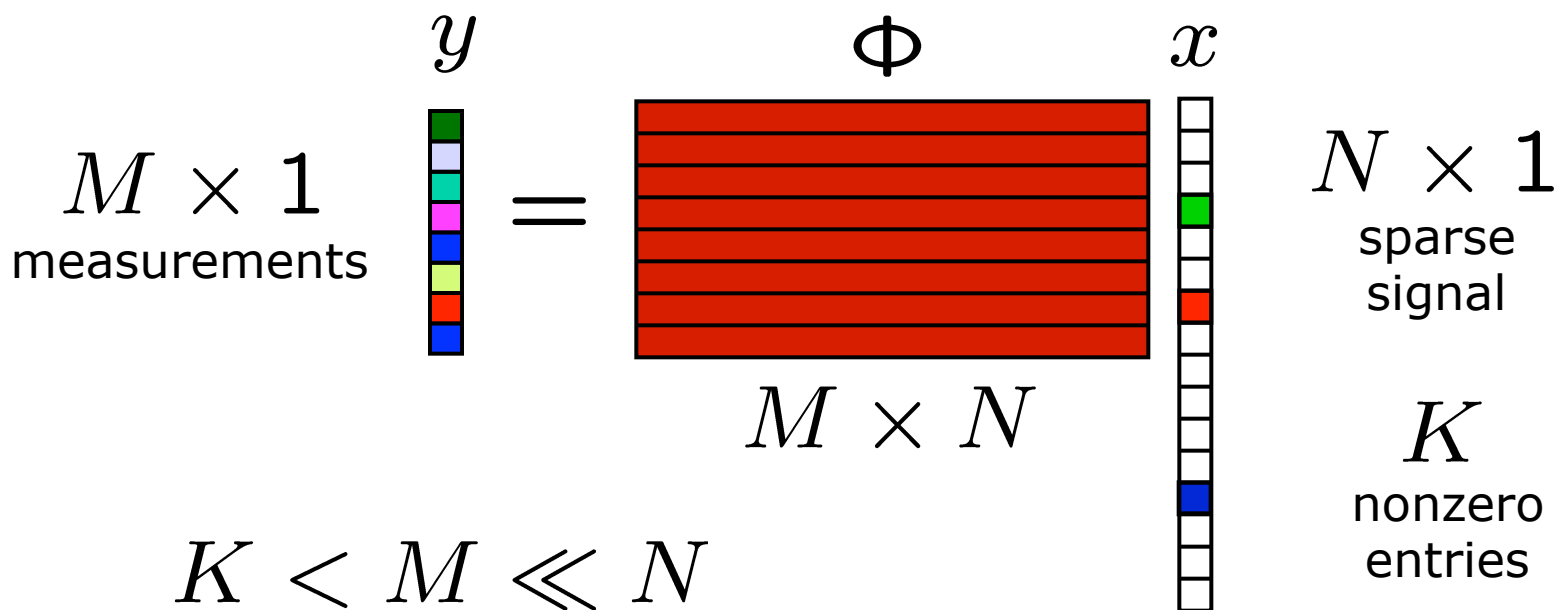
- **Sampling**



Compressive Sampling

- When data is sparse/compressible, can directly acquire a **condensed representation** with no/little information loss through linear **dimensionality reduction**

$$y = \Phi x$$



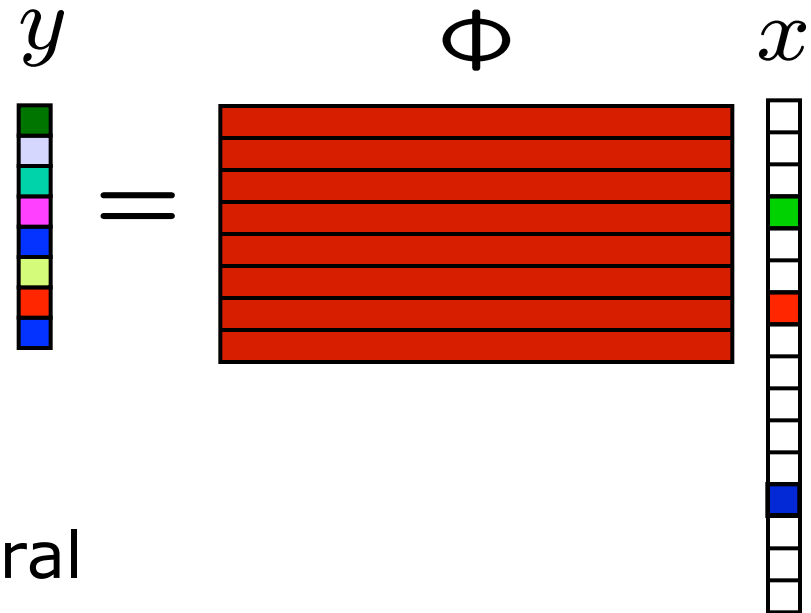
How Can It Work?

- Projection Φ
not full rank...

$$M < N$$

... and so

loses information in general



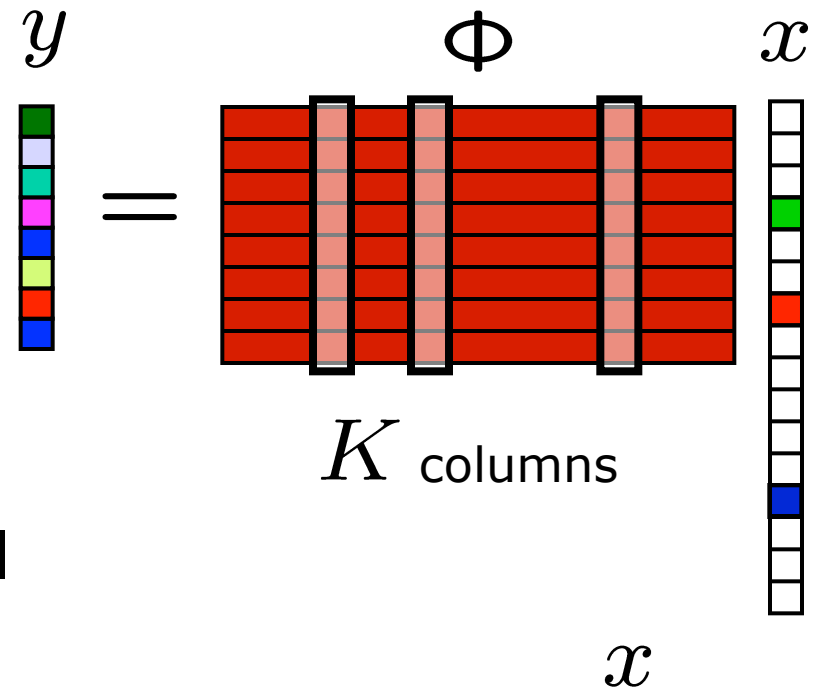
- Ex: Infinitely many x 's map to the same y
(null space)

How Can It Work?

- Projection Φ
not full rank...

$$M < N$$

... and so
loses information in general



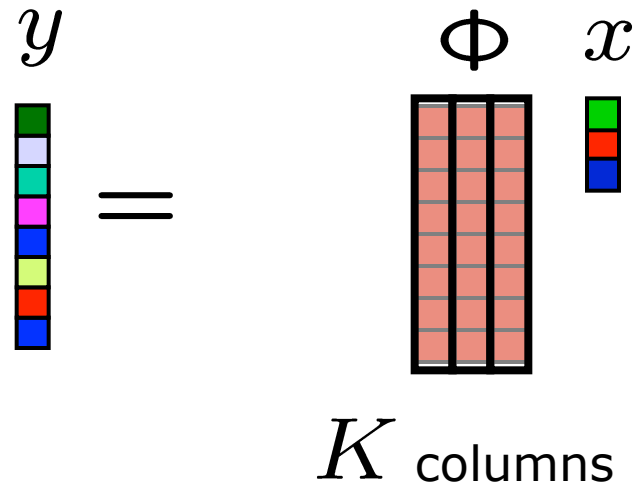
- But we are only interested in **sparse** vectors

How Can It Work?

- Projection Φ
not full rank...

$$M < N$$

... and so
loses information in general



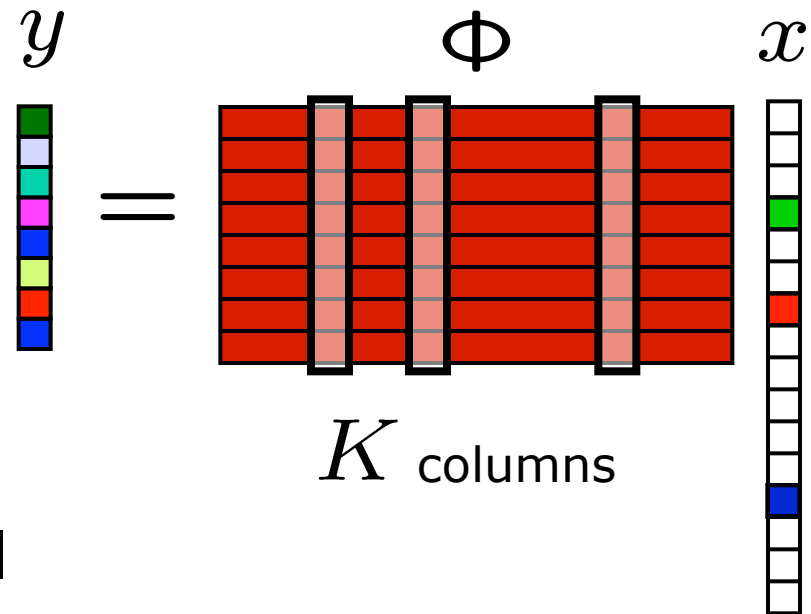
- But we are only interested in **sparse** vectors
- Φ is effectively $M \times K$

How Can It Work?

- Projection Φ
not full rank...

$$M < N$$

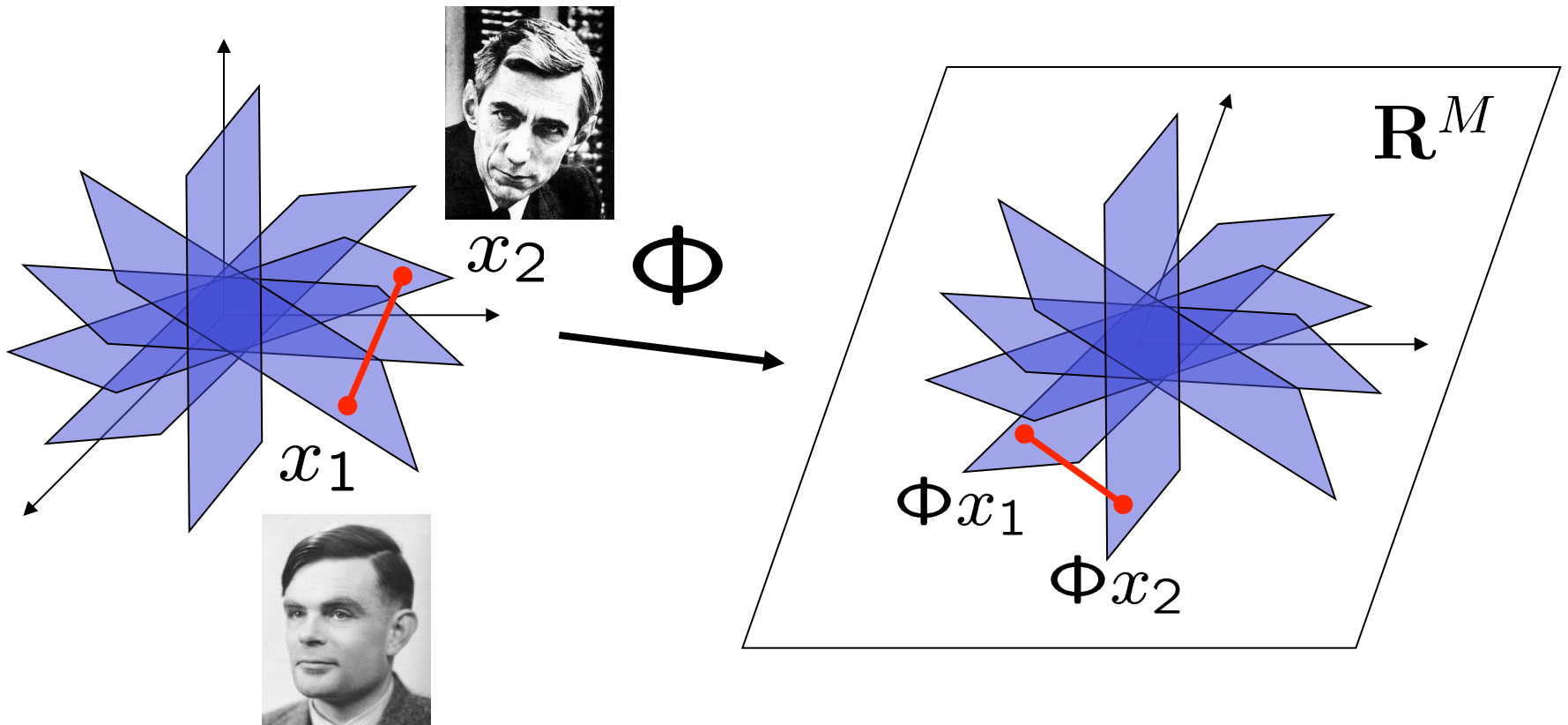
... and so
loses information in general



- But we are only interested in **sparse** vectors
- **Design** Φ so that each of its $M \times K$ submatrices are full rank (ideally close to orthobasis)
 - **Restricted Isometry Property (RIP)**

RIP = Stable Embedding

- An information preserving projection Φ preserves the **geometry** of the set of sparse signals



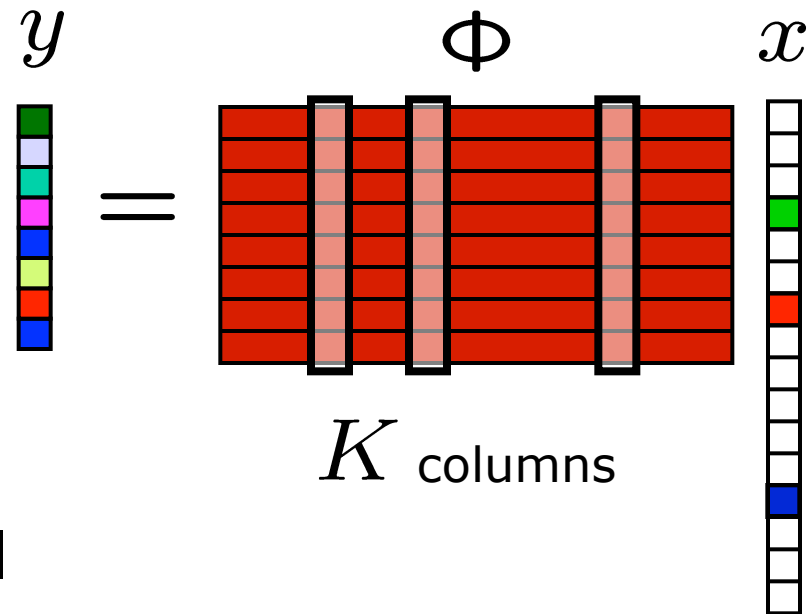
- RIP ensures that $\|x_1 - x_2\|_2 \approx \|\Phi x_1 - \Phi x_2\|_2$

How Can It Work?

- Projection Φ
not full rank...

$$M < N$$

... and so
loses information in general

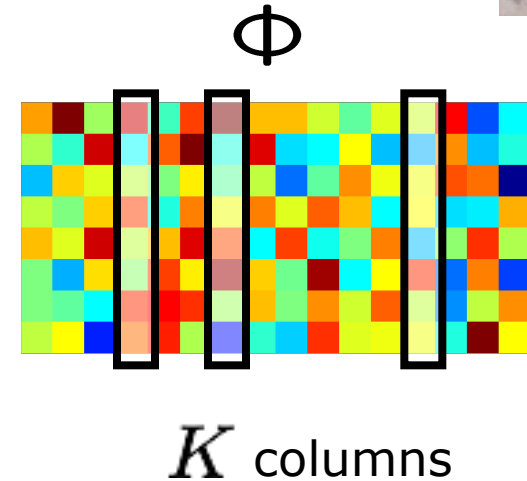


- **Design** Φ so that each of its $M \times K$ submatrices are full rank (RIP)
- Unfortunately, a combinatorial,
NP-Hard design problem

Insight from the 70's [Kashin, Gluskin]



- Draw Φ at **random**
 - iid Gaussian
 - iid Bernoulli ± 1
 - ...

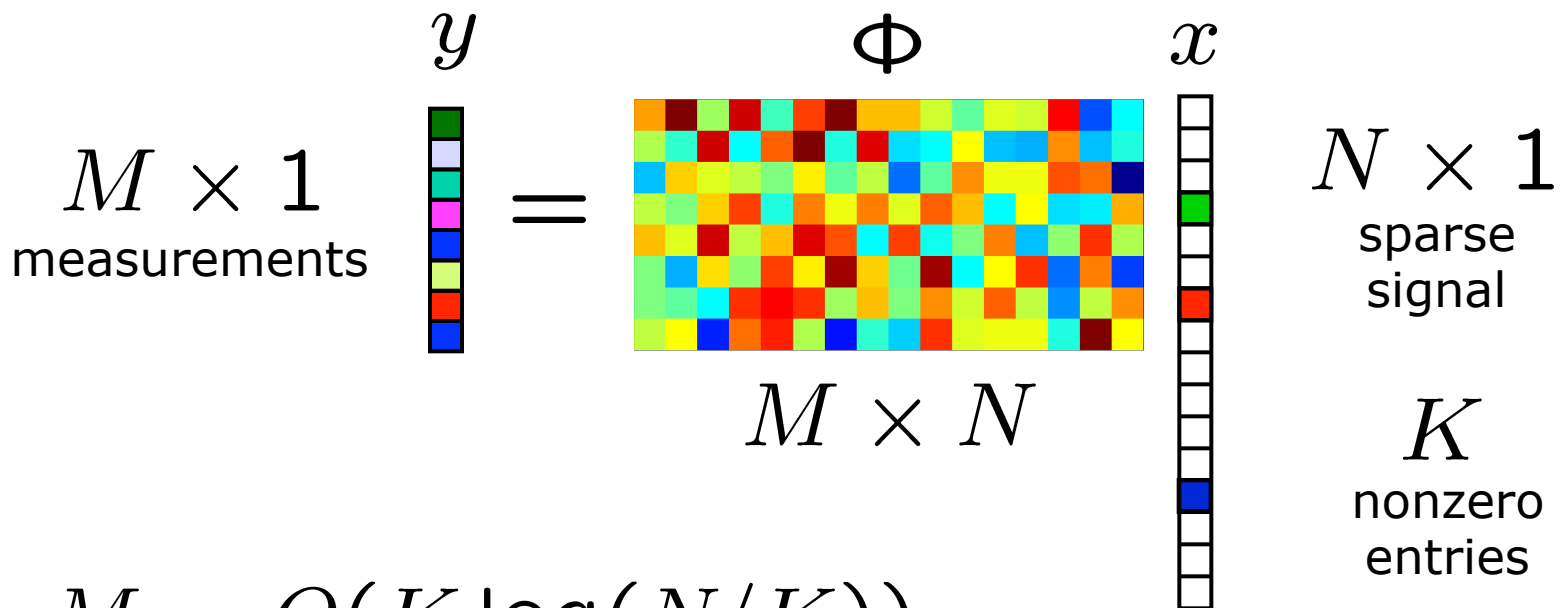


- Then Φ has the RIP with high probability provided

$$M = O(K \log(N/K)) \ll N$$

Randomized Sensing

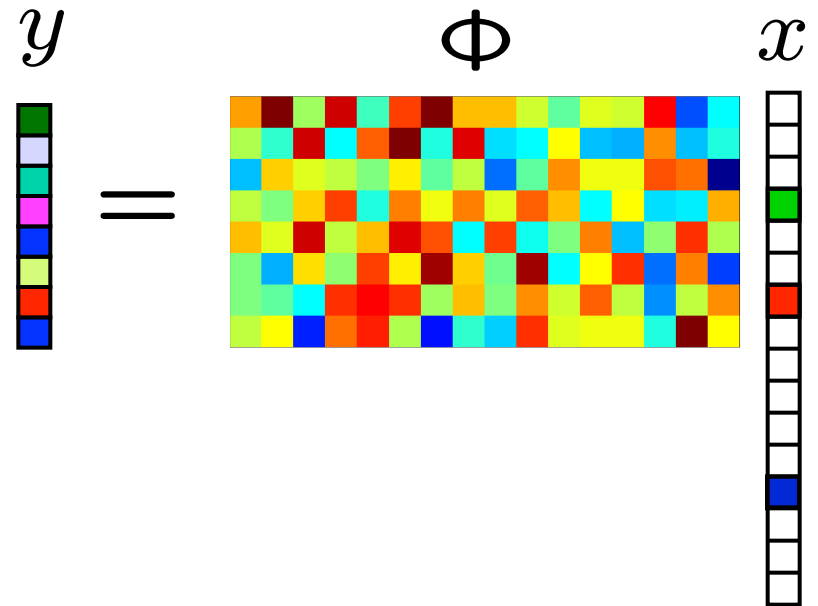
- Measurements $y =$ **random linear combinations** of the entries of x
- **No information loss** for sparse vectors x whp



$$M = O(K \log(N/K))$$

CS Signal *Recovery*

- **Goal:** Recover signal x from measurements y



- **Problem:** Random projection Φ not full rank (ill-posed inverse problem)
- **Solution:** Exploit the sparse/compressible ***geometry*** of acquired signal x

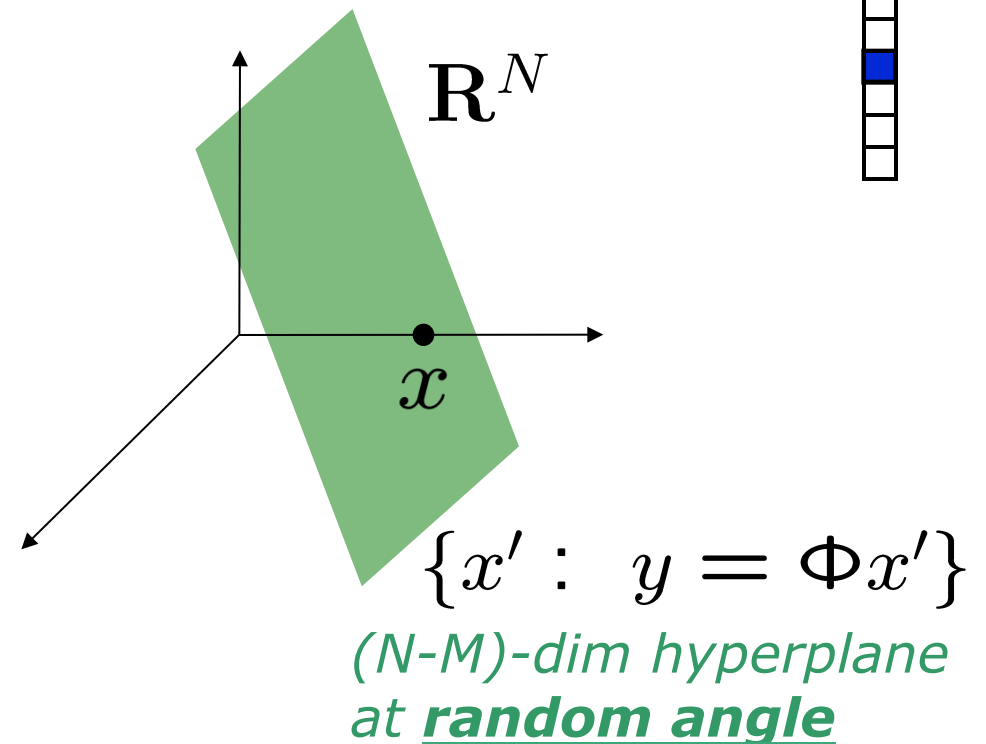
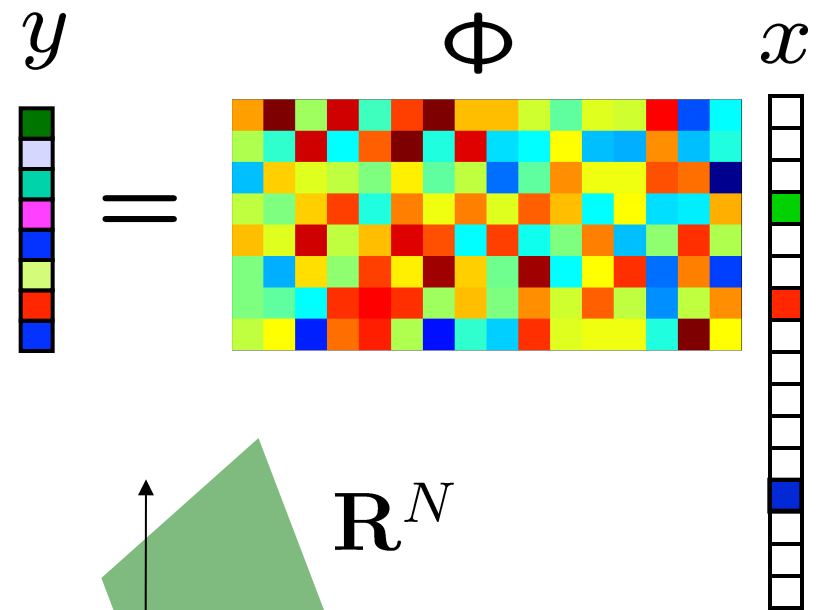
CS Signal Recovery

- Random projection Φ
not full rank

- Recovery problem:
given $y = \Phi x$
find x

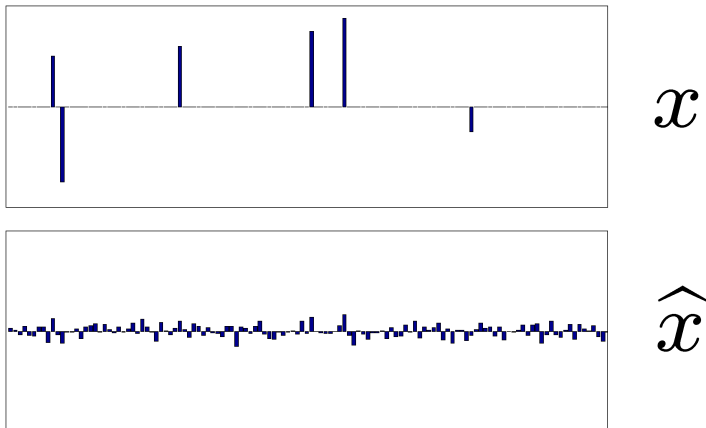
- **Null space**

- Search in null space
for the “best” x
according to some
criterion
 - ex: least squares



ℓ_2 Signal Recovery

- Recovery: (ill-posed inverse problem) given $y = \Phi x$
find x (sparse)
- Optimization: $\hat{x} = \arg \min_{y=\Phi x} \|x\|_2$
- Closed-form solution: $\hat{x} = (\Phi^T \Phi)^{-1} \Phi^T y$
- **Wrong answer!**



ℓ_2 Signal Recovery

- Recovery:
(ill-posed inverse problem)

$$\begin{array}{ll} \text{given} & y = \Phi x \\ \text{find} & x \text{ (sparse)} \end{array}$$

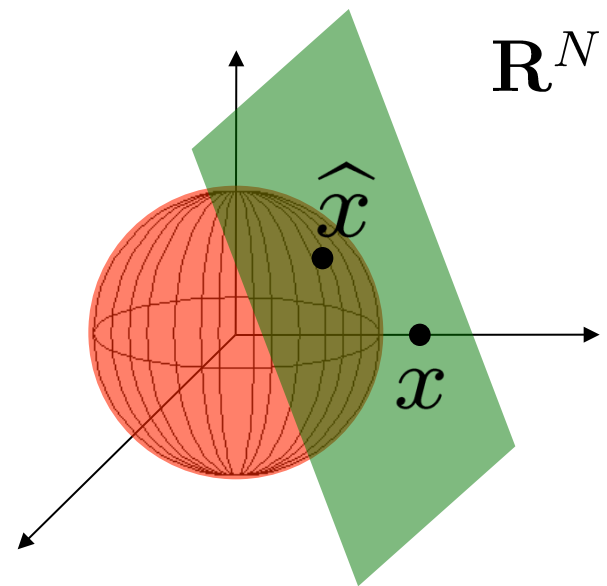
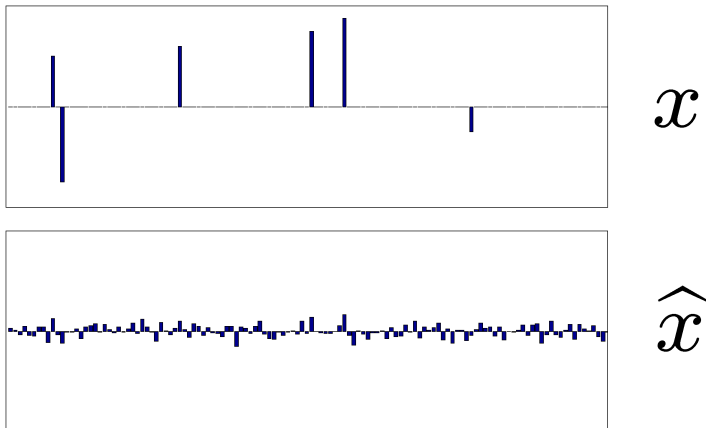
- Optimization:

$$\hat{x} = \arg \min_{y=\Phi x} \|x\|_2$$

- Closed-form solution:

$$\hat{x} = (\Phi^T \Phi)^{-1} \Phi^T y$$

- **Wrong answer!**



ℓ_0 Signal Recovery

- Recovery:
(ill-posed inverse problem)

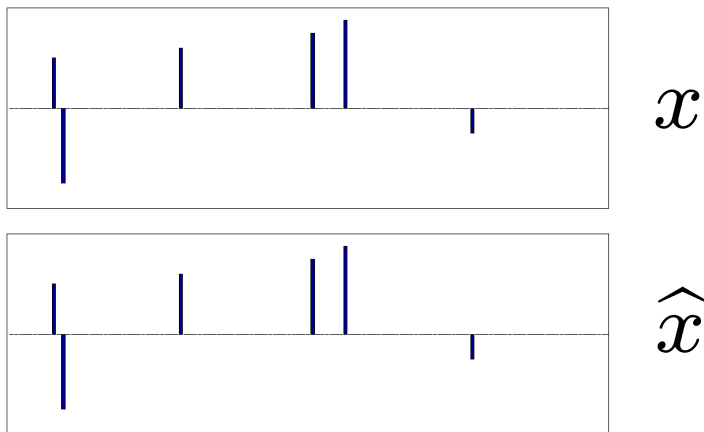
$$\begin{array}{ll} \text{given} & y = \Phi x \\ \text{find} & x \text{ (sparse)} \end{array}$$

- Optimization:

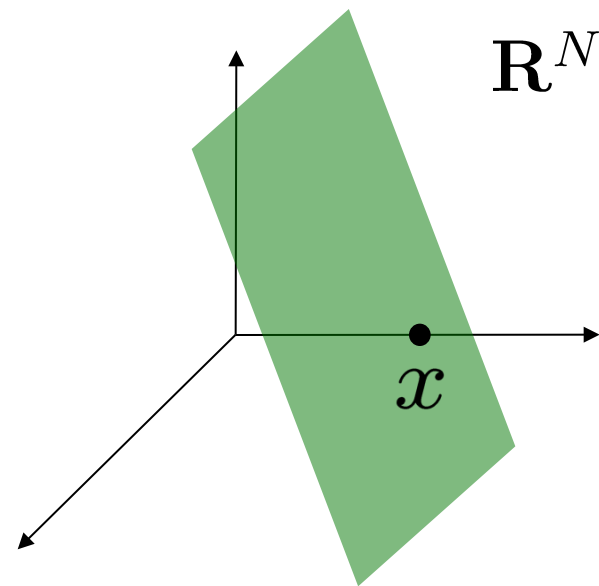
$$\hat{x} = \arg \min_{y=\Phi x} \|x\|_0$$

- **Correct!**


“find *sparsest* vector
in translated nullspace”



- But **NP-Complete** alg



ℓ_1 Signal Recovery

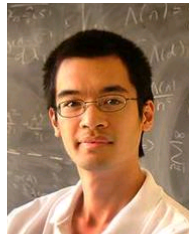
- Recovery: (ill-posed inverse problem) given $y = \Phi x$
find x (sparse)
- Optimization: $\hat{x} = \arg \min_{y=\Phi x} \|x\|_1$ 
- **Convexify** the ℓ_0 optimization



Candes



Romberg



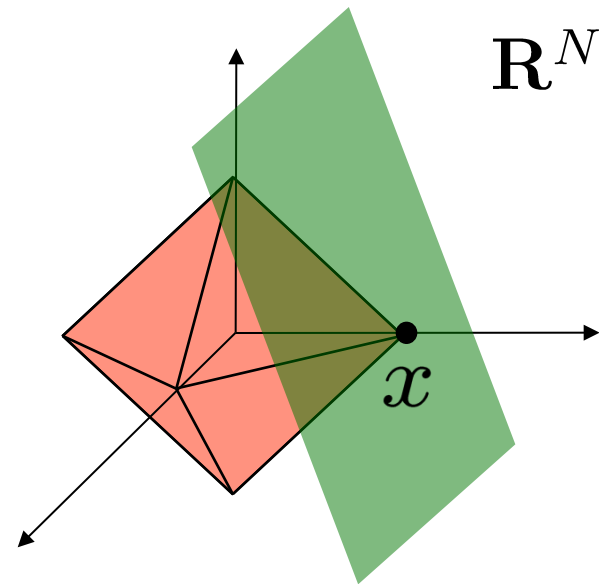
Tao



Donoho

ℓ_1 Signal Recovery

- Recovery: (ill-posed inverse problem) given $y = \Phi x$
find x (sparse)
- Optimization: $\hat{x} = \arg \min_{y=\Phi x} \|x\|_1$
- **Convexify** the ℓ_0 optimization
- **Correct!**
- **Polynomial time** alg (linear programming)



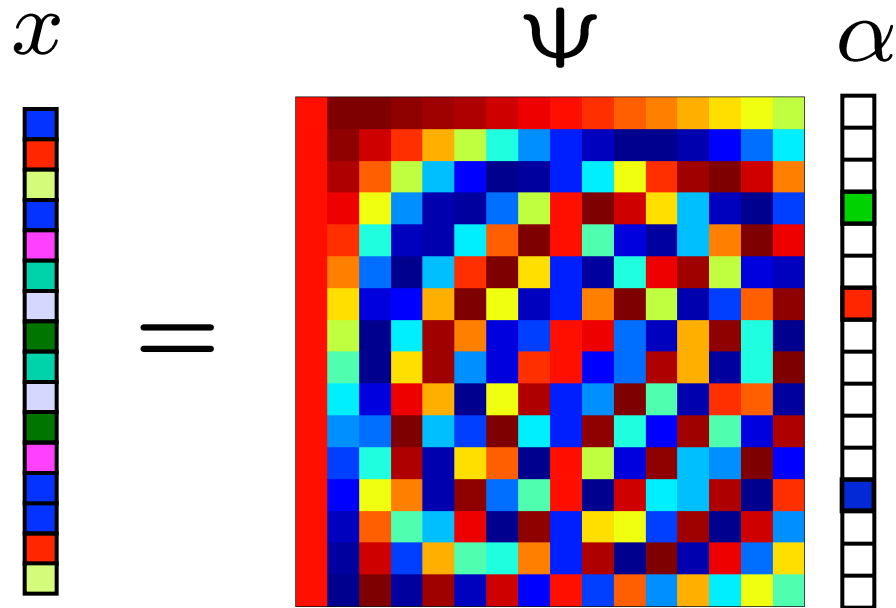
CS Hallmarks

- **Stable**
 - acquisition/recovery process is numerically stable
- **Asymmetrical** (most processing at decoder)
 - conventional: smart encoder, dumb decoder
 - CS: dumb encoder, smart decoder
- **Democratic**
 - each measurement carries the same amount of information
 - robust to measurement loss and quantization
 - “digital fountain” property
- Random measurements **encrypted**
- **Universal**
 - same random projections / hardware can be used for *any* sparse signal class (*generic*)

Universality

- Random measurements can be used for signals sparse in *any* basis

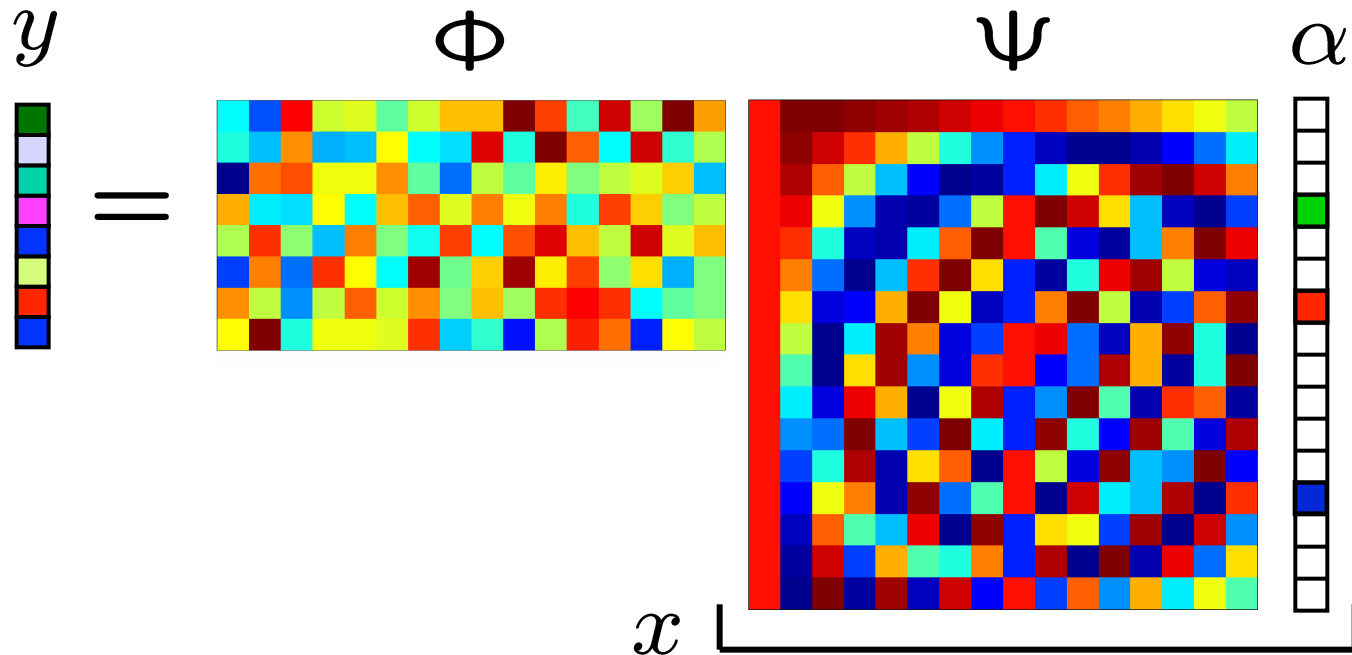
$$x = \Psi \alpha$$



Universality

- Random measurements can be used for signals sparse in *any* basis

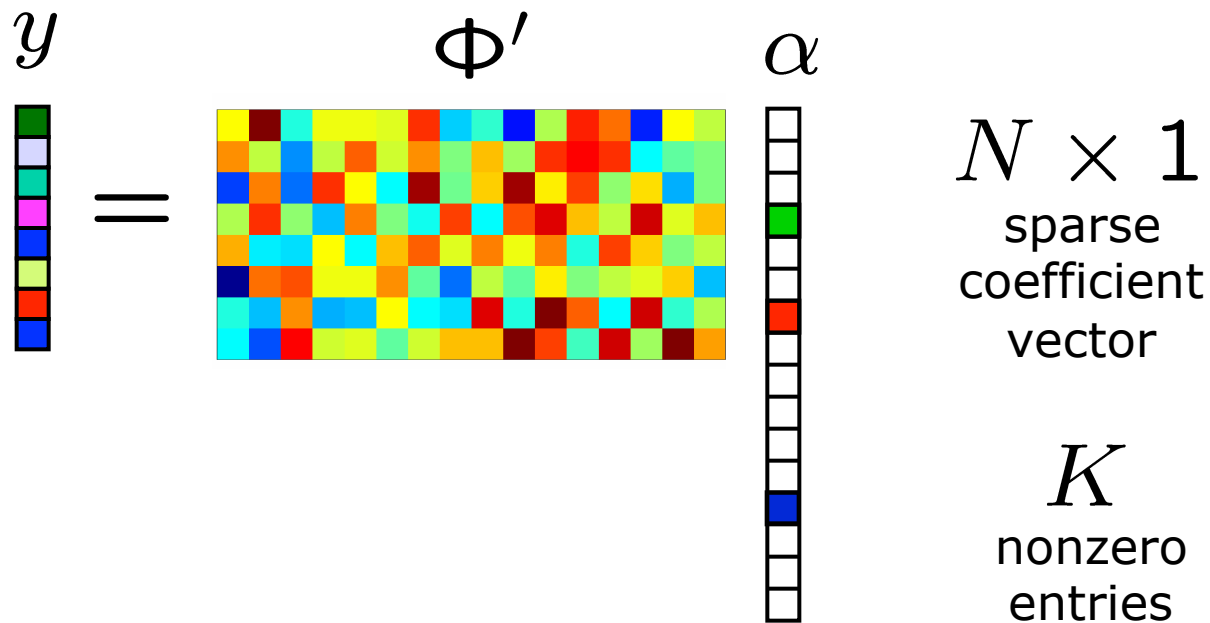
$$y = \Phi x = \Phi \Psi \alpha$$



Universality

- Random measurements can be used for signals sparse in *any* basis

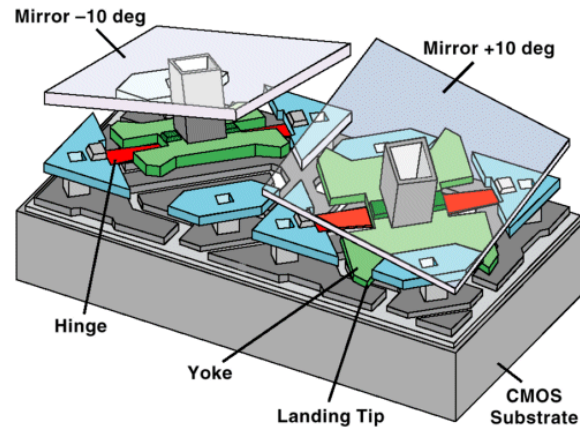
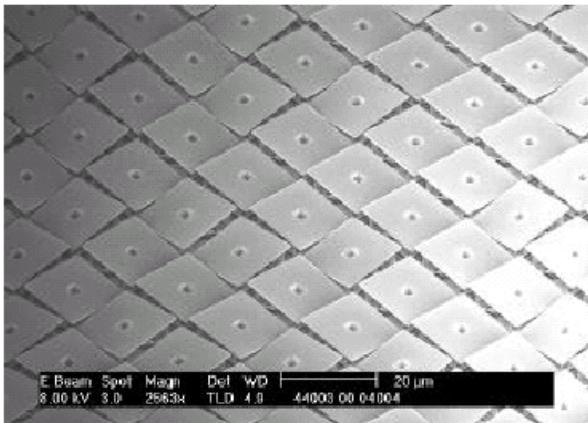
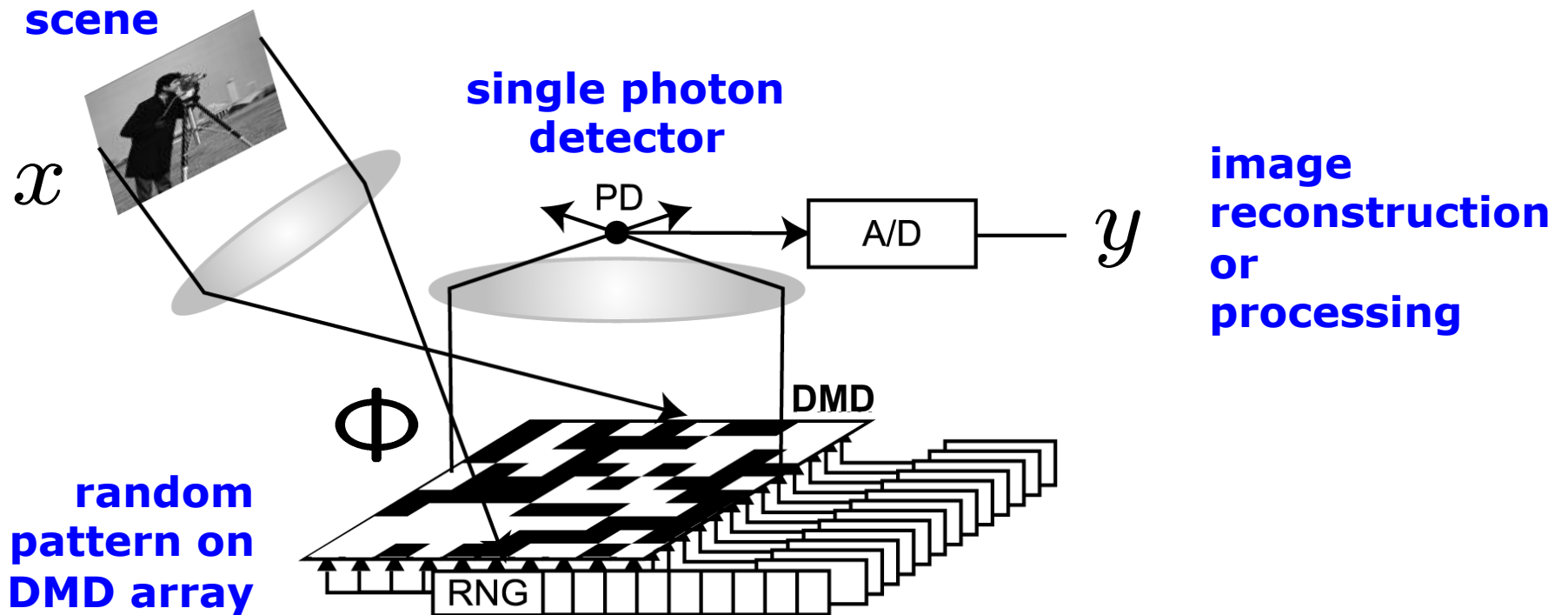
$$y = \Phi x = \Phi \Psi \alpha = \Phi' \alpha$$



Compressive Sensing ***In Action***

Cameras

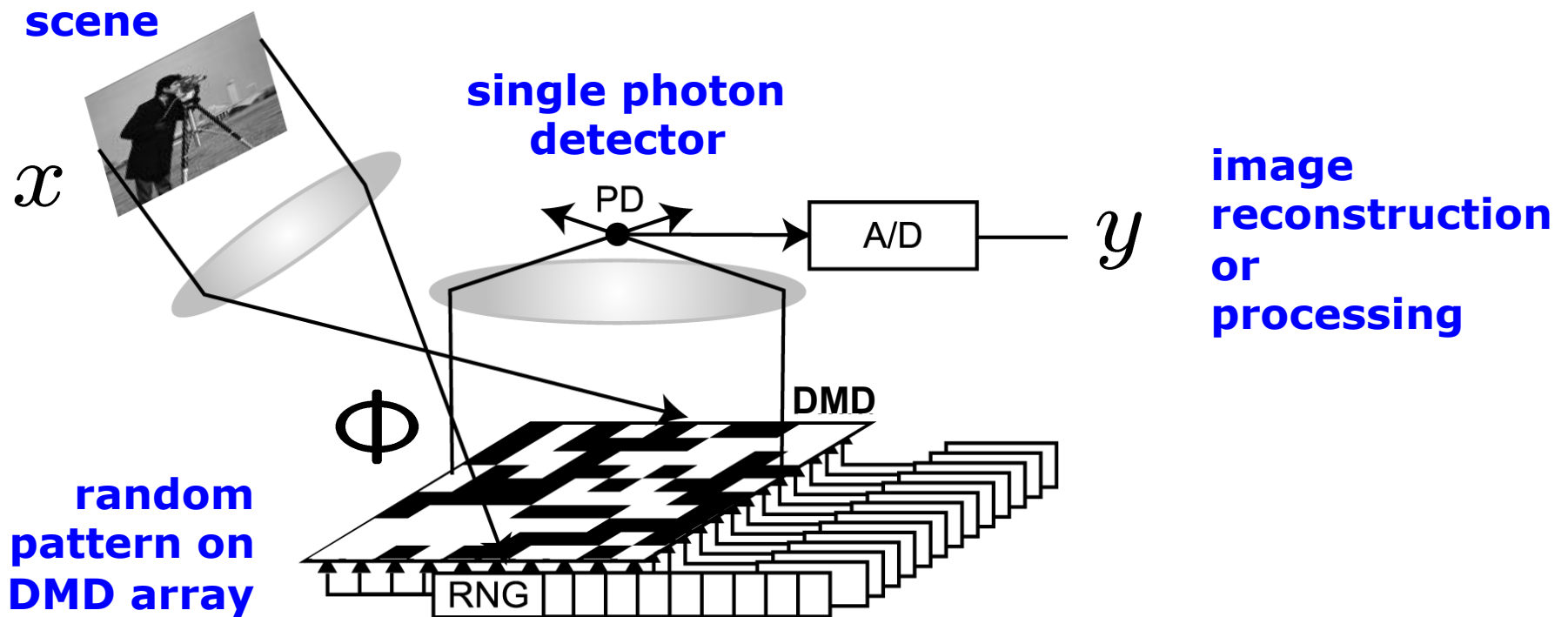
“Single-Pixel” CS Camera



w/ Kevin Kelly

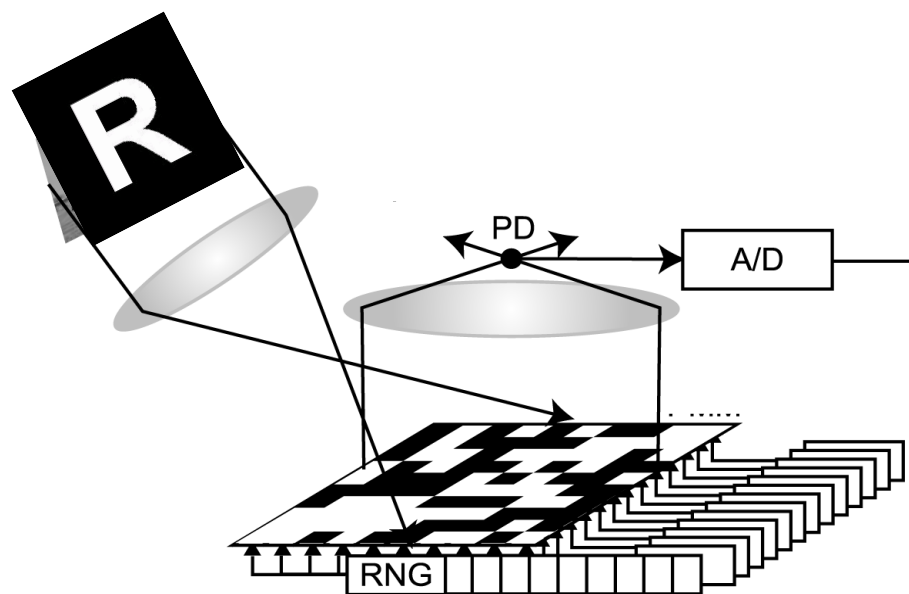


“Single-Pixel” CS Camera



- Flip mirror array M times to acquire M measurements
- Sparsity-based (linear programming) recovery

First Image Acquisition



target
65536 pixels



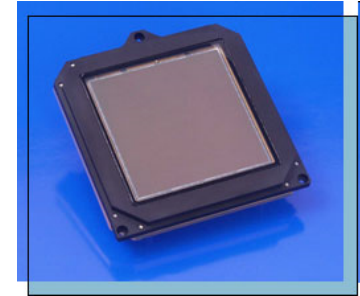
11000 measurements
(16%)



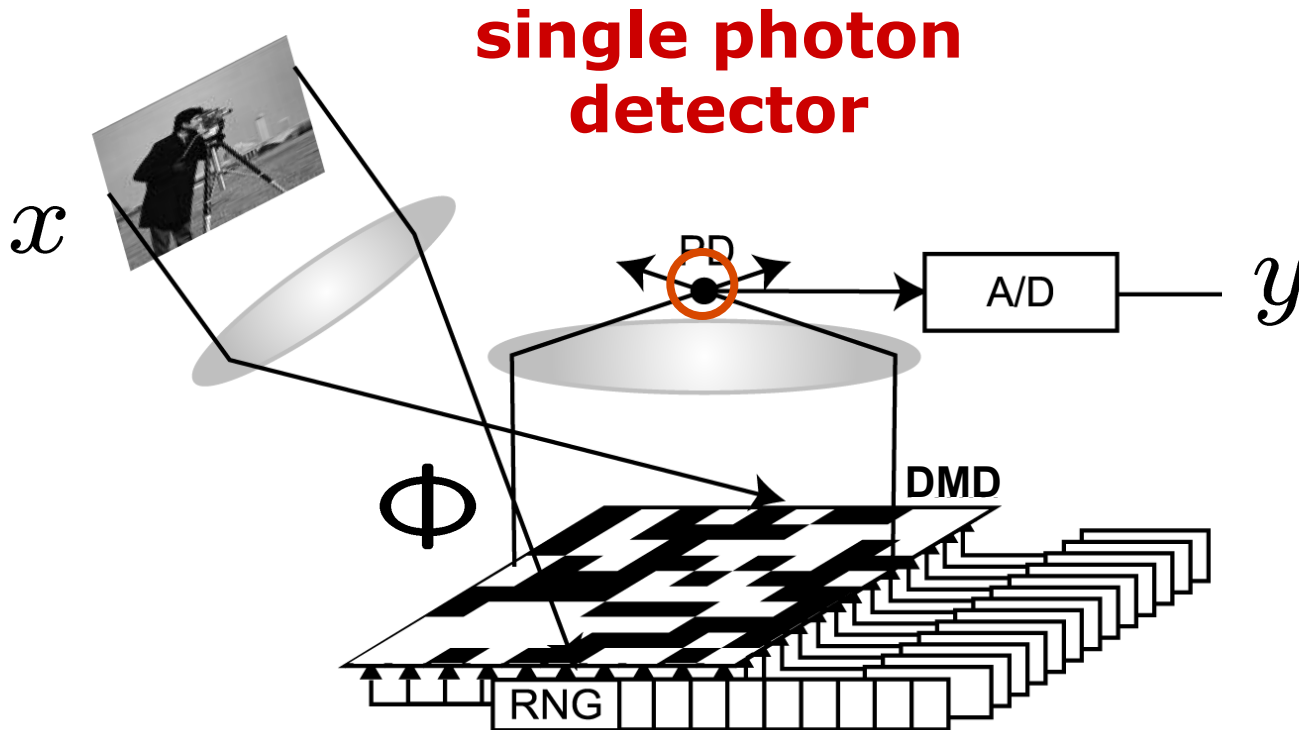
1300 measurements
(2%)



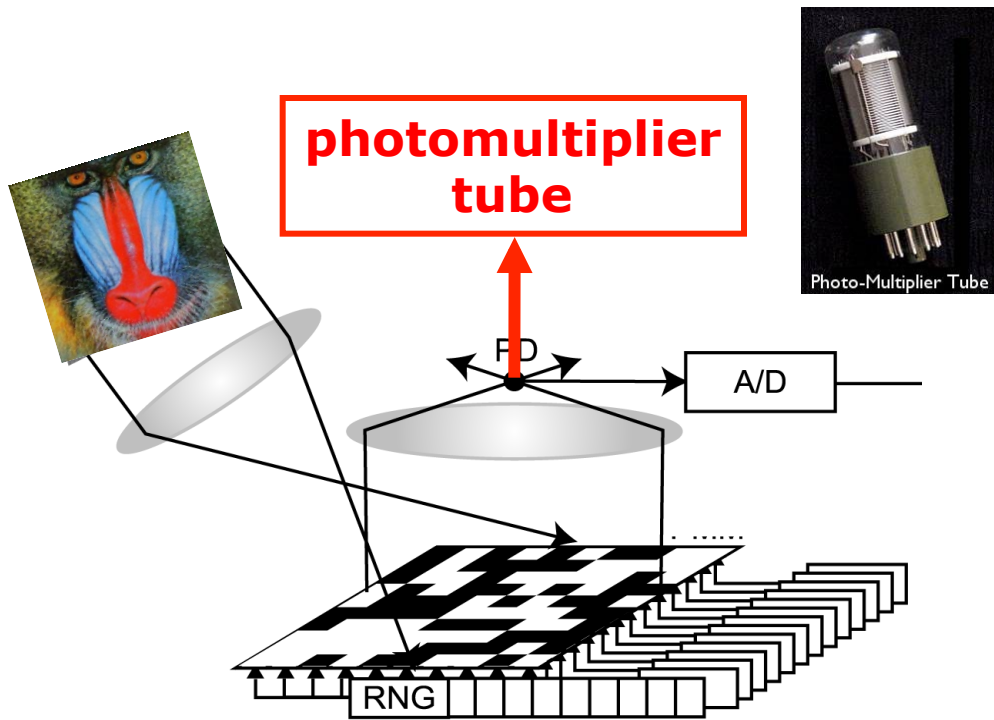
Utility?



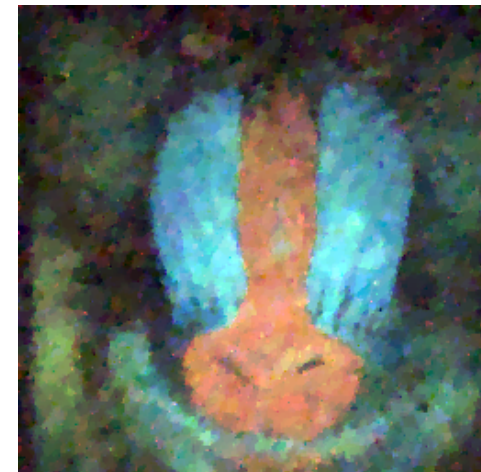
Fairchild
100Mpixel
CCD



CS Low Light Imager



target



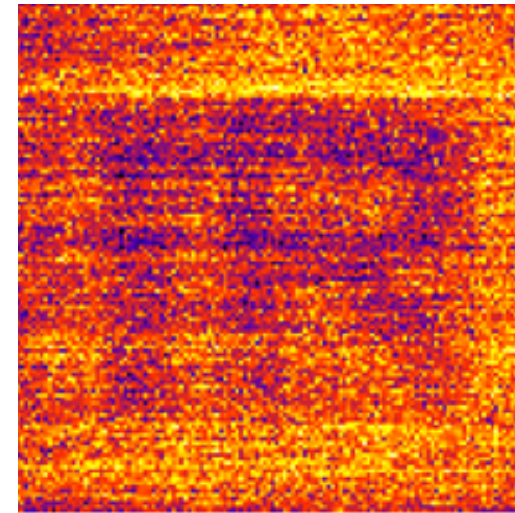
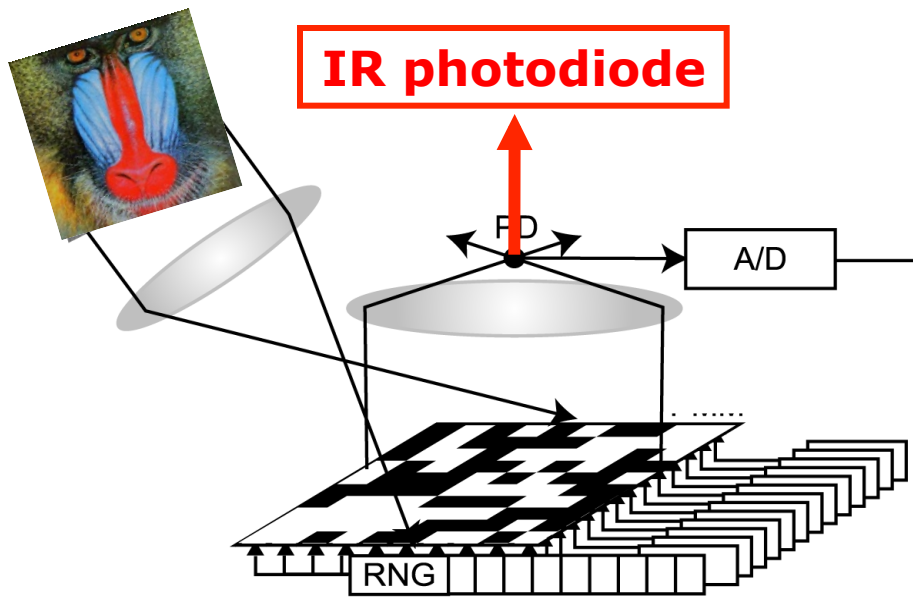
low light image

true color low-light imaging

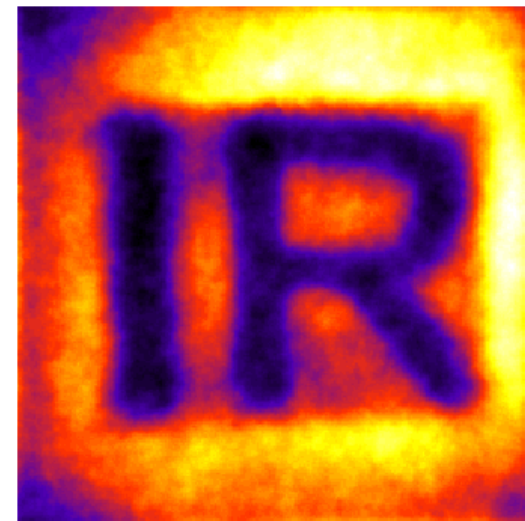
256 x 256 image with 10:1
compression

[Nature Photonics, April 2007]

CS Infrared Imager

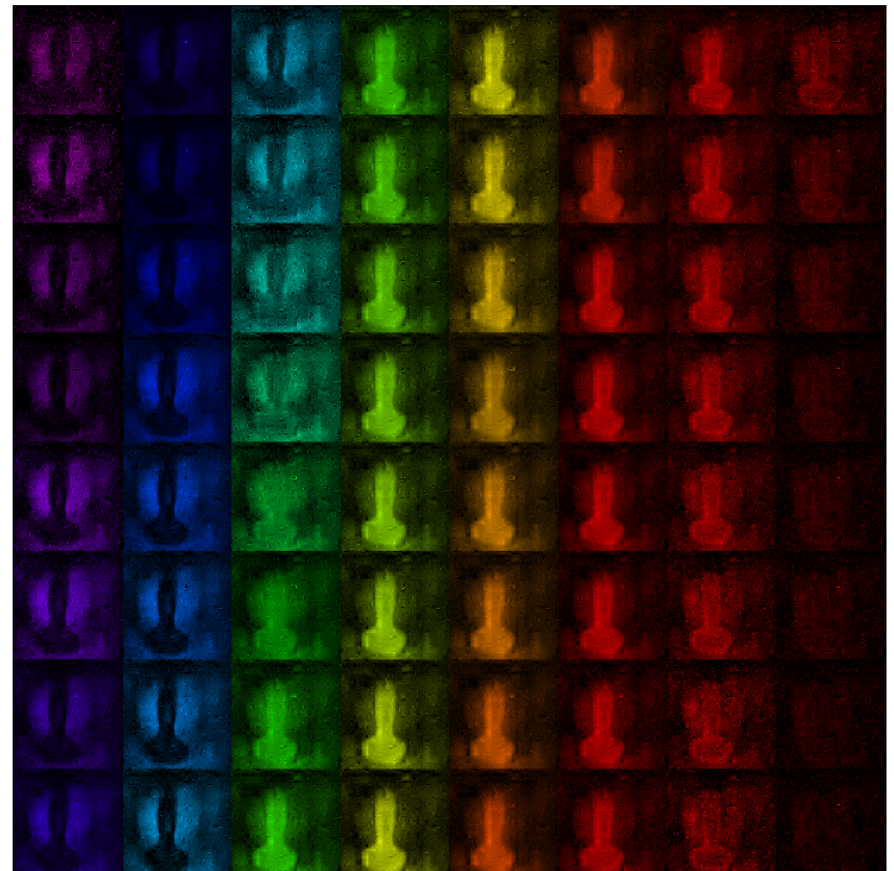
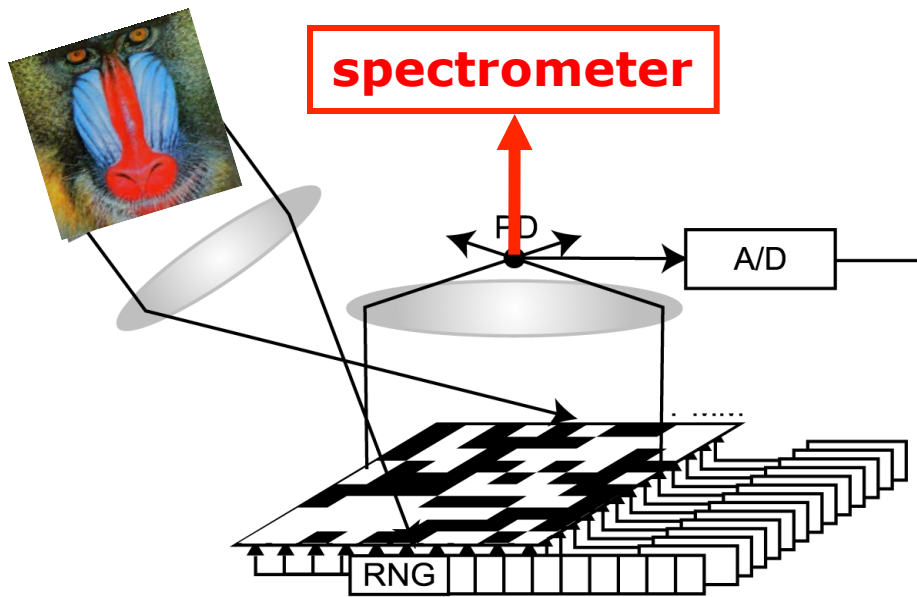


raster scan IR



CS IR

CS Hyperspectral Imager



hyperspectral data cube

450-850nm

$N=1\text{M}$ space x wavelength voxels

$M=200\text{k}$ random measurements

Compressive Sensing ***In Action***

Video Acquisition

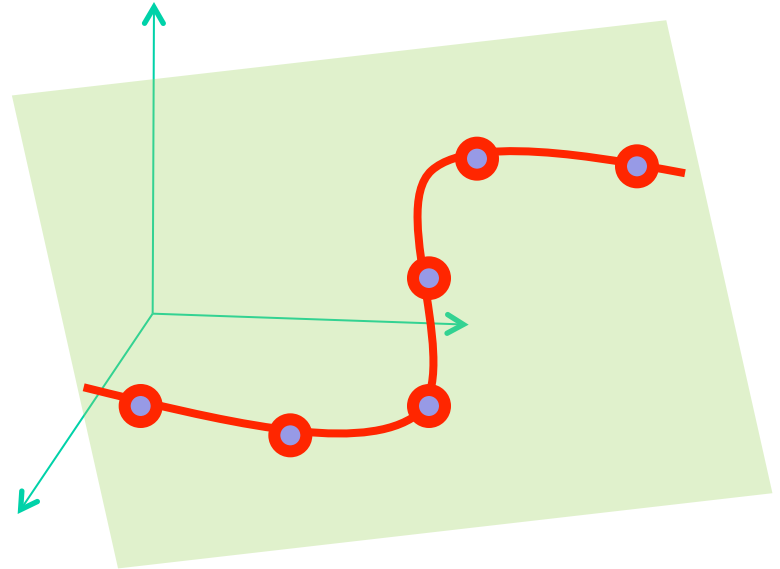
From Image to Video Sensing



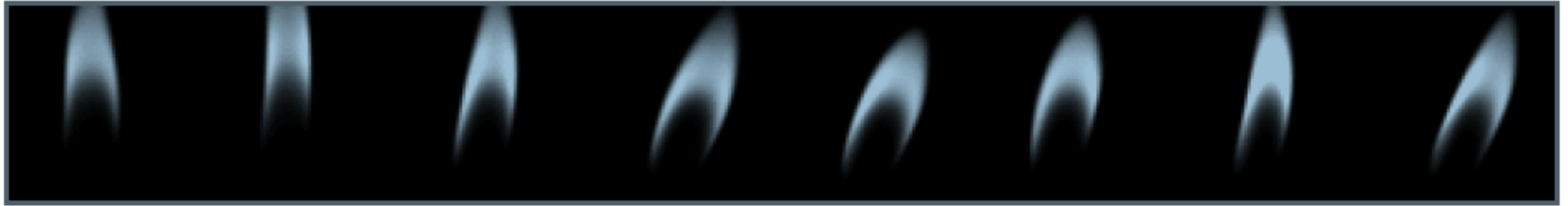
- Nontrivial extension of CS image acquisition
 - immoral to treat time as 3rd spatial dimension
- **Ephemeral** temporal events
 - should measure temporal events at their “information rate”
 - fleeting events hard to predict and capture
- Computational **complexity** involved in recovering billions of video voxels

Simple LDS Model

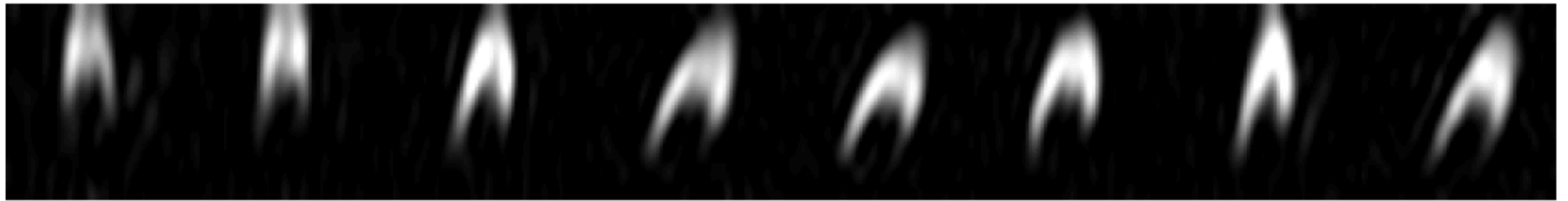
- **Linear dynamical system** model
 - image sequence lies along a curve on a linear subspace
- Reasonable model for certain physical phenomena
 - flows, waves, ...
- Leverage modern **state space techniques** to estimate image sequence from compressive measurements



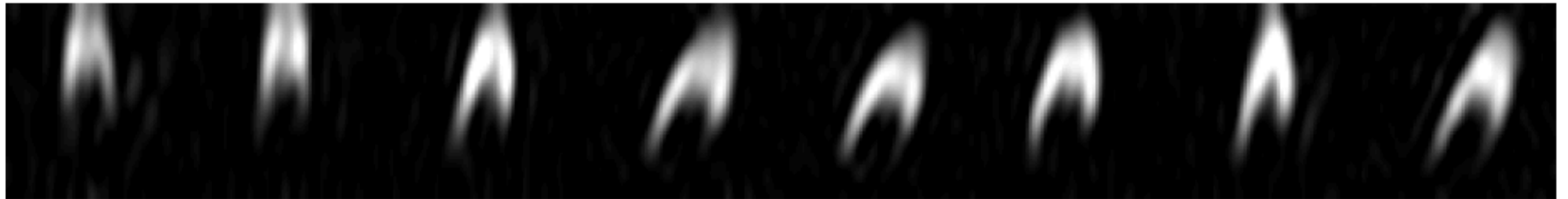
Flame Video



(a) Ground truth



(b) $f_s = 256$ Hz, $\tilde{M} = 30$, $\tilde{M} = 170$, Meas. rate = 5%, SNR = 13.73 dB.



(c) $f_s = 512$ Hz, $\tilde{M} = 30$, $\tilde{M} = 70$, Meas. rate = 2.44%, SNR = 13.73 dB.



(d) $f_s = 1024$ Hz, $\tilde{M} = 30$, $\tilde{M} = 20$, Meas. rate = 1.22%, SNR = 12.63 dB.

Traffic Video

ground truth



CS video recovery

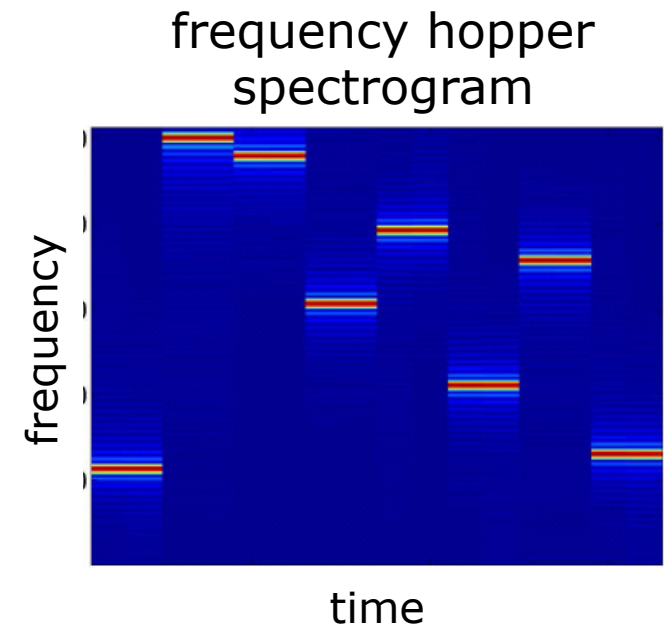
measurement rate = 4%

Compressive Sensing ***In Action***

A/D Converters

Analog-to-Digital Conversion

- Nyquist rate limits reach of today's ADCs
- “Moore's Law” for ADCs:
 - technology Figure of Merit incorporating sampling rate and dynamic range doubles every **6-8 years**
- Analog-to-Information (A2I) converter
 - wideband signals have high Nyquist rate but are often sparse/compressible
 - develop new ADC technologies to exploit
 - new tradeoffs among Nyquist rate, sampling rate, dynamic range, ...

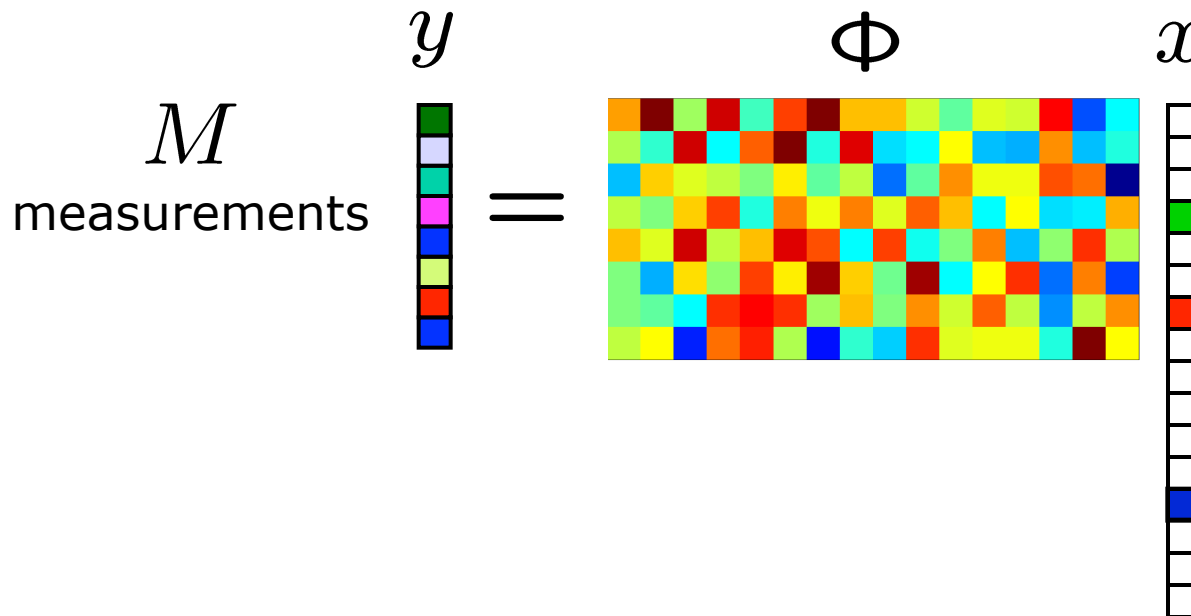


Streaming Measurements

- Streaming applications: cannot fit entire signal into a processing buffer at one time

$$y = \Phi x$$

streaming requires special Φ

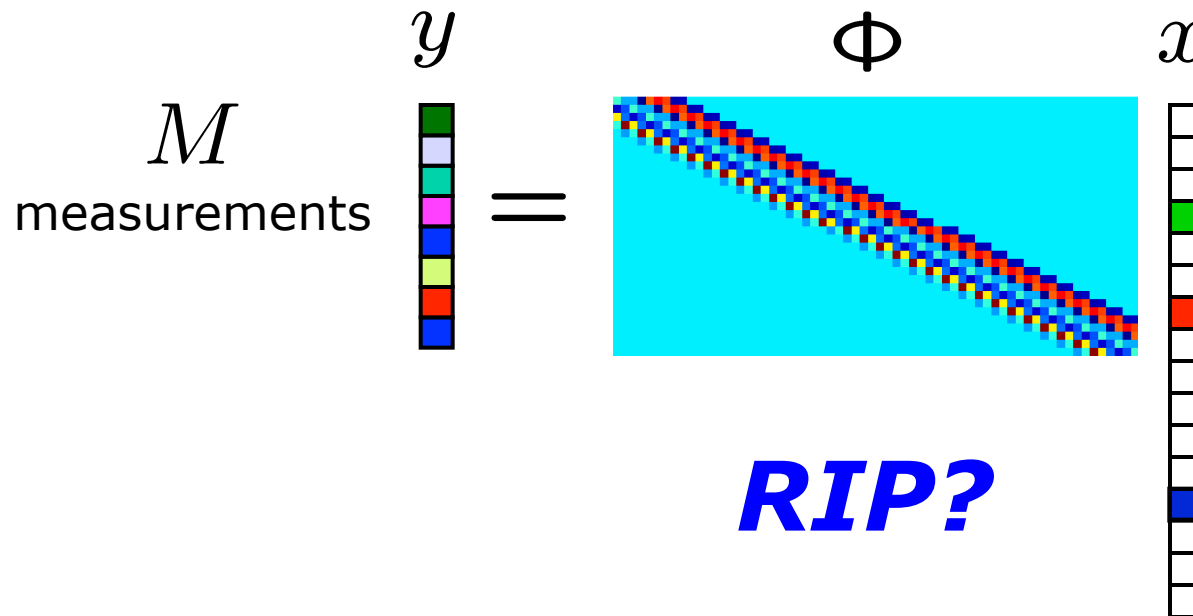


Streaming Measurements

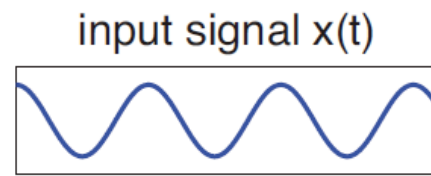
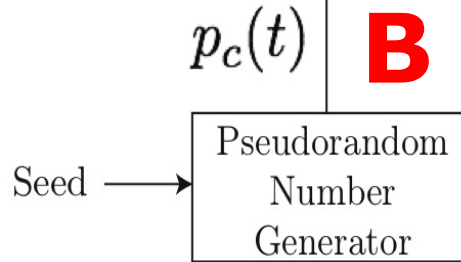
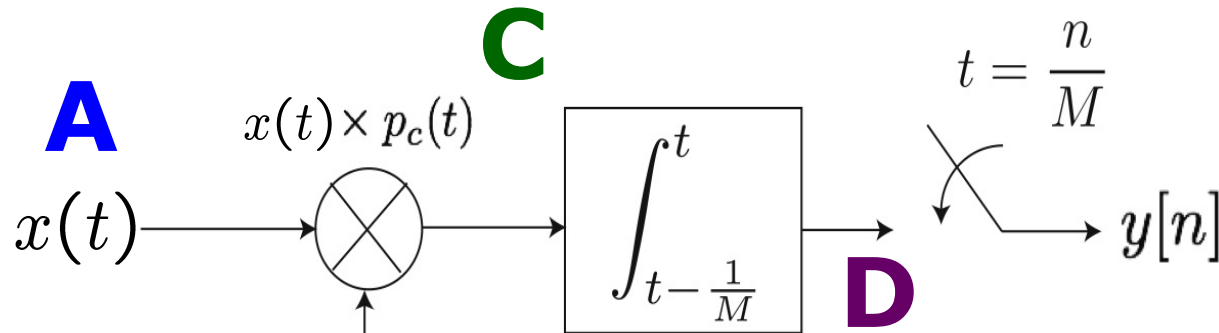
- Streaming applications: cannot fit entire signal into a processing buffer at one time

$$y = \Phi x$$

streaming requires special Φ



Random Demodulator

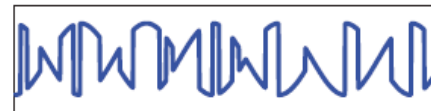


\times
pseudorandom
sequence $p_c(t)$



=

modulated input

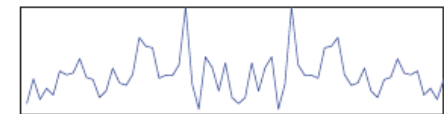


A

input signal $X(\omega)$

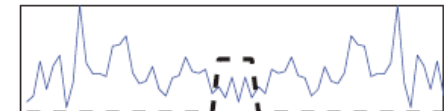


$*$
pseudorandom sequence
spectrum $P_c(\omega)$



=

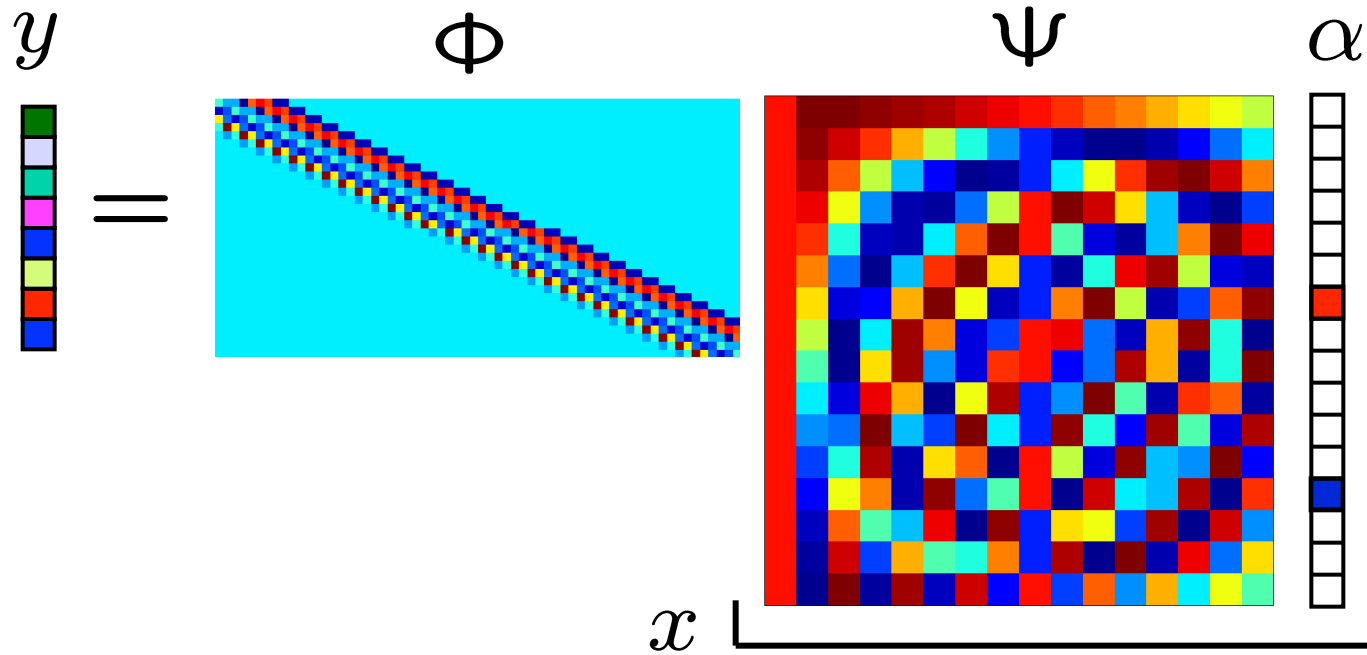
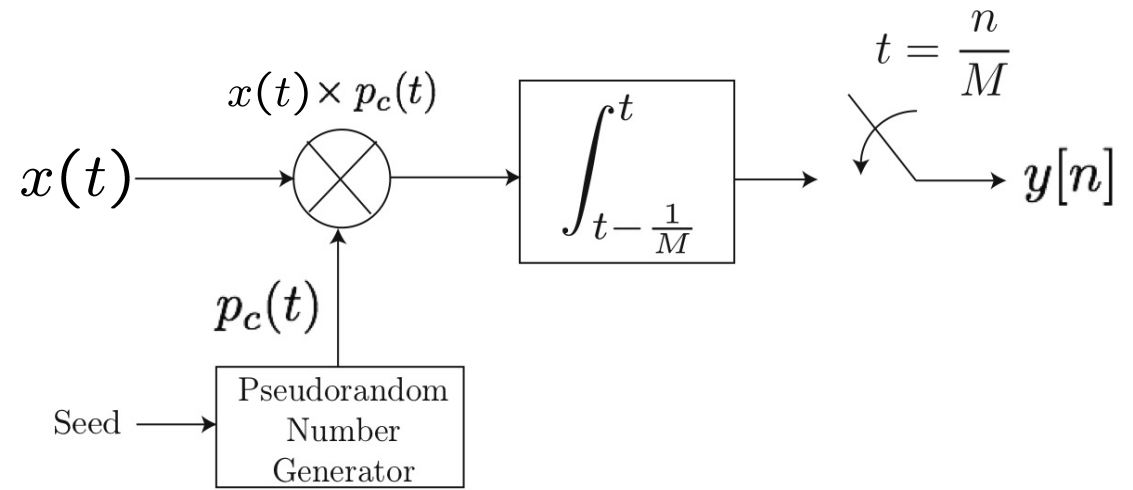
modulated input and
integrator (low-pass filter)



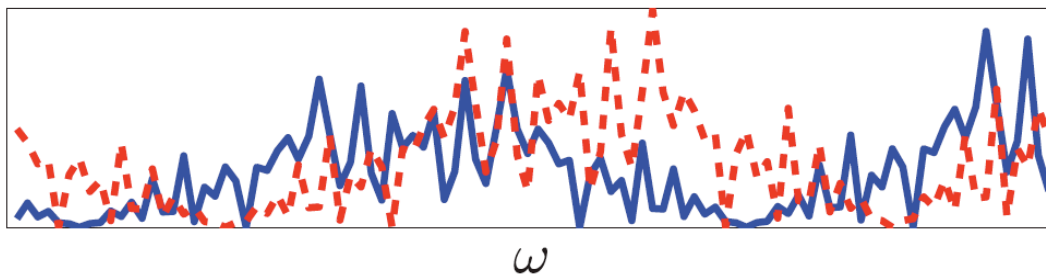
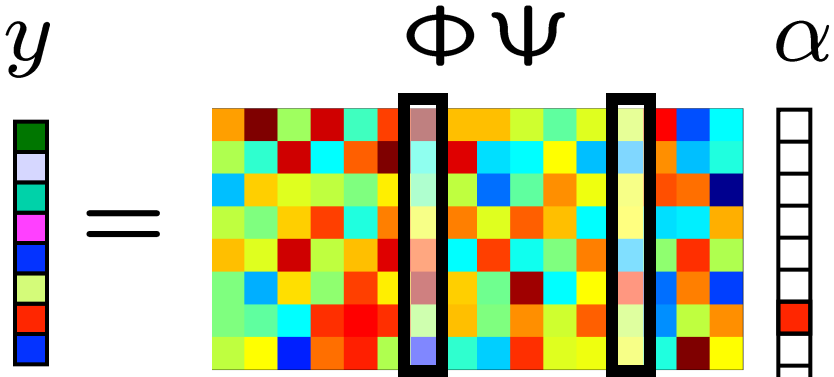
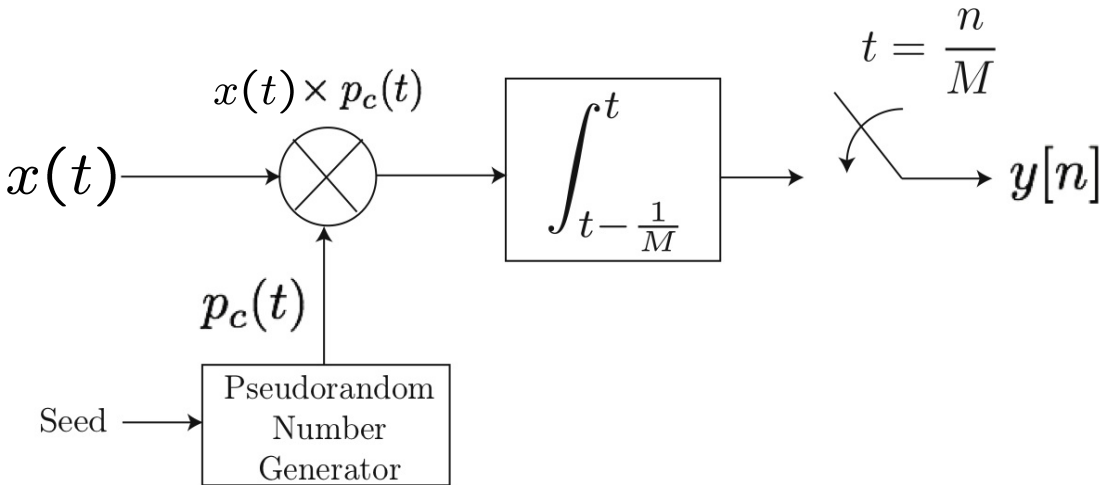
C

D

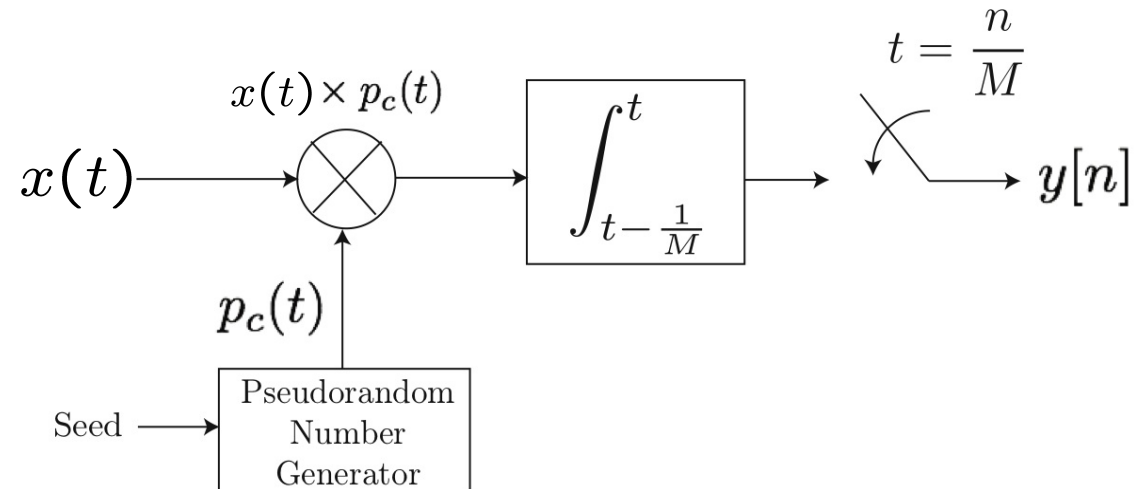
Random Demodulator



Random Demodulator



Sampling Rate



- **Goal:** Sample near signal's (low) "information rate" rather than its (high) Nyquist rate

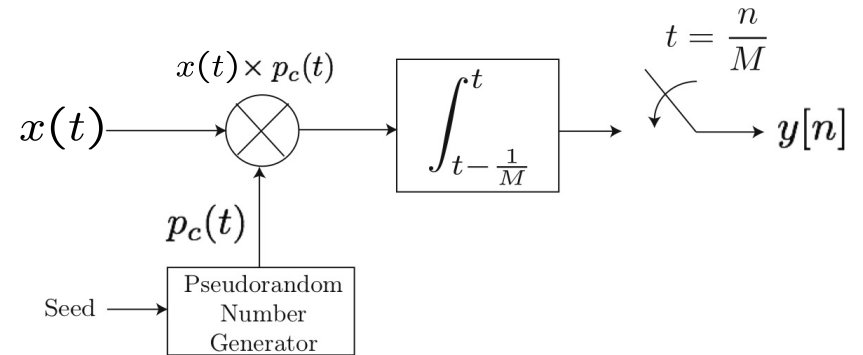
$$M = O(K \log(N/K))$$

A2I sampling rate

number of tones / window

Nyquist bandwidth

Sampling Rate



- **Theorem** [Tropp, B, et al 2007]

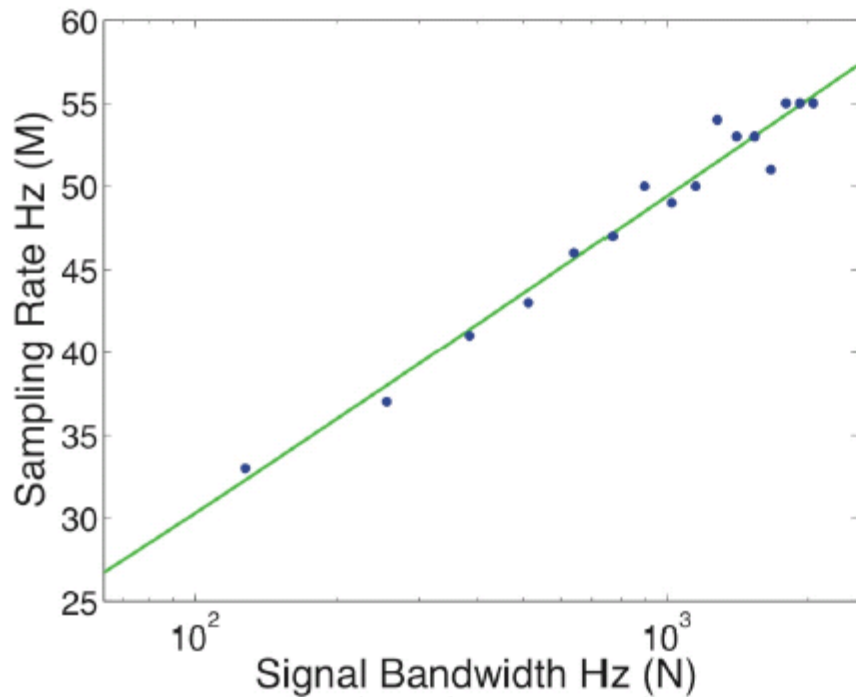
If the sampling rate satisfies

$$M > cK \log^2(N/\delta), \quad 0 < \delta < 1$$

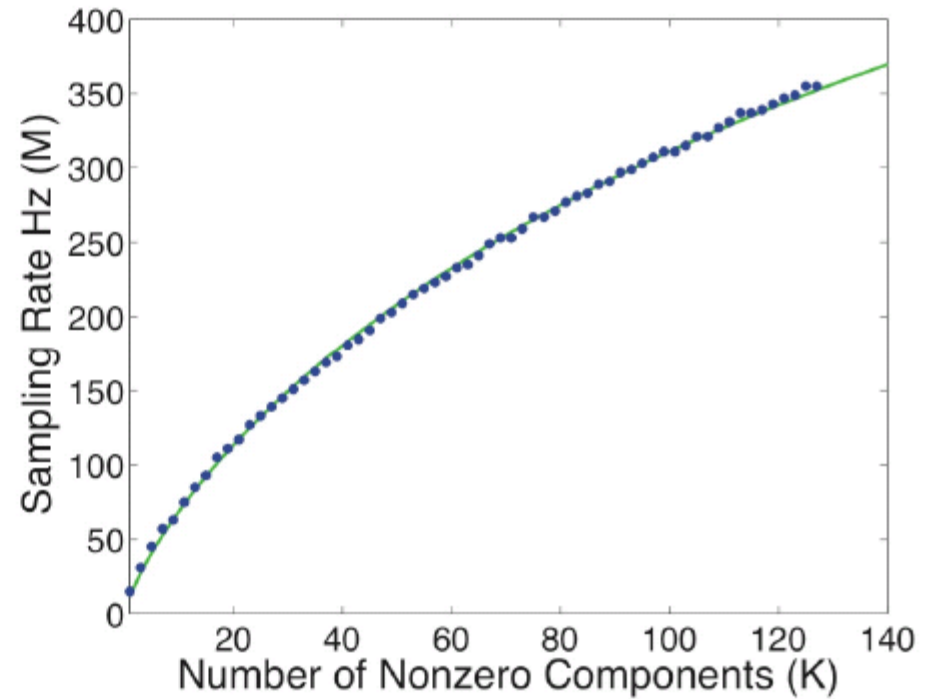
then locally Fourier K -sparse signals can be recovered exactly with probability

$$1 - \delta$$

Empirical Results



$$1.69K \log(N/K + 1) + 4.51$$

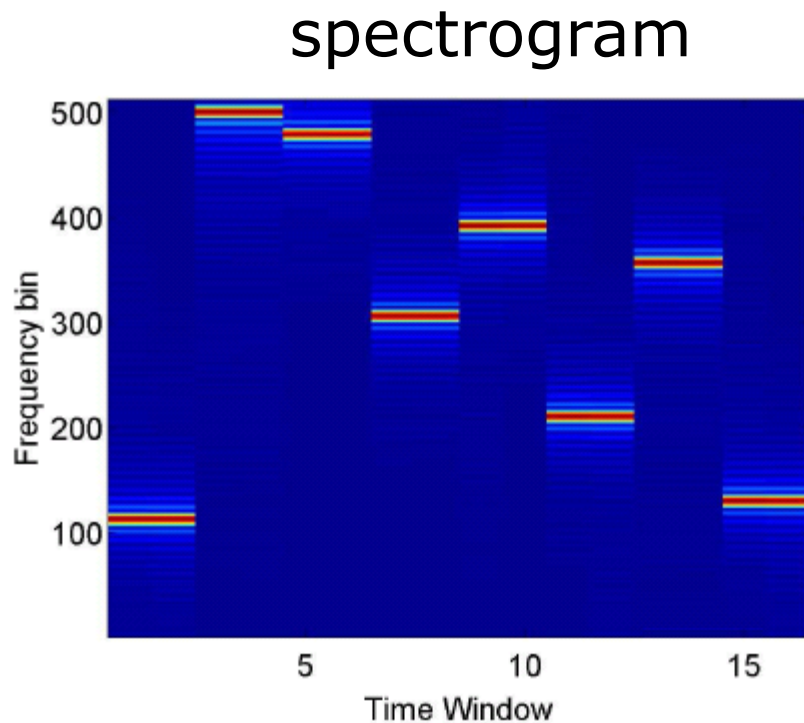


$$1.71K \log(N/K + 1) + 1$$

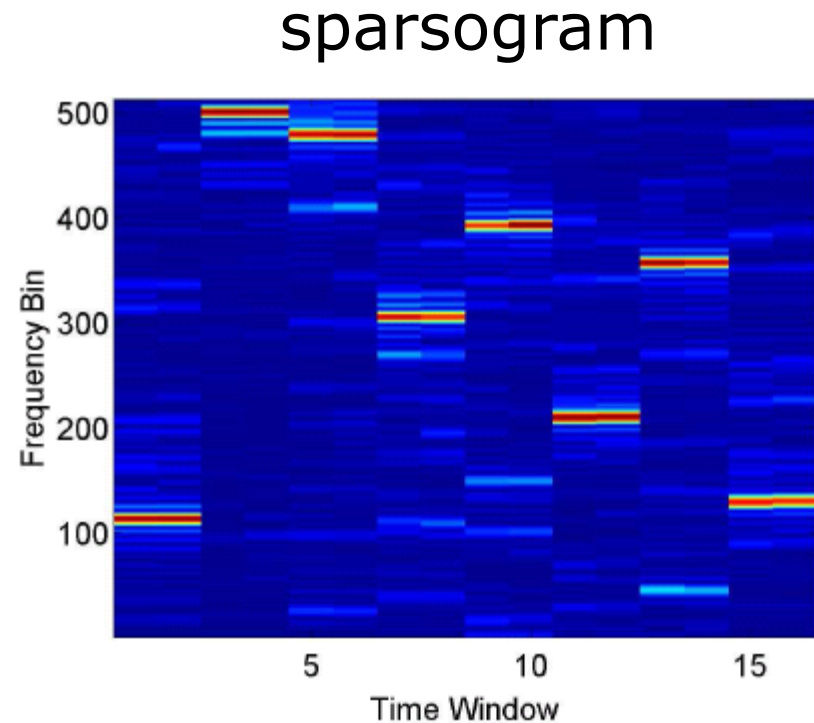
$$M \leq CK \log(N/K + 1)$$
$$C \sim 1.7$$

Example: Frequency Hopper

Nyquist rate sampling



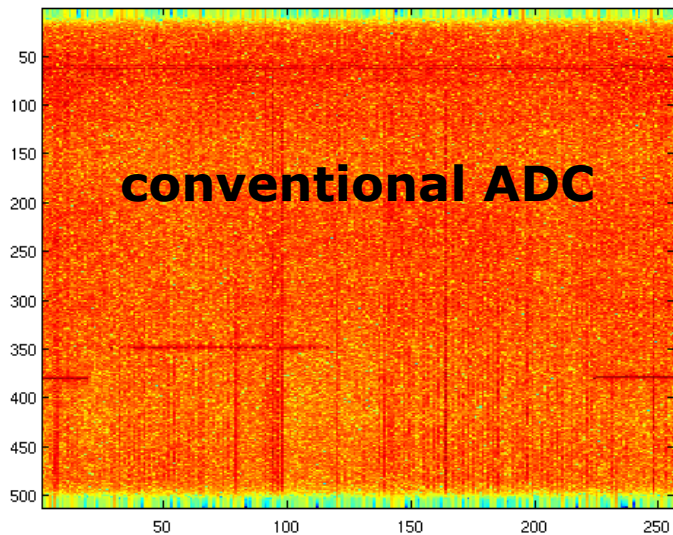
20x sub-Nyquist sampling



Example: Frequency Hopper

Nyquist rate sampling

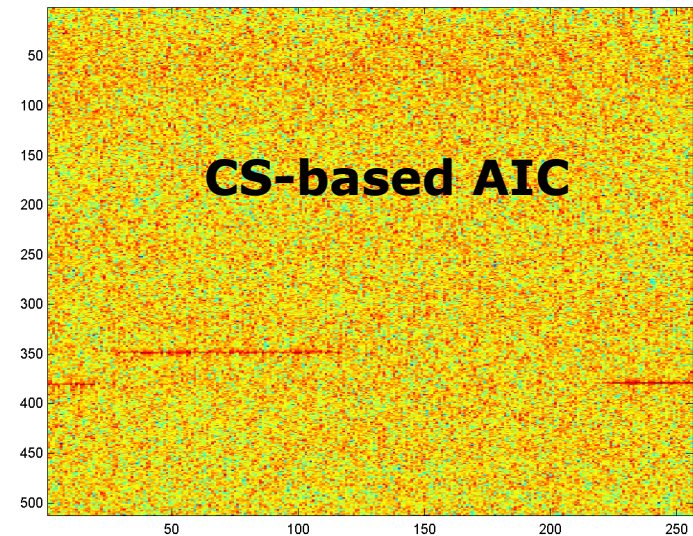
spectrogram



20MHz sampling rate

20x sub-Nyquist
sampling

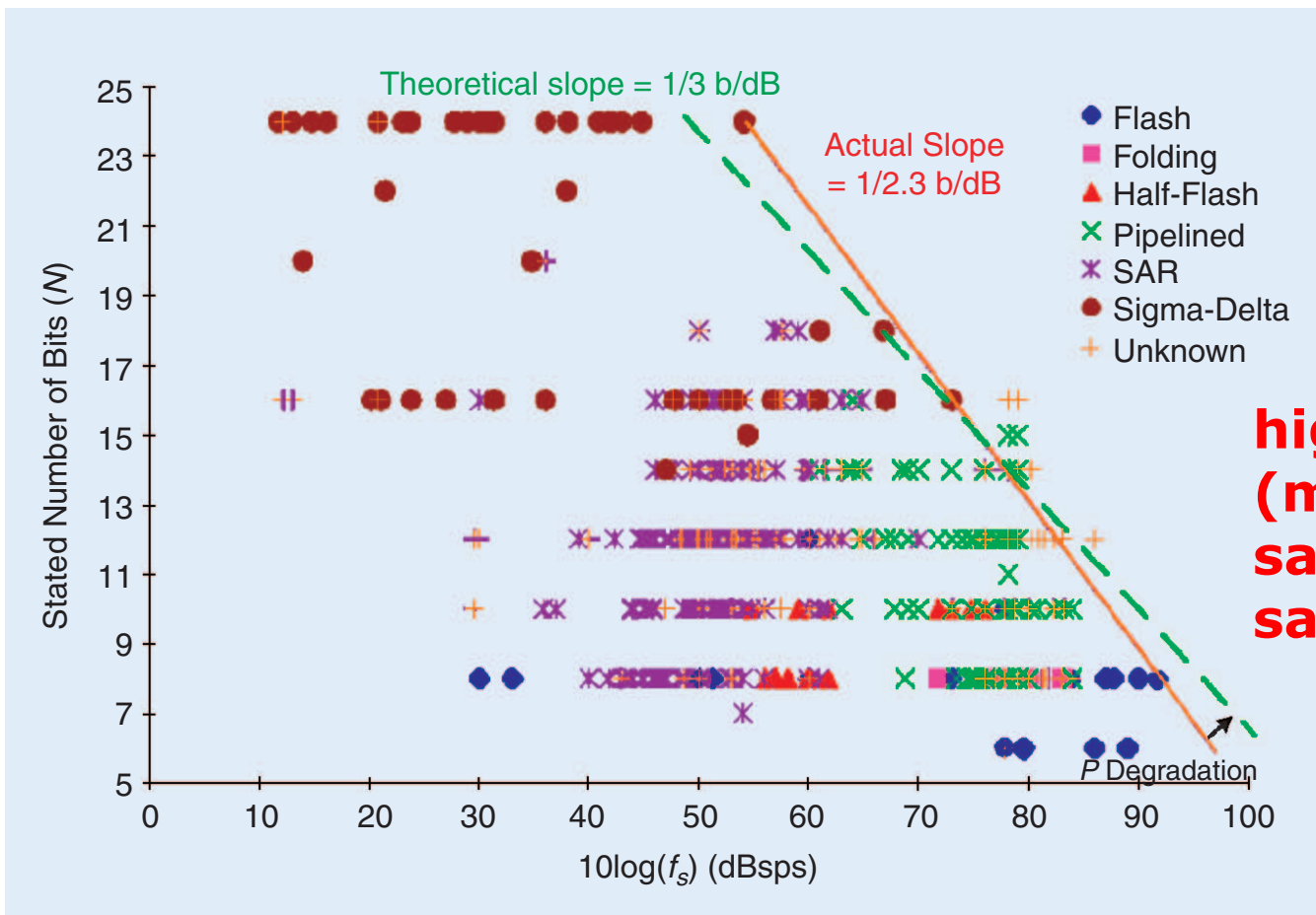
sparsogram



1MHz sampling rate

Dynamic Range

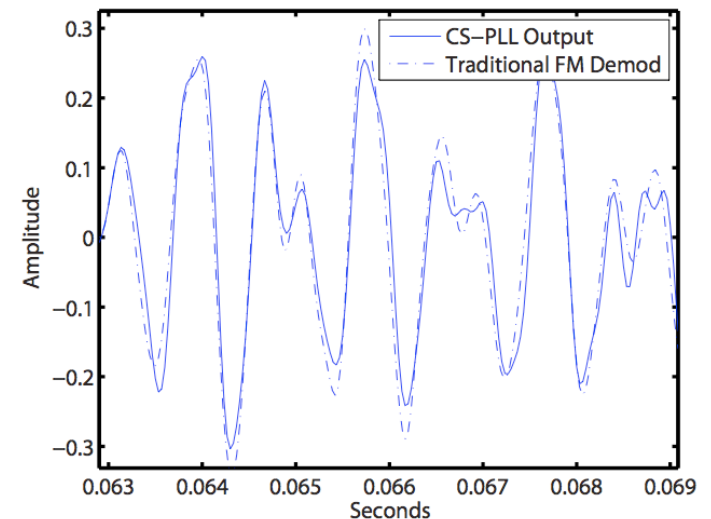
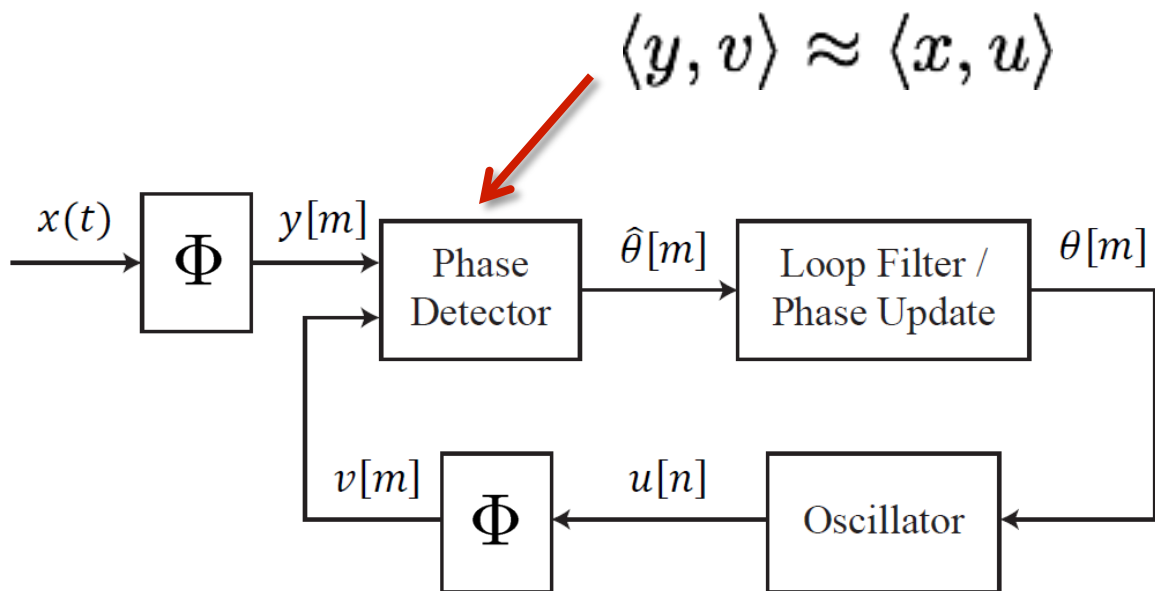
- **Key result:** Random measurements don't affect dynamic range



**high resolution
(many bits per
sample) only at low
sampling rate**

Application: Frequency Tracking

- Compressive **Phase Locked Loop (PLL)**
 - key idea: phase detector in PLL computes inner product between signal and oscillator output
 - RIP ensures we can compute this inner product between corresponding low-rate CS measurements



CS-PLL w/ 20x undersampling

Summary: CS

- **Compressive sensing**
 - randomized dimensionality reduction
 - exploits signal **sparsity** information
 - integrates sensing, compression, processing
- Why it works: **with high probability, random projections preserve information in signals with concise geometric structures**
- Enables new sensing architectures
 - ADCs, radios, cameras, ...
- Can process signals/images directly from their compressive measurements

Open Research Issues

- Links with **information theory**
 - new encoding matrix design via codes (LDPC, fountains)
 - new decoding algorithms (BP, etc.)
 - quantization and rate distortion theory
- Links with **machine learning**
 - Johnson-Lindenstrauss, manifold embedding, RIP
- **Processing/inference** on random projections
 - filtering, tracking, interference cancellation, ...
- **Multi-signal CS**
 - array processing, localization, sensor networks, ...
- **CS hardware**
 - ADCs, receivers, cameras, imagers, radars, ...



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